

## ALM2403-Q1

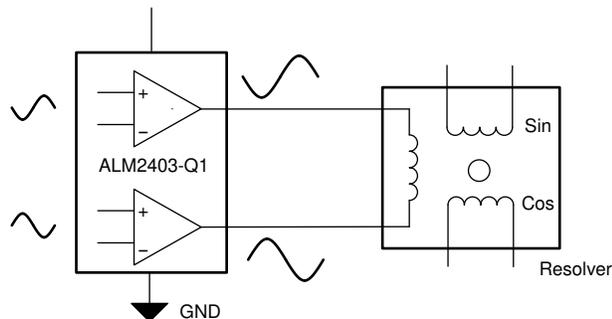
# レゾルバ駆動向け保護機能搭載、車載用、低歪み、デュアル・チャンネル・オペアンプ

### 1 特長

- 車載アプリケーション用に AEC-Q100 認定済み:
  - 温度グレード 1:  $T_A = -40^{\circ}\text{C} \sim +125^{\circ}\text{C}$
- 機能安全に対応
  - 機能安全システムの設計に役立つ資料を利用可能
- 大出力電流の駆動: ピーク 500mA (チャンネル毎)
  - ディスクリート・オペアンプおよびトランジスタの代替
- 両方の電源について広い電源電圧範囲 (最高 24V)
- 過熱シャットダウン
- 電流制限
- 低  $I_Q$  アプリケーション用のシャットダウン・ピン
- 21MHz のゲイン帯域幅、50V/ $\mu\text{s}$  のスルーレート
- パッケージ: 14 ピン HTSSOP (PWP)

### 2 アプリケーション

- レゾルバを使用した車載用および産業用アプリケーション
- インバータおよびモーター制御
- ブレーキ・システム
- 電動パワー・ステアリング (EPS)
- リアビュー・ミラー・モジュール
- 車載用電子ミラー
- サーボ・ドライブの電力段モジュール
- 飛行制御システム



簡略回路図

### 3 概要

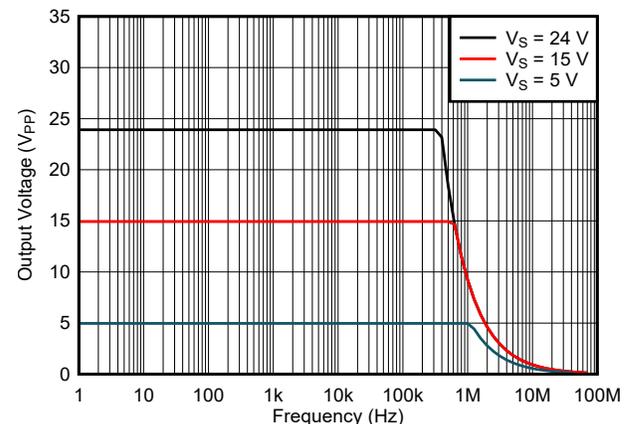
ALM2403-Q1 は、レゾルバを使用したアプリケーションに最適な機能と性能を備えたデュアルパワー・オペアンプです。大出力電流を連続的に駆動できると同時に、デバイスのゲイン帯域幅が広く、スルーレートが大きいため、レゾルバの 1 次コイルの励起に必要な低歪みの差動大振幅励磁の実現に適しています。電流制限と過熱検出により、特にフォルトが発生しやすい有線によるアナログ信号の駆動時に、システム全体の堅牢性を高めることができます。

サーマル・パッド付きで  $R_{\theta JA}$  が小さい小型の HTSSOP パッケージにより、基板面積を最小化すると同時に大電流を負荷に供給できます。ALM2403-Q1 のゲイン帯域幅が広いこと、高出力駆動能力を保持した状態で、デバイスをフィルタ段として構成できます。その結果、レゾルバ駆動のシグナル・チェーン全体のソリューション・サイズを大幅に縮小できます。このソリューション・サイズの縮小は、ALM2403-Q1 を車載用および産業用アプリケーションで使用する場合の主な利点のひとつです。

#### パッケージ情報

部品番号	パッケージ <sup>(1)</sup>	本体サイズ (公称)
ALM2403-Q1	HTSSOP (14)	5.00mm × 4.40mm

- (1) 利用可能なパッケージについては、データシートの末尾にあるパッケージ・オプションについての付録を参照してください。



出力電圧と周波数との関係



## Table of Contents

1 特長.....	1	7.4 Device Functional Modes.....	14
2 アプリケーション.....	1	<b>8 Application and Implementation.....</b>	<b>15</b>
3 概要.....	1	8.1 Application Information.....	15
4 Revision History.....	2	8.2 Typical Application.....	15
5 Pin Configuration and Functions.....	3	8.3 Power Supply Recommendations.....	19
6 Specifications.....	4	8.4 Layout.....	19
6.1 Absolute Maximum Ratings.....	4	<b>9 Device and Documentation Support.....</b>	<b>22</b>
6.2 ESD Ratings.....	4	9.1 Documentation Support.....	22
6.3 Recommended Operating Conditions.....	4	9.2 ドキュメントの更新通知を受け取る方法.....	22
6.4 Thermal Information.....	4	9.3 サポート・リソース.....	22
6.5 Electrical Characteristics.....	5	9.4 Trademarks.....	22
6.6 Typical Characteristics.....	7	9.5 静電気放電に関する注意事項.....	22
7 Detailed Description.....	12	9.6 用語集.....	22
7.1 Overview.....	12	<b>10 Mechanical, Packaging, and Orderable Information.....</b>	<b>22</b>
7.2 Functional Block Diagram.....	12		
7.3 Feature Description.....	13		

## 4 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from Revision * (November 2020) to Revision A (May 2022)	Page
• 「特長」の出力電流を 650mA から 500mA に変更.....	1
• 表紙のプロットでタイトルと Y 軸の単位を変更.....	1
• Changed pin names to synchronize pin naming throughout document.....	3
• Changed thermal pad description text for clarity.....	3
• Changed voltage range for $V_{OTF/SH\_DN}$ in the <i>Absolute Maximum Ratings</i> .....	4
• Changed all $V_S$ voltages to single-supply nomenclature in the <i>Electrical Characteristics and Typical Characteristics</i> .....	5
• Deleted test conditions from enable high and low input voltages in the <i>Electrical Characteristics</i> .....	5
• Moved shutdown current parameter to Power Supply section in the <i>Electrical Characteristics</i> .....	5
• Changed Figures 6-12 through 6-16 to correct axis units and values.....	7
• Changed functional block diagram to correct inaccuracies.....	12
• Changed EMC capacitance from 50 nF to 10 nF in Table 8-1, <i>Design Parameters</i> .....	16
• Added test condition to first bullet of <i>Detailed Design Procedure</i> .....	17
• Changed R3 to R2 in 2nd paragraph of <i>Filter Design</i> section.....	17
• Changed terms in Equation 4 to Equation 6 for clarity.....	18
• Changed Figure 8-4, <i>2nd-Order MFB LP Filter AC Output Characteristics</i> .....	19
• Changed values in Table 8-2, <i>Signal Attenuation vs Frequency</i> .....	19
• Changed Figure 8-6, <i>ALM2403-Q1 Layout Example</i> , to match EVM layout.....	20

## 5 Pin Configuration and Functions

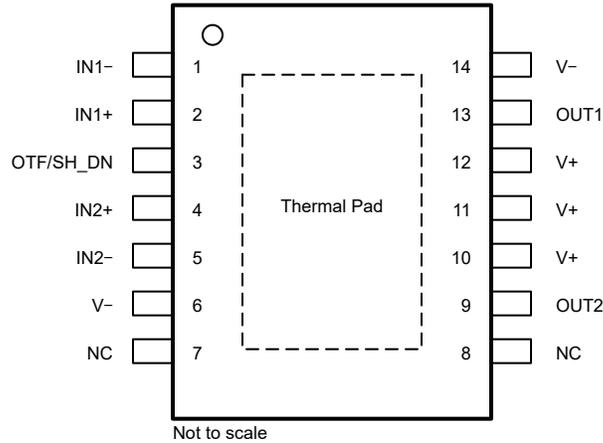


图 5-1. PWP Package, 14-Pin HTSSOP (Top View)

表 5-1. Pin Functions

PIN		TYPE	DESCRIPTION
NO.	NAME		
1	IN1-	Input	Inverting op amp input for channel 1
2	IN1+	Input	Noninverting op amp input for channel 1
3	OTF/SH_DN	Input/Output	Overtemperature flag and shutdown (see 表 7-1, Shutdown Truth Table)
4	IN2+	Input	Noninverting op amp input for channel 2
5	IN2-	Input	Inverting op amp input for channel 2
6, 14	V-	—	Negative supply pin (both negative supply pins must be used and connected together)
7, 8	NC	—	No internal connection (do not connect)
9	OUT2	Output	Op amp output for channel 2
10, 11, 12	V+	—	Positive supply pin
13	OUT1	Output	Op amp output for channel 1
Thermal Pad	Thermal Pad	—	Connect the exposed thermal pad to the most negative supply on the device, V-, for best thermal performance. The thermal pad can also be left floating electrically; the heat spread of the pad can be thermally maximized and conducted into the PCB.

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

			MIN	MAX	UNIT
V <sub>S</sub>	Supply voltage	Single-supply, V <sub>S</sub> = (V+) – GND	26		V
		Dual-supply, V <sub>S</sub> = (V+) – (V–)	±13		
	Signal input voltage	Common-mode	(V–) – 0.7	(V+) + 0.7	V
		Differential	(V+) – (V–) + 0.2		
V <sub>OTF/SH_DN</sub>	OTF/SH_DN pin voltage		(V–) – 0.2	(V–) + 5.7	V
	Signal input current		±10		mA
	Output short circuit <sup>(2)</sup>		Continuous	Continuous	
T <sub>A</sub>	Operating temperature		–55	150	°C
T <sub>J</sub>	Junction temperature			150	°C
T <sub>stg</sub>	Storage temperature		–65	150	°C

- (1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
- (2) Short-circuit to ground, one amplifier per package.

### 6.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup> HBM ESD classification level 2	±2000	V
		Charged-device model (CDM), per AEC Q100-011 CDM ESD classification level C5	±750	

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V <sub>S</sub>	Supply voltage	Single-supply, V <sub>S</sub> = (V+) – GND	5		24	V
		Dual-supply, V <sub>S</sub> = (V+) – (V–)	±2.5		±12	
T <sub>A</sub>	Operating temperature		–40		125	°C

### 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		ALM2403-Q1	UNIT
		PWP (HTSSOP)	
		14 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	46.9	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	42.1	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	22.6	°C/W
Ψ <sub>JT</sub>	Junction-to-top characterization parameter	1.2	°C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	22.5	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	5.9	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 6.5 Electrical Characteristics

at  $T_A = 25^\circ\text{C}$ ,  $V_S = V_+ = 24\text{ V}$ ,  $V_- = \text{GND}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>OFFSET VOLTAGE</b>						
$V_{OS}$	Input offset voltage			$\pm 6$	$\pm 25$	mV
$dV_{OS}/dT$	Input offset voltage drift	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		$\pm 15$	$\pm 50$	$\mu\text{V}/^\circ\text{C}$
PSRR	Power-supply rejection ratio	$V_S = 5\text{ V}$ to $24\text{ V}$		$\pm 10$	$\pm 47$	$\mu\text{V}/\text{V}$
		$V_S = 5\text{ V}$ to $24\text{ V}$ , $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			$\pm 50$	
	Channel separation	$f = 10\text{ kHz}$		120		dB
<b>INPUT BIAS CURRENT</b>						
$I_B$	Input bias current			10	$\pm 100$	pA
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$				$\pm 100$
$I_{OS}$	Input offset current			10	$\pm 200$	pA
		$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$				$\pm 100$
<b>NOISE</b>						
	Input voltage noise	$f = 0.1\text{ Hz}$ to $10\text{ Hz}$		8		$\mu\text{V}_{RMS}$
$e_N$	Input voltage noise density	$f = 1\text{ kHz}$		150		$\text{nV}/\sqrt{\text{Hz}}$
		$f = 100\text{ kHz}$		22		
$i_N$	Input current noise	$f = 1\text{ kHz}$		48		$\text{fA}/\sqrt{\text{Hz}}$
<b>INPUT VOLTAGE</b>						
$V_{CM}$	Common-mode voltage		$(V_-) - 0.2$		$(V_+) + 0.2$	V
CMRR	Common-mode rejection ratio	$(V_-) - 0.5\text{ V} < V_{CM} < (V_+) + 0.5\text{ V}$ , $10\text{ V} \leq V_S < 24\text{ V}$	49	72		dB
		$(V_-) - 0.2\text{ V} < V_{CM} < (V_+) + 0.2\text{ V}$ , $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $10\text{ V} < V_S < 24\text{ V}$	52			
		$(V_-) + 2.5\text{ V} < V_{CM} < (V_+) - 2.5\text{ V}$ , $10\text{ V} < V_S < 24\text{ V}$	80	94		
		$(V_-) + 2.5\text{ V} < V_{CM} < (V_+) - 2.5\text{ V}$ , $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $10\text{ V} < V_S < 24\text{ V}$	75			
		$(V_-) - 0.5\text{ V} < V_{CM} < (V_+) + 0.5\text{ V}$ , $5\text{ V} < V_S < 24\text{ V}$	44	59		
<b>INPUT CAPACITANCE</b>						
$Z_{ID}$	Differential			$1 \parallel 2$		$\text{G}\Omega \parallel \text{pF}$
$Z_{ICM}$	Common-mode			$1 \parallel 2$		
<b>OPEN-LOOP GAIN</b>						
$A_{OL}$	Open-loop voltage gain	$(V_-) + 0.5\text{ V} < V_O < (V_+) - 0.5\text{ V}$ , $V_S = 24\text{ V}$		103	111	dB
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	96		
		$(V_-) + 1.5\text{ V} < V_O < (V_+) - 1.5\text{ V}$ , $R_L = 225\ \Omega$ , $V_S = 24\text{ V}$		96	104	
			$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	94		
<b>FREQUENCY RESPONSE</b>						
GBW	Gain-bandwidth product	$V_S = 24\text{ V}$		21		MHz
SR	Slew rate	10-V step, gain = +1		50		$\text{V}/\mu\text{s}$
$t_s$	Settling time	To 0.1%, 10-V step, gain = +1, $C_L = 10\text{ pF}$		0.31		$\mu\text{s}$
		To 0.1%, 10-V step, gain = -1, $C_L = 10\text{ pF}$		0.40		
	Overload recovery time	$V_{IN} \times \text{gain} > V_S$		0.28		$\mu\text{s}$
THD+N	Total harmonic distortion + noise	$V_S = 15\text{ V}$ , $V_O = 10\text{ V}_{pp}$ , gain = -1, $f = 10\text{ kHz}$ , $R_L = 100\ \Omega$		74		dB

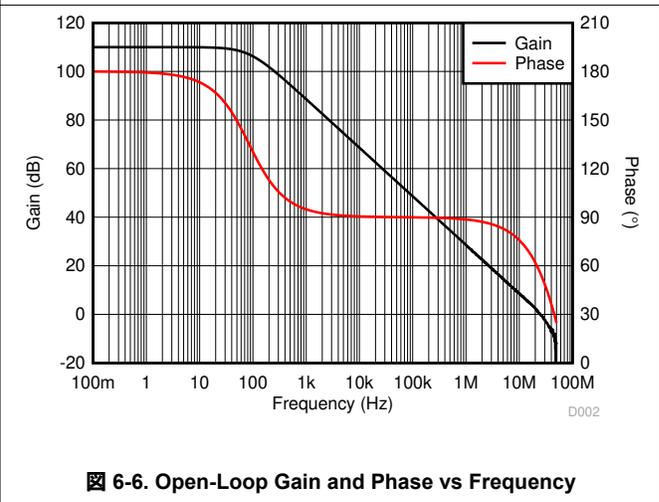
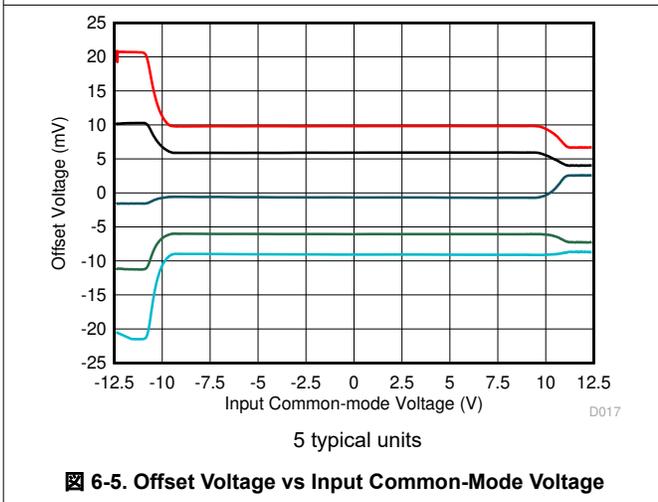
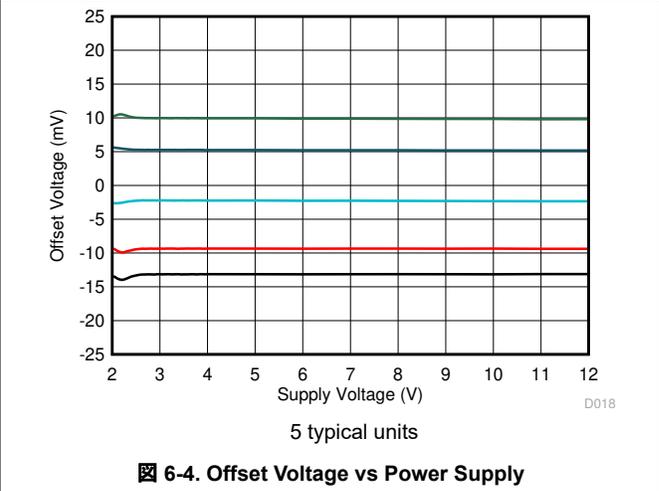
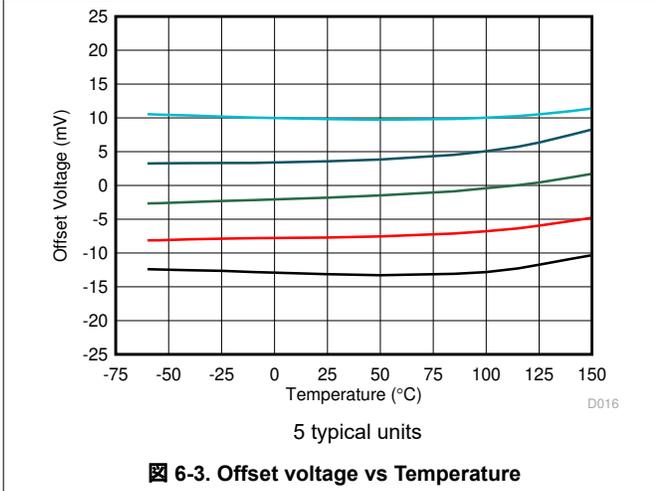
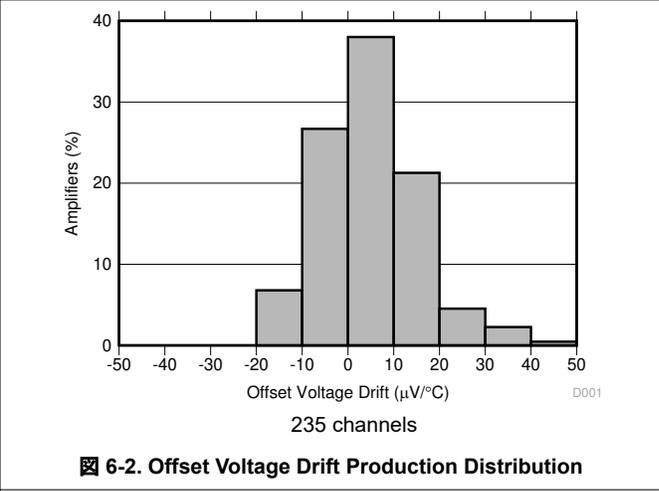
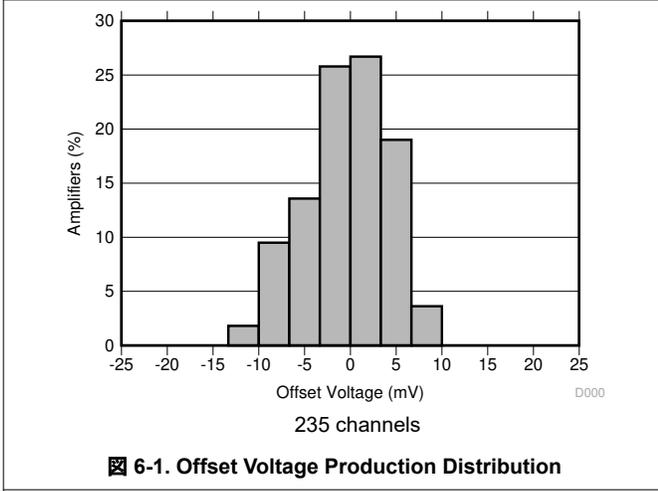
## 6.5 Electrical Characteristics (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_S = V_+ = 24\text{ V}$ ,  $V_- = \text{GND}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S / 2$ , and  $V_{CM} = V_{OUT} = V_S / 2$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>OUTPUT</b>						
	Voltage output swing from rail	$I_{OUT} = \pm 5\text{ mA}$		35	60	mV
$I_{SC}$	Short-circuit current	Sinking		400		mA
		Sourcing		500		
<b>ENABLE</b>						
$V_{IH\_OTF}$	Enable high input voltage		1.2			V
$V_{IL\_OTF}$	Enable low input voltage				0.5	V
	Enable hysteresis			220		mV
$t_{OTF/SH\_DN}$	Enable start-up time			5		$\mu\text{s}$
<b>POWER SUPPLY</b>						
$I_Q$	Total quiescent current	$I_O = 0\text{ A}$		3.6	5.5	mA
		$I_O = 0\text{ A}$ , $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			6	
$I_{SD}$	Shutdown current	$V_{OTF/SH\_DN} = 0\text{ V}$			260	$\mu\text{A}$
<b>TEMPERATURE</b>						
	Thermal shutdown			172		$^\circ\text{C}$
	Thermal shutdown recovery			150		$^\circ\text{C}$

## 6.6 Typical Characteristics

at  $T_A = 25^\circ\text{C}$ ,  $V_S = 24\text{ V}$ ,  $V_{CM} = V_S/2$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted)



### 6.6 Typical Characteristics (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_S = 24\text{ V}$ ,  $V_{CM} = V_S/2$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted)

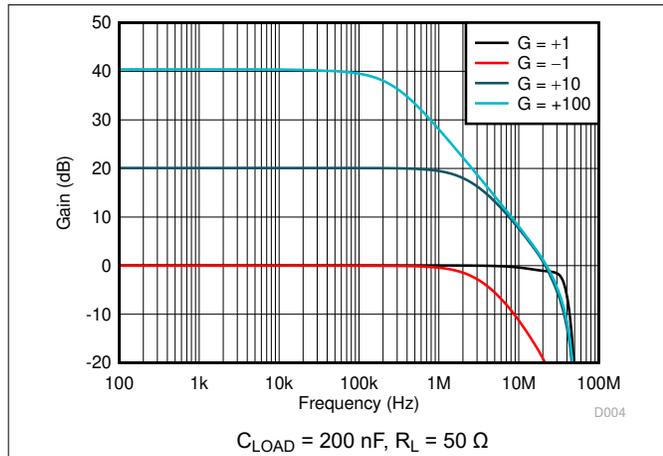


Figure 6-7. Closed-Loop Gain vs Frequency

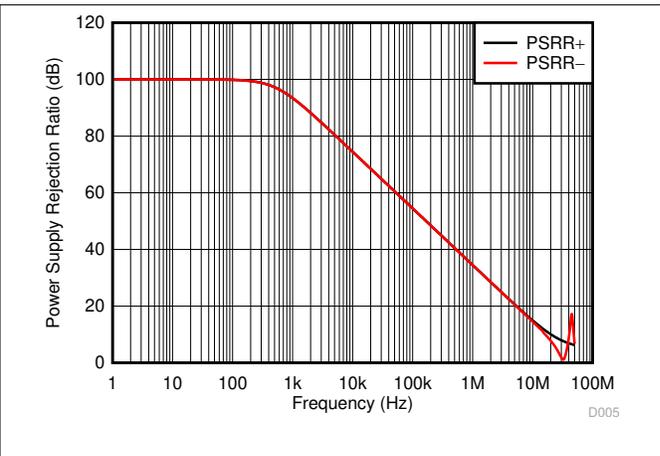


Figure 6-8. PSRR vs Frequency

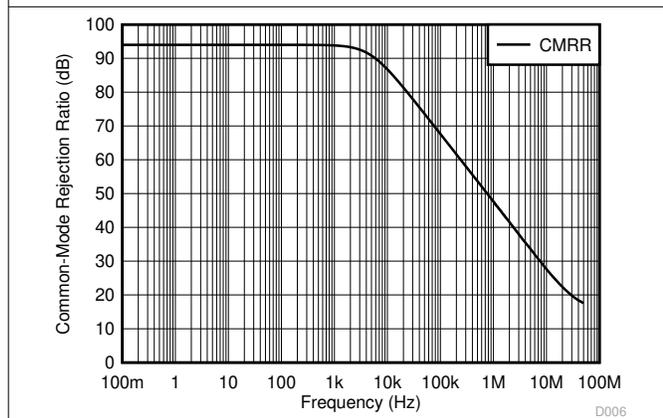


Figure 6-9. CMRR vs Frequency

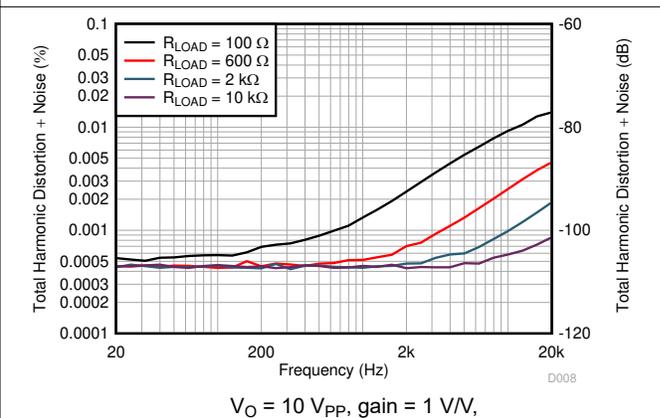


Figure 6-10. THD+N Ratio vs Frequency

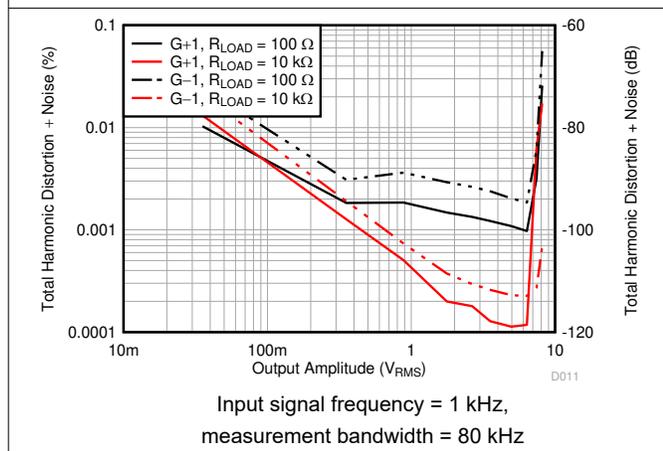


Figure 6-11. THD+N vs Output Amplitude

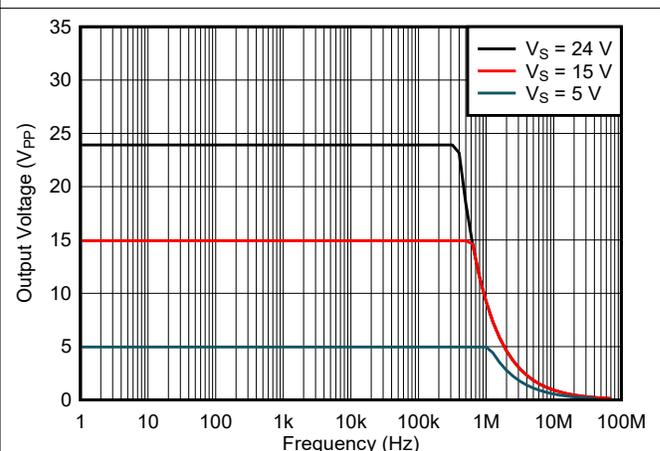
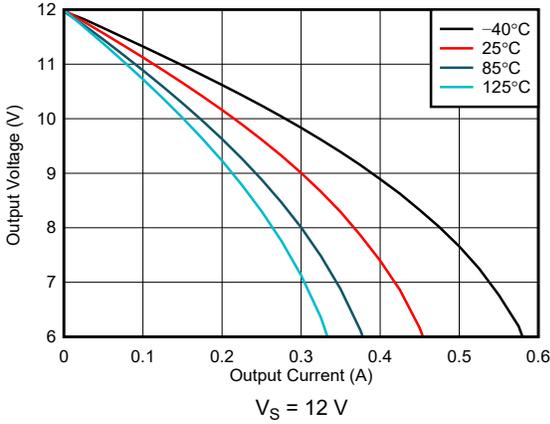


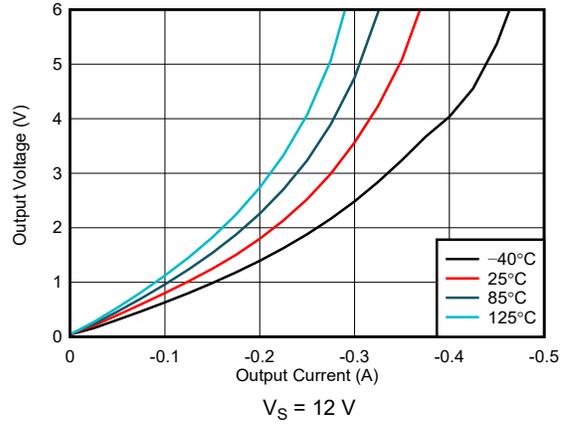
Figure 6-12. Output Voltage vs Frequency

### 6.6 Typical Characteristics (continued)

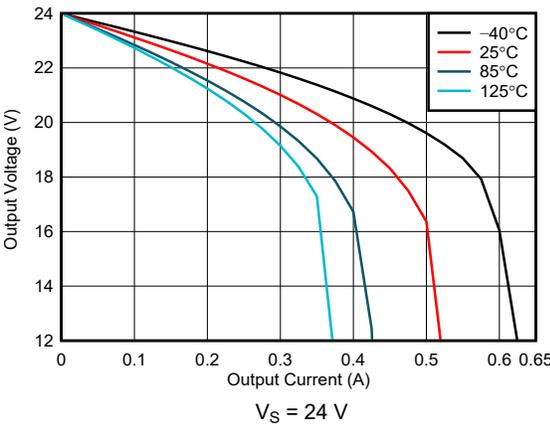
at  $T_A = 25^\circ\text{C}$ ,  $V_S = 24\text{ V}$ ,  $V_{CM} = V_S/2$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted)



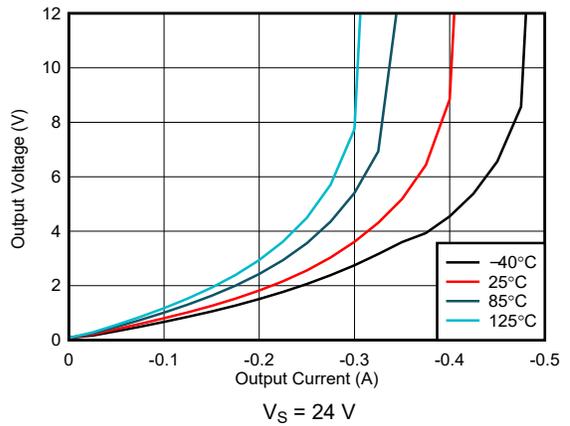
6-13. Output Voltage Swing vs Output Source Current



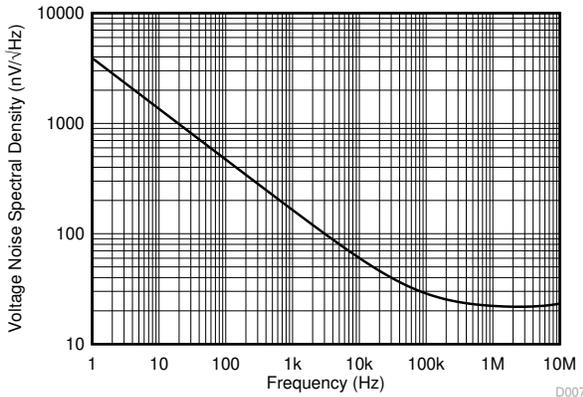
6-14. Output Voltage Swing vs Output Sink Current



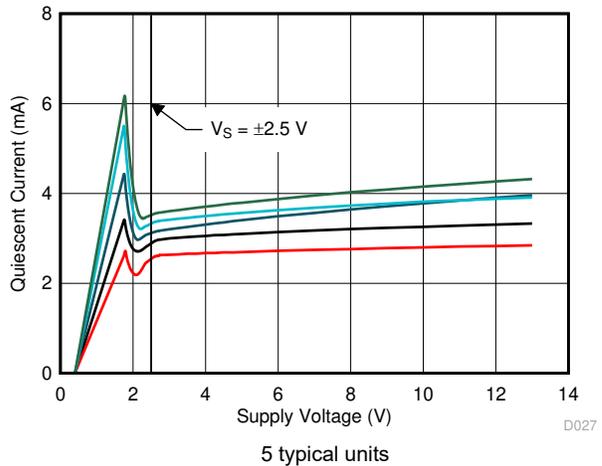
6-15. Output Voltage Swing vs Output Source Current



6-16. Output Voltage Swing vs Output Sink Current



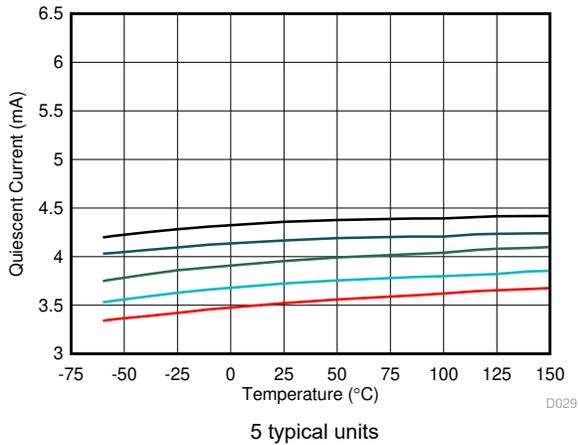
6-17. Input Voltage Spectral Noise Density vs Frequency



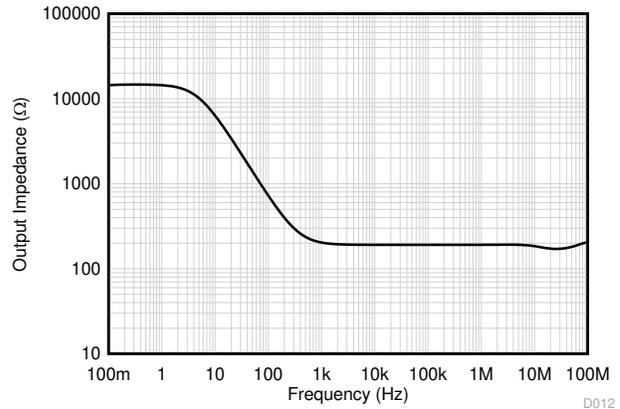
6-18. Quiescent Current vs Power Supply

### 6.6 Typical Characteristics (continued)

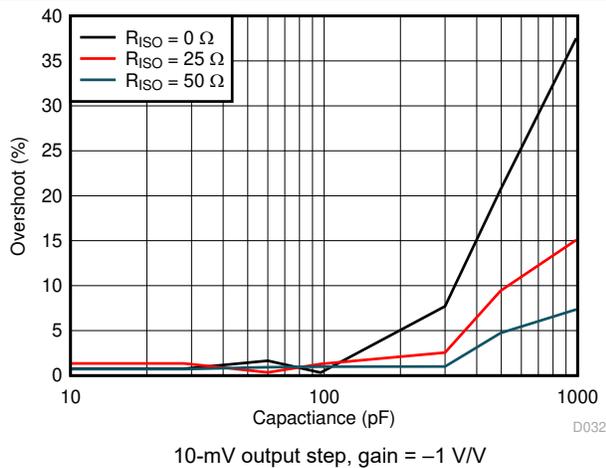
at  $T_A = 25^\circ\text{C}$ ,  $V_S = 24\text{ V}$ ,  $V_{CM} = V_S/2$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted)



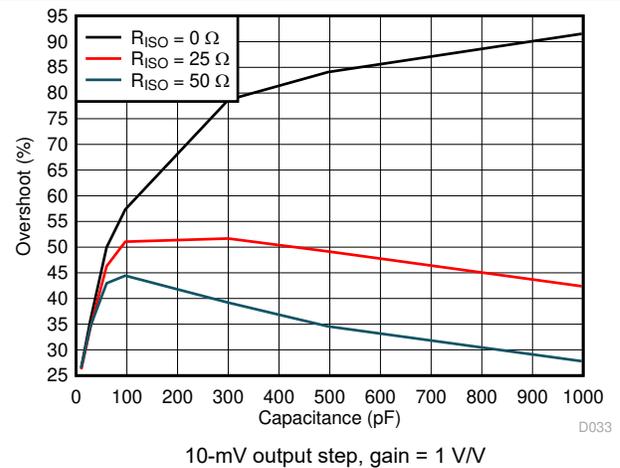
6-19. Quiescent Current vs Temperature



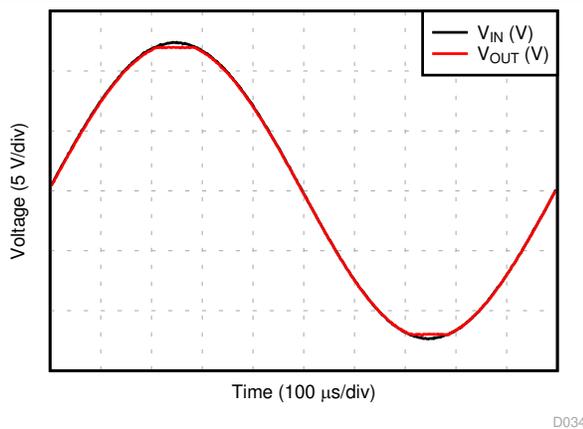
6-20. Open-Loop Output Impedance vs Frequency



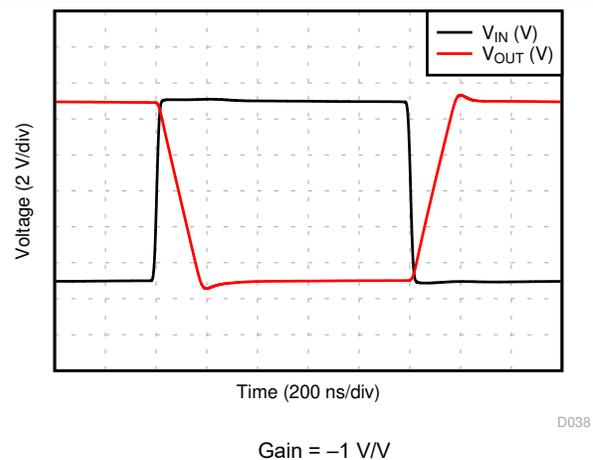
6-21. Small-Signal Overshoot vs Capacitive Load



6-22. Small-Signal Overshoot vs Capacitive Load



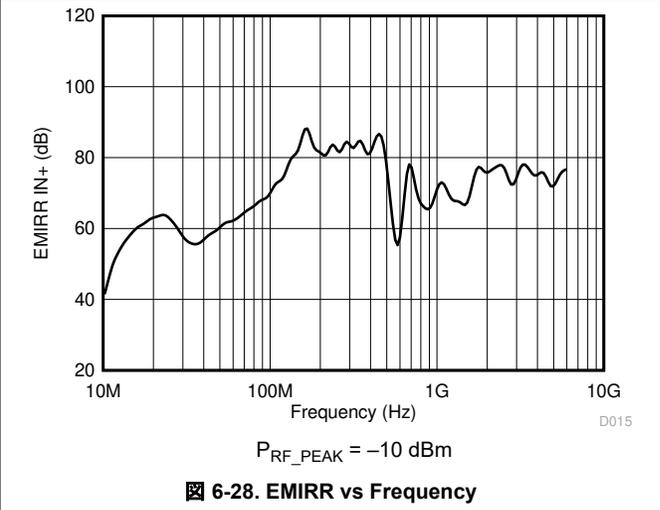
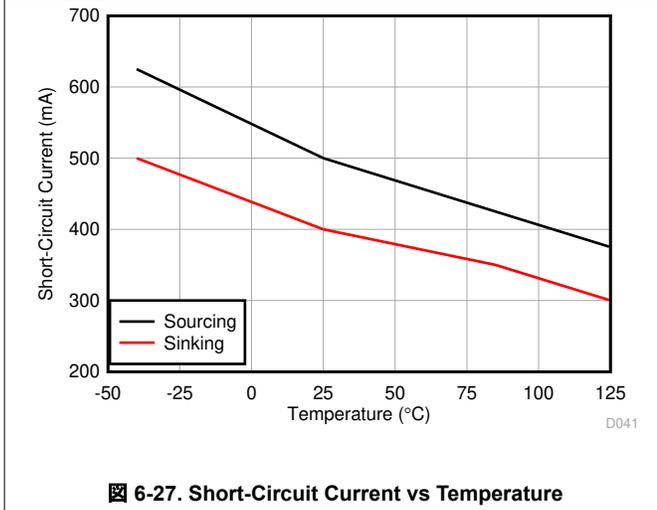
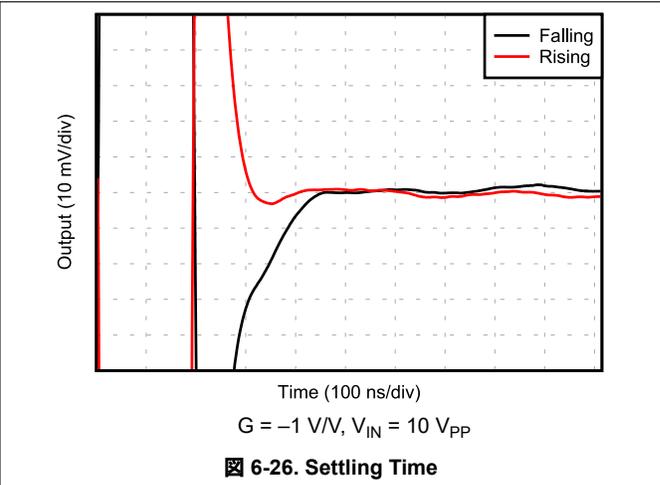
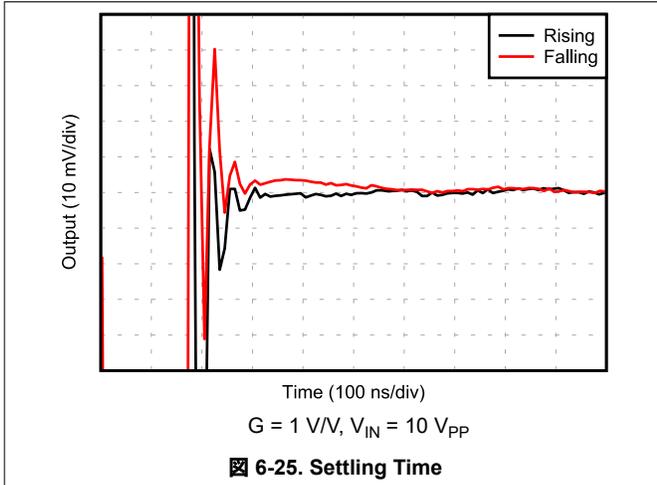
6-23. No Phase Reversal



6-24. Large-Signal Step Response

## 6.6 Typical Characteristics (continued)

at  $T_A = 25^\circ\text{C}$ ,  $V_S = 24\text{ V}$ ,  $V_{CM} = V_S/2$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted)

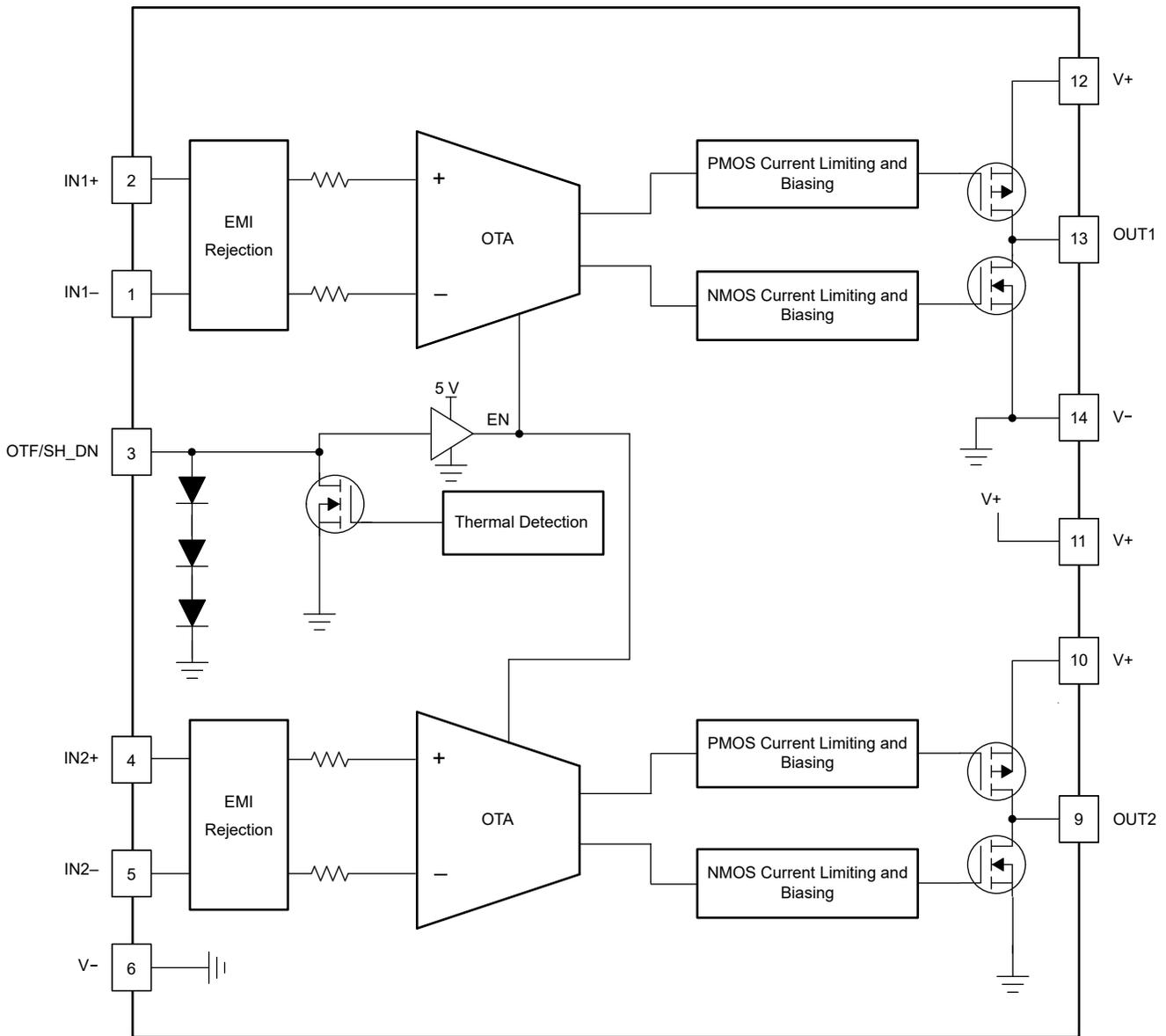


## 7 Detailed Description

### 7.1 Overview

The ALM2403-Q1 is a dual-power op amp qualified for use in automotive applications. Key features for this device are low offset voltage, high output current drive capability, and high FPBW capability. The device also offers protection features such as thermal shutdown and current limit. The 14-pin HTSSOP package minimizes board space and power dissipation.

### 7.2 Functional Block Diagram



## 7.3 Feature Description

### 7.3.1 Overtemperature and Shutdown Pin (OTF/SH\_DN)

The overtemperature and shutdown pin, OTF/SH\_DN, is bidirectional and allows both op amps to be put into a low  $I_Q$  state (approximately 200  $\mu\text{A}$  per amplifier) when forced low or to less than  $V_{IL\_OTF}$ . As a result of being bidirectional, and the respective enable and disable functionality, this pin must be pulled high or greater than  $V_{IH\_OTF}$  through a pullup resistor. The use of a 10-k $\Omega$  pullup resistor leads to a drive current of approximately 210  $\mu\text{A}$  when used with a pullup voltage of 3.3 V.

When the junction temperature of the ALM2403-Q1 exceeds the specified limits, OTF/SH\_DN goes low to alert the application that both the outputs have turned off because of an overtemperature event.

When OTF/SH\_DN is pulled low and the op amps are shut down, the op amps are in an open loop, even when there is negative feedback applied. This occurrence is due to the loss of the open-loop gain in the op amps when the biasing is disabled.

### 7.3.2 Thermal Shutdown

If the die temperature exceeds safe limits, all outputs are disabled, and the OTF/SH\_DN pin is driven low. After the die temperature has fallen to a safe level, operation automatically resumes. The OTF/SH\_DN pin is released after operation has resumed.

When operating the die at a high temperature, the op amp toggles on and off between the thermal shutdown hysteresis. In this event, the safe limits for the die temperature must be taken in to account. Do not continuously operate the device in thermal hysteresis for long periods of time.

### 7.3.3 Current-Limit and Short-Circuit Protection

Each op amp in the ALM2403-Q1 has separate internal current limiting for the PMOS (high-side) and NMOS (low-side) output transistors. If the output is shorted to ground, then the PMOS (high-side) current limit is activated, and limits the current to 500 mA nominally. If the output is shorted to supply, then the NMOS (low-side) current limit is activated and limits the current to 400 mA nominally at 25°C. The current limit value is inversely proportional to temperature; therefore, the current limit value increases at low temperatures.

When current is limited, the safe limits for the die temperature must be taken in to account. With too much power dissipation, the die temperature can surpass thermal shutdown limits; the op amp shuts down and reactivates after the die has fallen below thermal limits.

#### 注意

Do not continuously operate the device in thermal hysteresis for long periods of time because this action may cause irreversible damage to the device.

### 7.3.4 Input Common-Mode Range

The input common-mode range of the ALM2403-Q1 is between  $(V_-) - 0.2\text{ V}$  and  $(V_+) + 0.2\text{ V}$ . Staying within this range allows the op amps to perform and operate within specification. Operating beyond these limits can cause distortion and nonlinearities.

### 7.3.5 Reverse Body Diodes in Output-Stage Transistors

Designed as a high-voltage, high current operational amplifier, the ALM2403-Q1 delivers robust output drive capability. A class AB output stage with common-source transistors is used to achieve full rail-to-rail output swing capability. Different load conditions change the ability of the amplifier to swing close to the rails.

Each output transistor has internal reverse diodes between drain and source that conduct if the output is forced to greater than the supply or less than ground (reverse current flow). These diodes can be used as flyback protection in inductive-load-driving applications. Limit the use of these diodes to pulsed operation in order to minimize junction temperature overheating due to  $(V_F \times I_F)$ . Internal current-limiting circuitry does not operate when current is flown in the reverse direction and the reverse diodes are active. A method to protect these reverse body diodes is shown in [セクション 8.2.2.1.2](#).

### 7.3.6 EMI Filtering

Op amps vary with regard to the susceptibility of the device to electromagnetic interference (EMI). If conducted EMI enters the op amp, the dc offset observed at the amplifier output may shift from the nominal value while EMI is present. This shift is a result of signal rectification associated with the internal semiconductor junctions. While all op-amp pin functions can be affected by EMI, the signal input pins are likely to be the most susceptible. The ALM2403-Q1 incorporates an internal input low-pass filter that reduces the amplifiers response to EMI. Both common-mode and differential mode filtering are provided by this filter.

Texas Instruments has developed the ability to accurately measure and quantify the immunity of an operational amplifier over a broad frequency spectrum extending from 10 MHz to 990 MHz. The EMI rejection ratio (EMIRR) metric allows op amps to be directly compared by the EMI immunity. Detailed information can also be found in the [EMI Rejection Ratio of Operational Amplifiers application report](#), available for download from [www.ti.com](http://www.ti.com).

## 7.4 Device Functional Modes

### 7.4.1 Open-Loop and Closed-Loop Operation

As a result of the very-high, open-loop dc gain of the ALM2403-Q1, the device functions as a comparator in open loop for most applications. A majority of electrical characteristics are verified in negative feedback, closed-loop configurations. Certain dc electrical characteristics, like offset, may have a higher drift across temperature and lifetime when continuously operated in open loop over the lifetime of the device.

### 7.4.2 Shutdown

When the OTF/SH\_DN pin is left floating or is grounded, the op amp shuts down to a low  $I_Q$  state and does not operate; the op amp outputs go to a high-impedance state.

**表 7-1. Shutdown Truth Table**

PIN NAME	LOGIC STATE	OP AMP STATE
OTF/SH_DN	High ( $> V_{IH\_OTF}$ )	Operating
	Low ( $< V_{IL\_OTF}$ )	Shutdown (low $I_Q$ state)

## 8 Application and Implementation

### 注

以下のアプリケーション情報は、TI の製品仕様に含まれるものではなく、TI ではその正確性または完全性を保証いたしません。個々の目的に対する製品の適合性については、お客様の責任で判断していただくこととなります。お客様は自身の設計実装を検証しテストすることで、システムの機能を確認する必要があります。

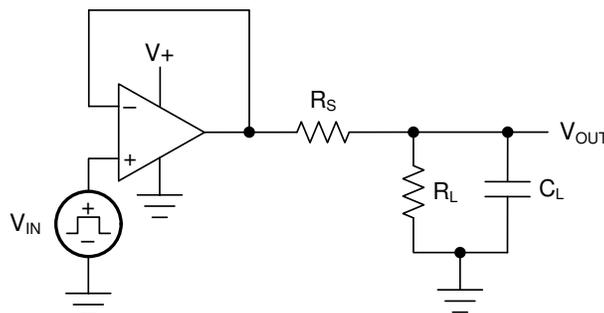
### 8.1 Application Information

The ALM2403-Q1 is a dual-power op amp with performance and protection features that are optimal for many applications. For op amps, there are many general design consideration that must be taken into account. The following subsections describe what to consider for most closed-loop applications. [セクション 8.2](#) gives a specific example of the ALM2403-Q1 being used in a resolver application.

#### 8.1.1 Capacitive Load and Stability

The ALM2403-Q1 is designed for applications where driving a capacitive load is required. As with all op amps, specific instances can occur where the ALM2403-Q1 device can become unstable. The particular op-amp circuit configuration, layout, gain, and output loading are some of the factors to consider when establishing whether or not an amplifier is stable in operation. An op amp in a unity-gain (1-V/V) buffer configuration that drives a capacitive load exhibits a greater tendency to become unstable compared to an amplifier operated at a higher-noise gain. The capacitive load, in conjunction with the op-amp output resistance, creates a pole within the feedback loop that degrades the phase margin. The degradation of the phase margin increases as the capacitive loading increases. When operating in a unity-gain configuration, the ALM2403-Q1 remains stable with a pure capacitive load up to approximately 30 pF. Increasing the amplifier closed-loop gain allows the amplifier to drive increasingly larger capacitance. This increased capability is evident when observing the overshoot response of the amplifier at higher voltage gains.

One technique for increasing the capacitive load drive capability of the amplifier operating in a unity-gain configuration is to insert a small resistor ( $R_S$ ; typically, 100 m $\Omega$  to 10  $\Omega$ ) in series with the output, as shown in [図 8-1](#). This resistor significantly reduces the overshoot and ringing associated with large capacitive loads.



**図 8-1. Capacitive Load Drive**

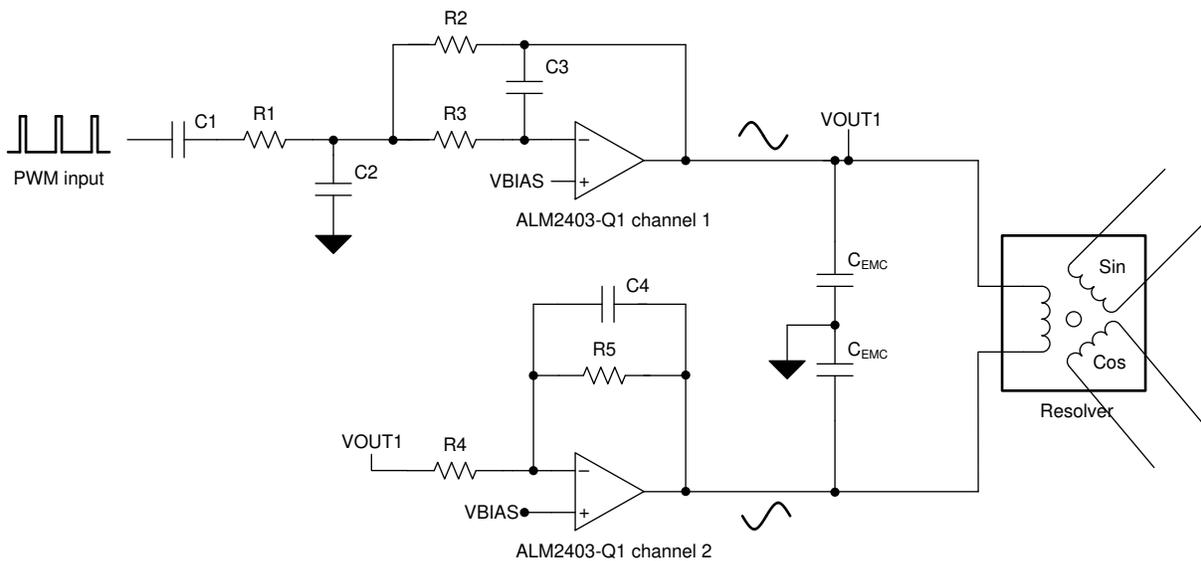
### 8.2 Typical Application

High-power ac and brushless dc (BLDC) motor-drive applications need position feedback to efficiently and accurately drive the motor. Position feedback can be achieved by using optical encoders, hall sensors, or resolvers. Resolvers are the main choice when environmental or longevity requirements are challenging and extensive.

A resolver acts as a transformer with one primary coil and two secondary coils. The primary coil, or excitation coil, is located on the rotor of the resolver. As the rotor of the resolver spins, the excitation coil induces a current into the sine and cosine sensing coils. These coils are oriented 90 degrees from one another, and the voltage from the sine and cosine coils is translated into a vector position by the microcontroller or resolver-to-digital converter chip.

Resolver excitation coils can have a very low dc resistance ( $< 100 \Omega$ ), requiring a sink and a source of up to 200 mA from the excitation driver. The ALM2403-Q1 can source and sink this current while providing current-limiting and thermal-shutdown protection. Incorporating these protections in a resolver design can increase the life of the end product.

The input to the ALM2403-Q1 can be an analog sine wave generated by the resolver-to-digital converter chip or a pulse-width modulation (PWM) signal generated from a microcontroller I/O pin. In the case of the latter, a filter stage is needed to extract a lower bandwidth sine wave from the PWM signal. This sine wave would then be the input signal to the ALM2403-Q1. As a result of high gain bandwidth, the ALM2403-Q1 can be configured as a filter stage while providing the required output drive. This configuration significantly reduces the total solution size and design complexity of the resolver-drive signal chain. The fundamental design steps to achieve this functionality are shown in this application example, and can be applied to other inductive-load applications as well.



**图 8-2. Resolver-Based Application**

### 8.2.1 Design Requirements

For this design example, use the parameters listed in [表 8-1](#) as the input parameters.

**表 8-1. Design Parameters**

DESIGN PARAMETER	EXAMPLE VALUE
Ambient temperature range	$-40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$
Available supply voltages	15 V
EMC capacitance (CL)	10 nF
Resolver excitation input voltage	$7 V_{\text{RMS}}$
Excitation frequency	10 kHz
PWM signal frequency	320 kHz
PWM signal amplitude	3.3 V
Functional safety capable	Yes
Short-to-battery protection	Yes

## 8.2.2 Detailed Design Procedure

When using the ALM2403-Q1 in a resolver application, determine:

- Resolver excitation input impedance or resistance and inductance:  $Z_O = 100 + j188$ ,  $R = 100 \Omega$ , and  $L = 3 \text{ mH}$  at 10 kHz
- Resolver transformation ratio ( $V_{\text{SINCOS}} / V_{\text{EXC}}$ ): 0.5 V/V at 10 kHz
- Package and  $R_{\theta\text{JA}}$ : HTSSOP, 46.9°C/W
- Op amp maximum junction temperature: 150°C
- Op amp bandwidth: 21 MHz
- Op amp slew rate: 50 V/ $\mu\text{S}$

### 8.2.2.1 Resolver Excitation Amplifier Combined With MFB 2nd-Order, Low-Pass Filter

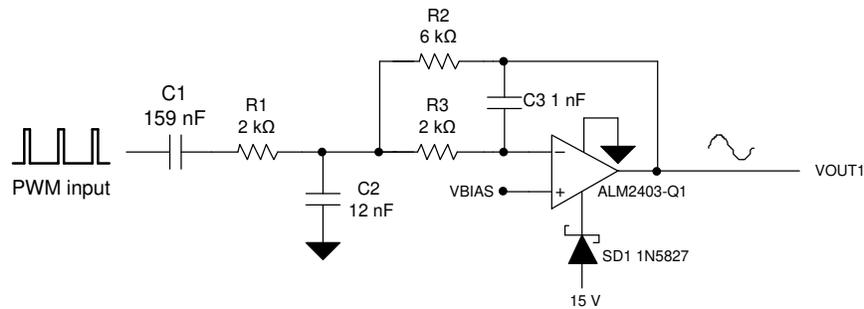


FIG 8-3. Two-Pole MFB Filter

When designing a low-pass filter, the most important design criteria is to decide the corner frequency. In this design example, the resolver excitation frequency is 10 kHz and PWM frequency is 320 kHz. Thus, we want to make sure that the low-pass filter corner frequency is greater than 10 kHz, and there is maximum attenuation of harmonic interference generated from the PWM signal. FIG 8-3 shows a single channel of the ALM2403-Q1 configured as a 2-pole multiple feedback (MFB) filter with a  $-40 \text{ dB/decade}$  rolloff. The MFB topology enables a steep rolloff while reducing BOM count. The output from this circuit is a sine wave that can then be inverted using the second channel of the ALM2403-Q1; see FIG 8-2. Thus, both ALM2403-Q1 channels combined provide the required resolver excitation signal.

#### 8.2.2.1.1 Filter Design

The corner frequency of the 2nd-order MFB filter is set to approximately twenty times less than the PWM frequency. The corner frequency defined at  $-3 \text{ dB}$  is shown in 式 1.

$$f_p = \frac{1}{2 \times \pi \times \sqrt{R_3 \times C_3 \times R_2 \times C_2}} \quad (1)$$

The 2nd-order MFB active filter uses an inverted input topology and the op amp gain is determined by the ratios of resistors R2 and R1:

$$\text{Gain} = -\frac{R_2}{R_1} \quad (2)$$

The gain settings are based on the output drive requirements and PWM signal amplitude. With different gain settings, the filter characteristics, such as rolloff, can change. The design must be fine-tuned to meet optimal performance needs.

The quality (Q) factor of the low-pass filter is configured with  $Q = 1$ . The purpose of designing for this Q factor is to minimize attenuation around the corner frequency of 10 kHz, thus extending the pass-band gain. The Q factor of the 2nd-order MFB filter is given by 式 3:

$$Q = \frac{\sqrt{C_2 / C_3}}{\sqrt{R_3 / R_2} + \sqrt{R_2 / R_3} + \sqrt{R_3 \times R_2 / R_1}} \quad (3)$$

### 8.2.2.1.2 Short-to-Battery Protection

Resolver-based applications require the power op amp stage to provide the resolver excitation signal over long cables. In many applications, such as automotive traction inverters, the cables are housed in a harness and a short-circuit condition between different cables in the same harness can occur. In this situation, the output of the ALM2403-Q1 can see a higher voltage than provided at the positive supply pin. This condition causes the body diode in the output stage PMOS to become forward-biased and start conducting. As a precaution, use a blocking diode in series with the positive power supply; see also [Figure 8-3](#).

For related information, see the [ALM2403-Q1 Overvoltage Protection of Resolver-Based Circuits application note](#).

### 8.2.2.2 Power Dissipation and Thermal Reliability

Power dissipation is critical to many industrial and automotive applications. Resolvers are typically chosen over other position feedback techniques because of reliability and accuracy in harsh conditions and high temperatures.

The ALM2403-Q1 is capable of high output current with power-supply voltages up to 24 V. Internal power dissipation increases when operating at high supply voltages. The power dissipated in the op amp ( $P_{OPA}$ ) is calculated using [Equation 4](#):

$$P_{OPA} = (V_S - V_{OUT}) \times I_{OUT} = (V_S - V_{OUT}) \times \frac{V_{OUT}}{R_L} \quad (4)$$

To calculate the worst-case power dissipation in the op amp, the ac and dc cases must be considered separately.

In the case of constant output current (dc) to a resistive load, the maximum power dissipation in the op amp occurs when the output voltage is half the positive supply voltage. This calculation assumes that the op amp is sourcing current from the positive supply to a grounded load. If the op amp sinks current from a grounded load, modify [Equation 5](#) to include the negative supply voltage instead of the positive.

$$P_{OPA(MAX\_DC)} = P_{OPA}\left(\frac{V_S}{2}\right) = \frac{(V_S)^2}{4 \times R_L} \quad (5)$$

The ac maximum of average power dissipation in the op amp for a sinusoidal output current (ac) to a resistive load occurs when the peak output voltage is  $2/\pi$  times the supply voltage, given symmetrical supply voltages, as shown in [Equation 6](#):

$$P_{OPA(PEAK\_AC)} = P_{OPA}\left(\frac{2 \times V_S}{\pi}\right) = \frac{2 \times (V_S)^2}{\pi^2 \times R_L} \quad (6)$$

After the total power dissipation is determined, the junction temperature at the worst expected ambient temperature case must be determined by using [Equation 7](#):

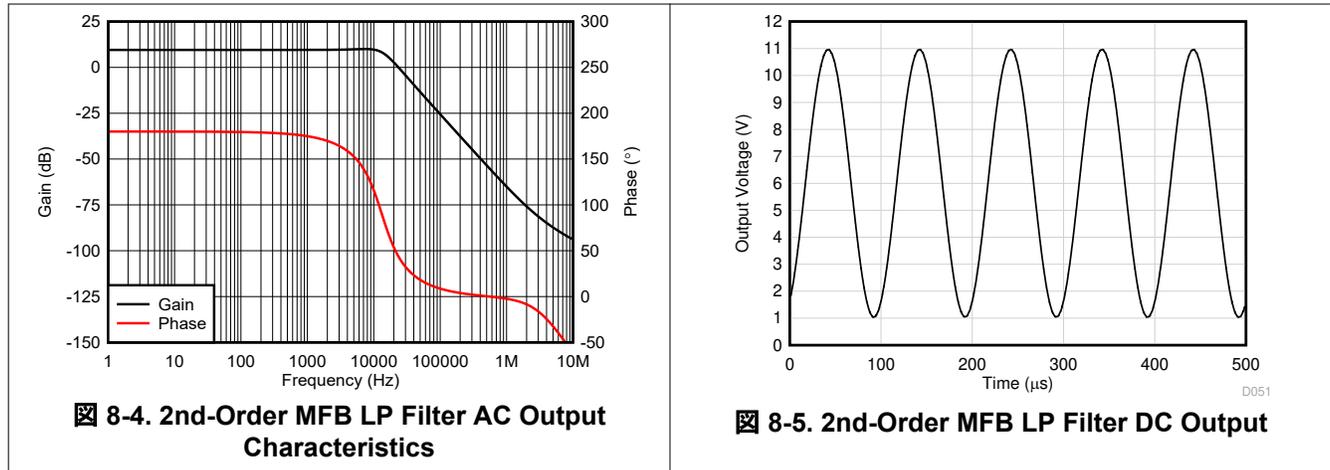
$$T_{J(MAX)} = P_{OPA} \times R_{\theta JA} + T_{A(MAX)} \quad (7)$$

#### 8.2.2.2.1 Improving Package Thermal Performance

The value of  $R_{\theta JA}$  depends on the printed circuit board (PCB) layout. An external heat sink, a cooling mechanism such as a cold air fan, or both, can help reduce  $R_{\theta JA}$ , and thus improve device thermal capabilities. See TI's design support web page at [www.ti.com/thermal](http://www.ti.com/thermal) for general guidance on improving device thermal performance.

### 8.2.3 Application Curves

The roll of characteristics and output waveform for the designed MFB filter are shown in [図 8-4](#) and [図 8-5](#). The attenuation is specified in [表 8-2](#).



**表 8-2. Signal Attenuation vs Frequency**

2ND-ORDER MFB LPF FREQUENCY (kHz)	ATTENUATION (dB)
DC	9.54
10.0	9.70
15.4	6.54
19	3.54
30	-4.38
320	-45.9

### 8.3 Power Supply Recommendations

The ALM2403-Q1 is recommended for continuous operation from 5 V to 24 V ( $\pm 2.5$  V to  $\pm 12$  V) for  $V_S$ , and many specifications apply from  $-40^\circ\text{C}$  to  $+125^\circ\text{C}$ .

Place 0.1- $\mu\text{F}$  bypass capacitors close to the power-supply pins to reduce errors coupling from noisy or high-impedance power supplies.

**注意**

Supply voltages larger than 26 V can permanently damage the device (see [セクション 6.1](#)).

### 8.4 Layout

#### 8.4.1 Layout Guidelines

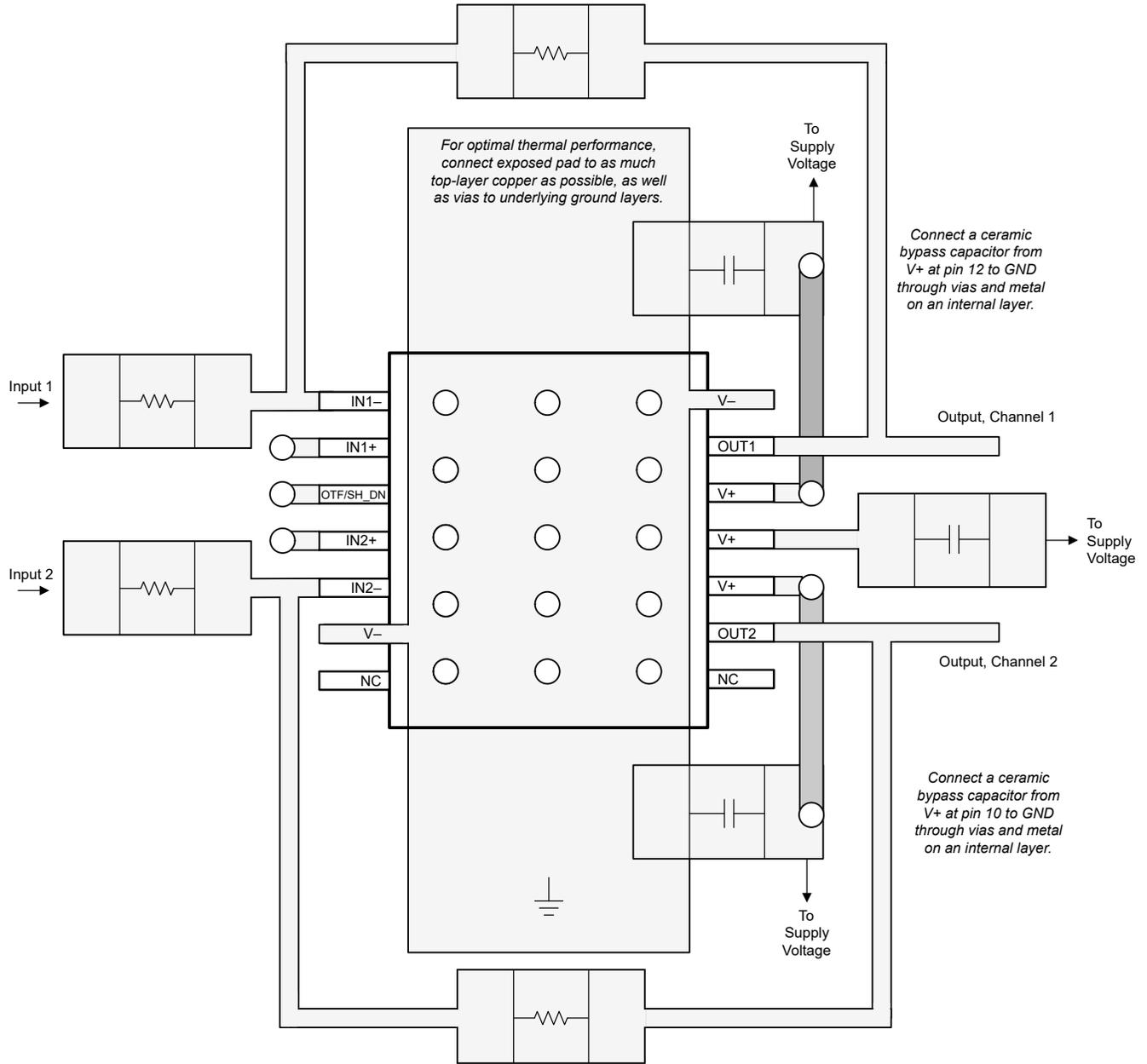
For best operational performance of the device, use good PCB layout practices, including:

- Noise can propagate into analog circuitry through the power pins of the circuit as a whole, as well as the operational amplifier. Bypass capacitors are used to reduce the coupled noise by providing low impedance power sources local to the analog circuitry.
  - Connect low-ESR, 0.1- $\mu\text{F}$  ceramic bypass capacitors between each supply pin and ground, placed as close as possible to the device. A single bypass capacitor from  $V_+$  to ground is applicable for single-supply applications.

- Separate grounding for analog and digital portions of circuitry is one of the simplest and most-effective methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground planes. A ground plane helps distribute heat and reduces EMI noise pickup. Make sure to physically separate digital and analog grounds, paying attention to the flow of the ground current. For more detailed information, see [Circuit Board Layout Techniques](#).
- To reduce parasitic coupling, run the input traces as far away as possible from the supply or output traces. If keeping the traces separate is not possible, then cross the sensitive trace perpendicular, as opposed to in parallel with the noisy trace.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.

#### 8.4.2 Layout Example

This layout does not verify optimum thermal impedance performance. See TI's design support web page at [www.ti.com/thermal](http://www.ti.com/thermal) for general guidance on improving device thermal performance.



**8-6. ALM2403-Q1 Layout Example**

## 9 Device and Documentation Support

### 9.1 Documentation Support

#### 9.1.1 Related Documentation

For related documentation see the following: [ALM2403-Q1 Evaluation Module user's guide](#).

### 9.2 ドキュメントの更新通知を受け取る方法

ドキュメントの更新についての通知を受け取るには、[ti.com](#) のデバイス製品フォルダを開いてください。「更新の通知を受け取る」をクリックして登録すると、変更されたすべての製品情報に関するダイジェストを毎週受け取れます。変更の詳細については、修正されたドキュメントに含まれている改訂履歴をご覧ください。

### 9.3 サポート・リソース

[TI E2E™ サポート・フォーラム](#)は、エンジニアが検証済みの回答と設計に関するヒントをエキスパートから迅速かつ直接得ることができる場所です。既存の回答を検索したり、独自の質問をしたりすることで、設計に必要な支援を迅速に得ることができます。

リンクされているコンテンツは、該当する貢献者により、現状のまま提供されるものです。これらは TI の仕様を構成するものではなく、必ずしも TI の見解を反映したものではありません。TI の[使用条件](#)を参照してください。

### 9.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.

すべての商標は、それぞれの所有者に帰属します。

### 9.5 静電気放電に関する注意事項



この IC は、ESD によって破損する可能性があります。テキサス・インスツルメンツは、IC を取り扱う際には常に適切な注意を払うことを推奨します。正しい取り扱いおよび設置手順に従わない場合、デバイスを破損するおそれがあります。

ESD による破損は、わずかな性能低下からデバイスの完全な故障まで多岐にわたります。精密な IC の場合、パラメータがわずかに変化するだけで公表されている仕様から外れる可能性があるため、破損が発生しやすくなっています。

### 9.6 用語集

[テキサス・インスツルメンツ用語集](#) この用語集には、用語や略語の一覧および定義が記載されています。

## 10 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ALM2403QPWPRQ1	ACTIVE	HTSSOP	PWP	14	2000	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 125	A2403Q	Samples

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

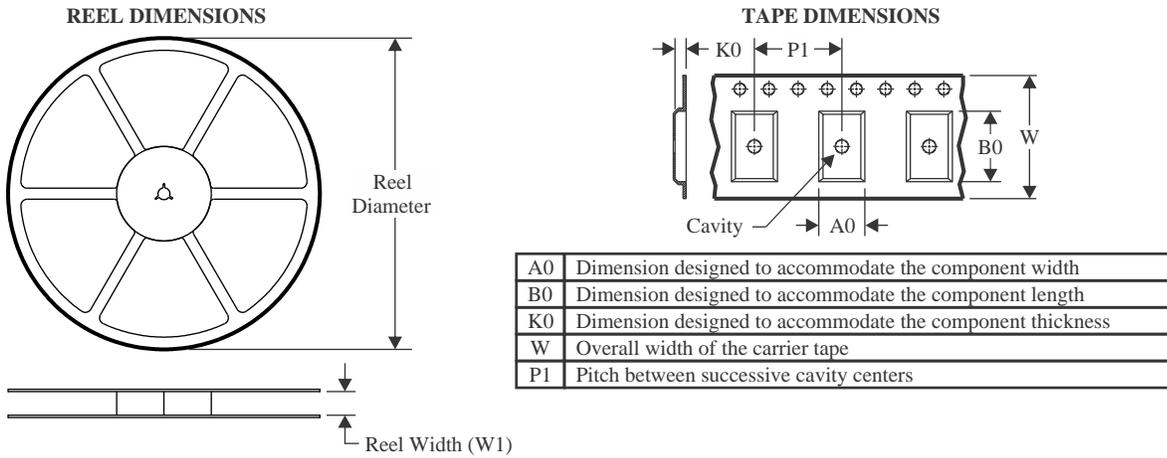
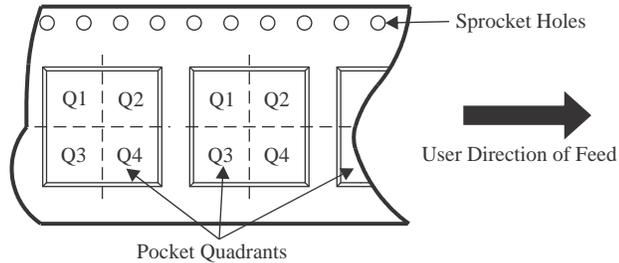
(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
ALM2403QPWRQ1	HTSSOP	PWP	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
ALM2403QPWPRQ1	HTSSOP	PWP	14	2000	356.0	356.0	35.0

## GENERIC PACKAGE VIEW

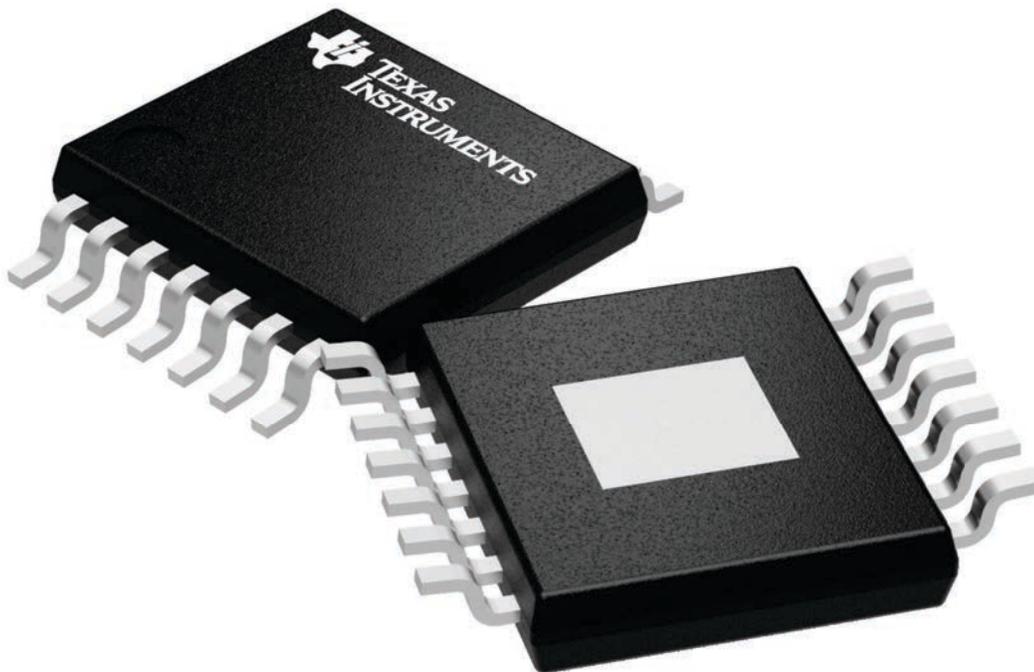
**PWP 14**

**PowerPAD TSSOP - 1.2 mm max height**

4.4 x 5.0, 0.65 mm pitch

PLASTIC SMALL OUTLINE

This image is a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.

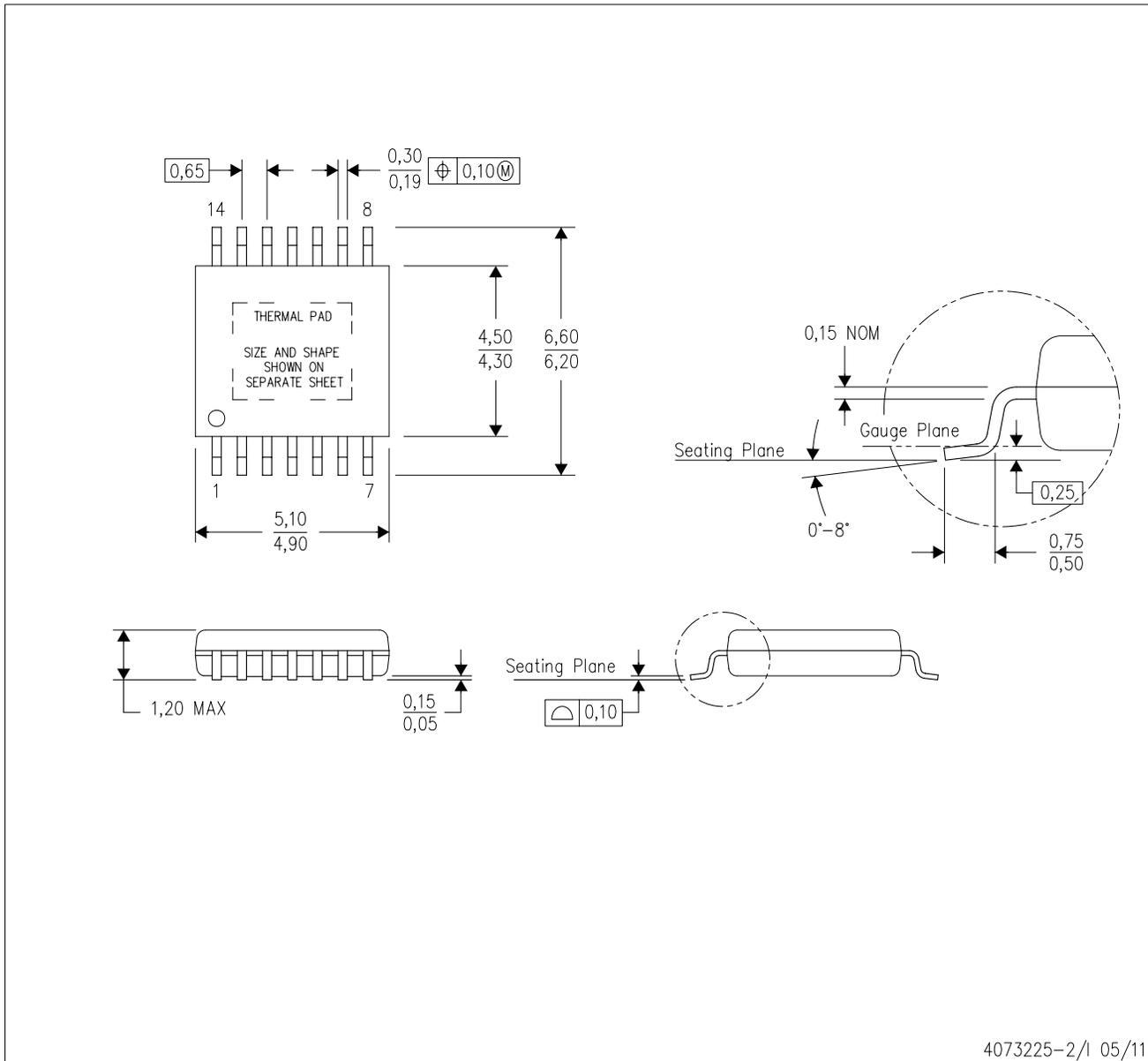


4224995/A

# MECHANICAL DATA

PWP (R-PDSO-G14)

PowerPAD™ PLASTIC SMALL OUTLINE



- NOTES:
- All linear dimensions are in millimeters.
  - This drawing is subject to change without notice.
  - Body dimensions do not include mold flash or protrusions. Mold flash and protrusion shall not exceed 0.15 per side.
  - This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at [www.ti.com](http://www.ti.com) <<http://www.ti.com>>.
  - See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
  - Falls within JEDEC MO-153

PowerPAD is a trademark of Texas Instruments.

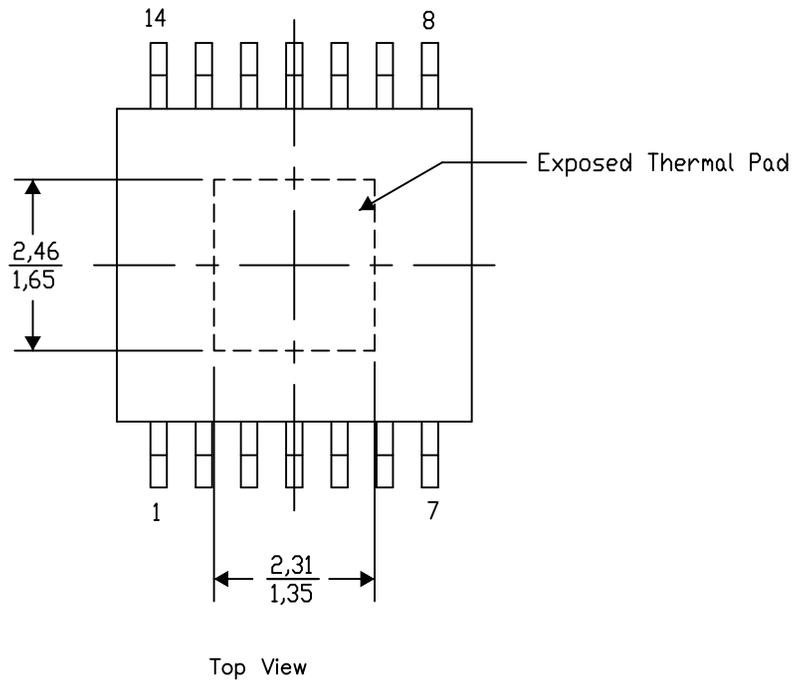
**PWP (R-PDSO-G14) PowerPAD™ SMALL PLASTIC OUTLINE**

**THERMAL INFORMATION**

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at [www.ti.com](http://www.ti.com).

The exposed thermal pad dimensions for this package are shown in the following illustration.



Exposed Thermal Pad Dimensions

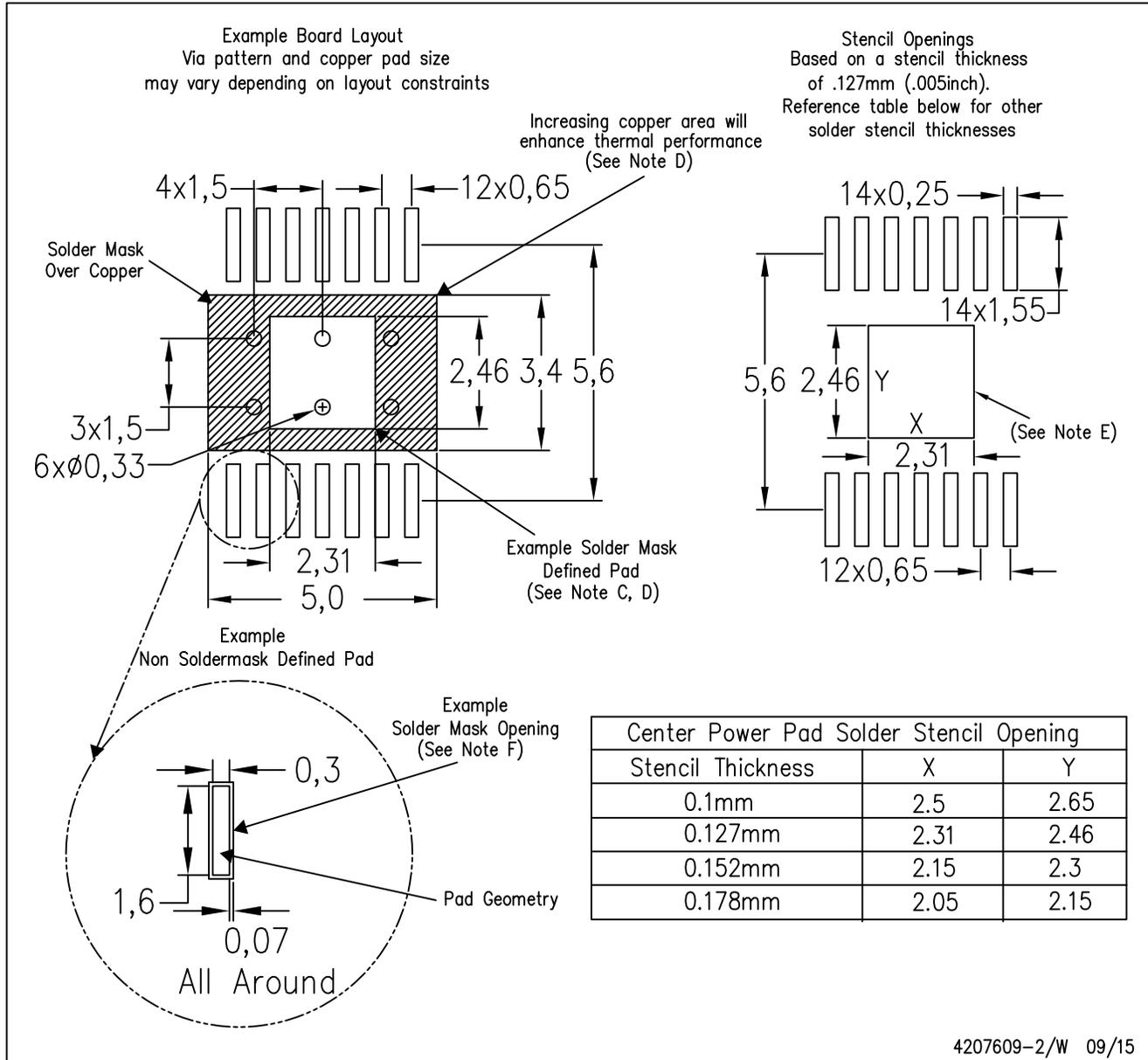
4206332-2/AO 01/16

NOTE: A. All linear dimensions are in millimeters

PowerPAD is a trademark of Texas Instruments

PWP (R-PDSO-G14)

PowerPAD™ PLASTIC SMALL OUTLINE



4207609-2/W 09/15

- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
  - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at [www.ti.com](http://www.ti.com) <<http://www.ti.com>>. Publication IPC-7351 is recommended for alternate designs.
  - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
  - F. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

## 重要なお知らせと免責事項

TI は、技術データと信頼性データ (データシートを含みます)、設計リソース (リファレンス・デザインを含みます)、アプリケーションや設計に関する各種アドバイス、Web ツール、安全性情報、その他のリソースを、欠陥が存在する可能性のある「現状のまま」提供しており、商品性および特定目的に対する適合性の黙示保証、第三者の知的財産権の非侵害保証を含むいかなる保証も、明示的または黙示的にかかわらず拒否します。

これらのリソースは、TI 製品を使用する設計の経験を積んだ開発者への提供を意図したものです。(1) お客様のアプリケーションに適した TI 製品の選定、(2) お客様のアプリケーションの設計、検証、試験、(3) お客様のアプリケーションに該当する各種規格や、その他のあらゆる安全性、セキュリティ、規制、または他の要件への確実な適合に関する責任を、お客様のみが単独で負うものとし、

上記の各種リソースは、予告なく変更される可能性があります。これらのリソースは、リソースで説明されている TI 製品を使用するアプリケーションの開発の目的でのみ、TI はその使用をお客様に許諾します。これらのリソースに関して、他の目的で複製することや掲載することは禁止されています。TI や第三者の知的財産権のライセンスが付与されている訳ではありません。お客様は、これらのリソースを自身で使用した結果発生するあらゆる申し立て、損害、費用、損失、責任について、TI およびその代理人を完全に補償するものとし、TI は一切の責任を拒否します。

TI の製品は、[TI の販売条件](#)、または [ti.com](https://www.ti.com) やかかる TI 製品の関連資料などのいずれかを通じて提供する適用可能な条項の下で提供されています。TI がこれらのリソースを提供することは、適用される TI の保証または他の保証の放棄の拡大や変更を意味するものではありません。

お客様がいかなる追加条項または代替条項を提案した場合でも、TI はそれらに異議を唱え、拒否します。

郵送先住所 : Texas Instruments, Post Office Box 655303, Dallas, Texas 75265  
Copyright © 2023, Texas Instruments Incorporated