

# Bipolar Voltage Outputs for the TLV56xx Family of DACs

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Precision Digital-to-Analog Converters

## ABSTRACT

A method for generating a symmetrical, bipolar, output swing voltage from a TI TLV56xx-family [digital-to-analog converter](#) (DAC) by using a bipolar operational amplifier (op amp), [TLE2142](#), is presented. The resulting output voltage has a wide range that is limited only by the choice of op amp used for conditioning the DAC output signal. The example in this report realizes an output voltage range of  $\pm 13.8$  V for a 10-k $\Omega$  load.

## Design Problem

Some applications require digital-to-analog signal conversion with a bipolar output-voltage range. The output-voltage range of a standard unipolar DAC is generally between zero and  $2 \times V_{\text{ref}}$ ; however, it can easily be signal-conditioned to produce a bipolar range.

## Solution

The DAC's output voltage is:

$$\text{OUT} = 2V_{\text{ref}} \times \frac{\text{CODE}}{(0x1000)}$$

where *CODE* is the DAC's digital input, *OUT* is its analog output, and  $V_{\text{ref}}$  is the reference voltage, which may be already integrated into the DAC. Within the 12-bit [TLV56xx](#) family of DACs, *CODE* can have any value between 0x000 and 0xFFFF.

The conversion of a strictly non-negative voltage range into a symmetrical bipolar range is achieved using a standard op amp connected as a difference amplifier as shown in [Figure 1](#).

Referring to [Figure 1](#), the output voltage of the op amp  $A_1$  is:

$$V_O = \frac{R_4}{R_3 + R_4} \left( 1 + \frac{R_2}{R_1} \right) \text{OUT} - \frac{R_2}{R_1} V_{\text{ref}} \quad (1)$$

When  $R_2 / R_1 = R_4 / R_3$  the op amp works as a real differential amplifier and, in this case, [Equation 1](#) simplifies to:

$$V_O = \frac{R_2}{R_1} (\text{OUT} - V_{\text{ref}}) = A_{\text{DM}} (\text{OUT} - V_{\text{ref}})$$

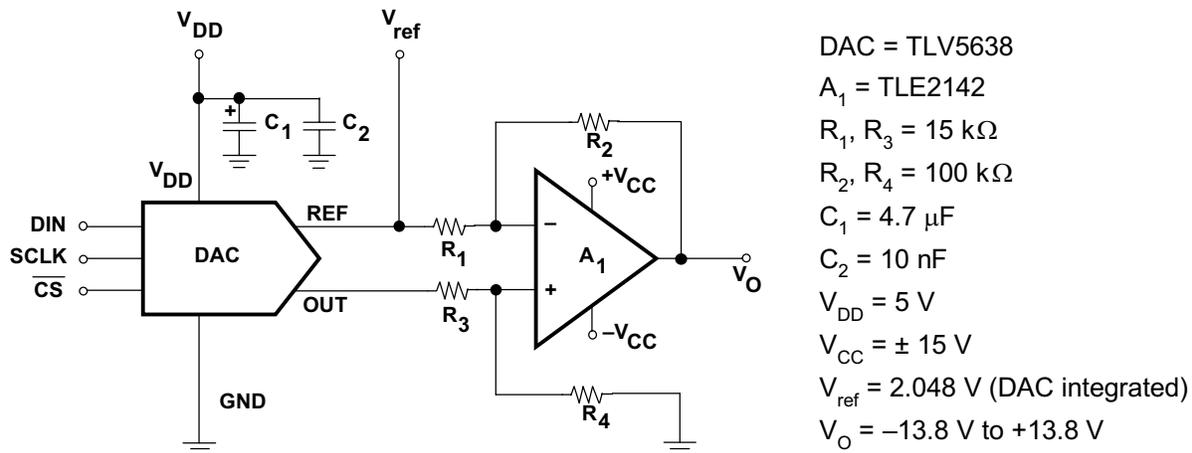


Figure 1. TLV56xx With Bipolar Output

$$\frac{R_4}{R_3} = (1 + x) \frac{R_2}{R_1}; \quad |x| \ll 1$$

In this case, Equation 1 becomes:

$$V_O = \frac{R_2}{R_1} \left[ \frac{\left(1 + \frac{R_2}{R_1}\right) (1 + x)}{1 + \frac{R_2}{R_1} (1 + x)} \text{OUT} - V_{\text{ref}} \right] \approx A_{\text{DM}} \left[ (\text{OUT} - V_{\text{ref}}) + \frac{\text{OUT}}{1 + \frac{R_2}{R_1}} x + O(x^2) \right]$$

When *OUT* and *V<sub>ref</sub>* share the common-mode voltage, *V<sub>CM</sub>*, the output voltage and the common-mode gain are nonzero and

$$A_{\text{CM}} = \frac{V_O}{V_{\text{CM}}} \approx \left( \frac{R_2}{R_1 + R_2} \right) x$$

The common-mode rejection ratio, CMRR, is then:

$$\text{CMRR} = \left| \frac{A_{\text{DM}}}{A_{\text{CM}}} \right| = \left( \frac{R_1 + R_2}{R_1} \right) \frac{1}{|x|} \approx \frac{R_2}{R_1} \times \frac{1}{|x|}; \quad R_2 \gg R_1$$

This result shows that it is crucial to choose very precise pairs of resistors to obtain an acceptably-high value of the common-mode rejection ratio.

## Conclusion

An easy, cost-effective method to generate bipolar outputs from a DAC is by using a bipolar difference amplifier to condition the DAC's output signal. The output voltage range depends mainly on the choice of op amp and its resistors. However, an acceptable common-mode rejection ratio can be obtained only by using resistor pairs of very high accuracy. Therefore, for those applications that are CMRR critical, an instrumental amplifier should be used instead.

## Revision History

Changes from Original (December 2000) to A Revision	Page
• Changed format to current TI application report template. ....	1

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

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