

# 3 to 6 Cells Battery-Management System Based On bq76925 + MSP430G2xx2

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### ABSTRACT

This application report describes how to use bq76925 and MSP430G2xx2 to implement a high-accuracy digital battery-management solution, which can support a complete pack monitoring, balancing, protection, and gas gauging system for 3 to 6 series cell Lithium-Ion / Polymer battery. This solution is designed to focus on power tool projects, while it can be also a reference design for other similar applications.

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## 1 Features

To implement a high-accuracy digital battery-management solution, bq76925 + MSP430G2xx2 can support a complete pack monitoring, balancing, protection, and gas gauging system for 3 to 6 series cell Lithium-Ion / Polymer battery. This solution is designed to focus on power tool projects, while it can also be a reference design for other similar applications.

## 2 Introduction

The bq76925 is a dedicated analog front end (AFE) for 3 to 6 series cell application that provides 3 analog outputs that allow a microcontroller to easily monitor cell voltage, current, and temperature. Cell voltages are level-shifted, scaled, and multiplexed to the VCOUT pin. Cell current is monitored through a sense resistor placed in series with the cell stack. The voltage across the sense resistor is amplified and driven to the VIOUT pin. The VTB pin supplies a switched bias to stimulate a thermistor network for temperature measurement. The bq76925 supplies a 3.3-V regulated output to power the MSP430G2xx2 and an accurate 3.3-V reference voltage for the MSP430G2xx2 analog-to-digital converter (ADC). The AFE also includes integrated cell balancing FETs that are under control of the MSP430G2xx2 for fast fault response.

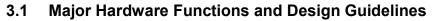
The MSP430G2xx2 series is an ultra-low-power microcontroller with one built-in 16-bit timer, a fast 10-bit ADC with integrated reference, built-in communication capability using the universal serial communication interface, and up to 24 I/O pins. The architecture, combined with five low-power modes, is optimized to achieve extended battery life in portable measurement applications. The MSP430G2xx2 application circuit is provided to read the analog signals from the bq76925 and let the user create custom firmware based on the application code.

With the AFE of bq76925, the MSP430G2xx2 is evaluated as the host to control bq76925 by I<sup>2</sup>C of USI module, the MSP430G2xx2 can read the data of analog signals from bq76925 by ADC10. Then the MSP430G2xx2 can easily monitor individual cell voltages, pack current, and temperature to achieve battery charge and discharge management. Although it is only a 10-bit ADC result for the MSP430G2xx2, bq76925 can calibrate it into a high-accuracy value with the particular compensated algorithmic of correction factors. For example, the ADC10 value deviation of cell voltage does not exceed 5 mV, which can strongly make this solution outstanding in this field requirement. To make this solution more competitive, the MSP430G2xx2 is selected for its advantages of low power and low cost.

The user can choose the part number of the MSP430G2xx2 from MSP430G2x32/G2x52, as there are eight part numbers for the different configurations, like flash size, Comparator\_A+. More software functions mean more flash size is need. The user can choose the right part number based on the necessary functions of the MSP430G2xx2.

## 3 Description

This application note, which can help designers evaluate the bq76925 + MSP430G2xx2 batterymanagement system, discusses the important setup and operation of the module, and contains schematics, bill of materials, and printed circuit board layout. Before designing a batterymanagement system based on the bq76925 + MSP430G2xx2 solution, designers are advised to read the bq76925 and MSP430G2xx2 data sheet.



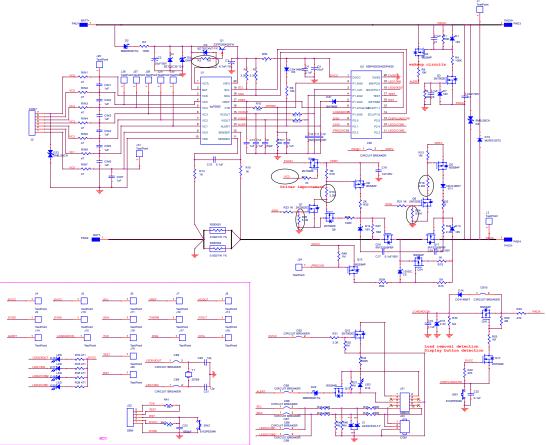


Figure 1. Overall Schematic

## 3.1.1 Cell Voltage Sense Circuits

At each cell voltage input port, an RC filter circuit is necessary (as are RIN and CIN in the following circuits), which can filter the input noise to improve the accuracy of cell voltage sense. Considering the internal cell balancing current, input resistor RIN cannot be selected too largely. For example, selected  $47-\Omega$  RIN can provide balancing current of about 40 to 50 mA. Moreover, because the pack current will be up to hundreds of amperes when SC occurs, a very large peak voltage exists when SC protection occurs, which may exceed the max input voltage range of the device. One TVS diode D23 can provide good protection for this situation.

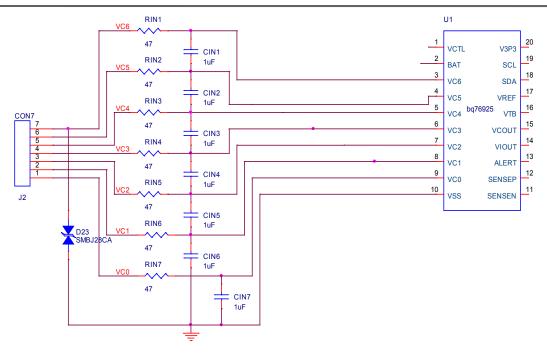


Figure 2. Cell Voltage Sense Circuits

## 3.1.2 External LDO

bq76925 integrates one internal LDO but the drive capability is only 3 mA. To provide enough power to the MCU and external controlling circuits, external LDO is necessary. bq76925 can support the external LDO extension. The VCTL pin can control an external NPN transistor or MOSFET to generate the 3.3-V power supply to bq76925 and other external circuits.

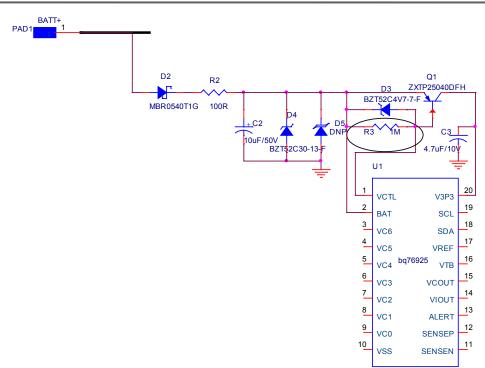


Figure 3. External LDO

## 3.1.3 SC Protection and Wake-up Circuits

bq76925 integrates a hardware comparator for current sense, which can be used for SC protection. The output of the comparator is the ALERT pin that can be connected to the external interrupt input pin of the MSP430G2xx2. When an SC event occurs, the protection activity can execute most quickly. The ALERT pin is an open-drain structure, so one pullup is necessary just as R8 and D6. Moreover, the ALERT pin can be used to wake up the bq76925 from sleep mode. As shown in Figure 4, R1, R4, R6, R9, Q2, and Q3 form the wake-up circuits that can wake up bq76925 when the charger is plugged in. D6 blocks the high voltage to 3.3-V power supply. The special attention that should be paid is that D21 is mandatory. The ESD structure of the MSP430G2xx2 I/O pin inducts a reverse diode to the VDD pin, which clamps the voltage level of the ALERT pin so that the bq76925 cannot be woken up successfully. Here, D21 is used merely to remove the clamp effect of the internal diode.

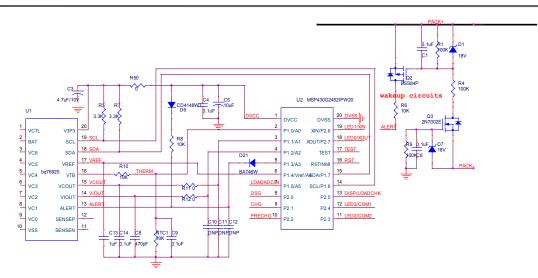


Figure 4. SC Protection and Wake-up Circuits

### 3.1.4 Voltage and Current Measurement

bq76925 can accurately transfer the cell voltage and current of the AFE to a single-end analog signal. This signal, in turn, can be output to the input channels of the ADC of the MSP430G2xx2. Thereby, the MSP430G2xx2 can accurately monitor the voltage, current, and temperature. Moreover, bq76925 provides a very accurate reference voltage that can be used for the ADC reference. Here, C8 and C14 are used to filter some noise. See Figure 5.

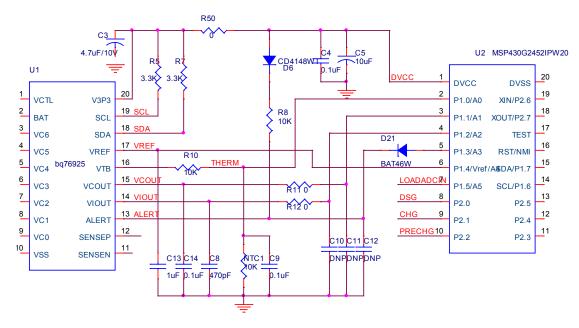


Figure 5. Voltage and Current Measurement



### 3.1.5 Drivers

The drivers must solve two problems. One problem is that the driver power must provide enough high voltage and driving capability; the other problem is to ensure that the FETs can be turned on and off quickly enough. Because the controlling signals come from the MSP430G2xx2, the voltage level transfer circuits are necessary. Certainly, the user must also carefully consider total power consumption.

R7, C16, and Q4 generate the driver power supply that avoids cell imbalance by getting the power from the top side of the pack, and also gets the suitable driver power level with the simplest structure. C16 is the buffer capacitor that can provide the dynamic current when the FETs are switched.

Q6, Q7, Q9, R16, R19, R20, and R26 form the discharging controlling circuits. When DSG is high, Q6 and Q7 are turned off and Q9 is turned on to turn off the discharging FET quickly. Special attention must be paid when selecting these resistors. R16 and R19 should not be too small; otherwise, the power consumption will be large. However, it is better to select bigger proportion for R20 and R26. Because the turn-off speed of Q7 is somewhat slow, both Q7 and Q9 are turned on for a short period. The low voltage on R26 must be able to turn off the discharging FET quickly and reliably. If necessary, one 0- $\Omega$  resistor can be used. But R20 must ensure enough driving capability.

Q5, Q8, R13, R18, R22, R28, and D11 form the charging controlling circuits. When CHG is high, the driver turned on the charging FET through R22. When CHG is low, Q5 is turned off then the charges on the gate of the charging FET are discharging through R28 to turn off the charging FET. Here, R28 cannot be too large; otherwise, the turn-off time will be too long. D11 is used to block the reverse high voltage when discharging FET is turned off.

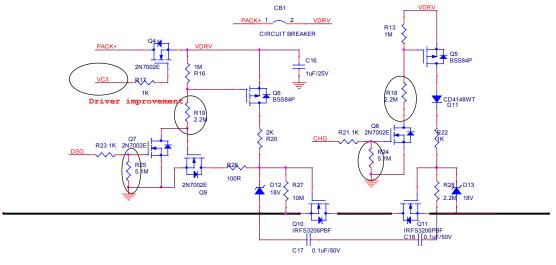
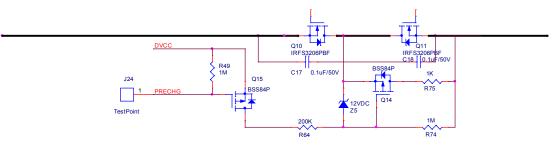


Figure 6. Drivers

## 3.1.6 Precharge Control

Q14, Q15, Z5, R49, R64, R74, and R75 form precharge controlling circuits that can support 0-V charging. When the pack voltage is 0 V, the chip does not work and DVCC is 0 V. When a charger is plugged in at the moment, Q14 is turned on and the pack is charged through R75. If the pack voltage arises up to a certain level, the chip is powered on and DVCC is 3.3 V. The MSP430G2xx2 will work and output high PRECHG signal. The pack continues to be charged through R75. When the pack voltage rises to exceed the threshold of rapid charging, PRECHG outputs low, the precharge path is turned off, and the main charging path is turned on.





## 3.1.7 Load Removal Detection

C19, D15, R29, Q16 (optional), and Q17 (optional) form the detection circuits for load removal. When an SCD or an OCD event occurs, the discharging FET is turned off. If the load is still present, the voltage of PACK– is close to that of PACK+. But if the load is removed, the voltage of PACK– drops to 0 V. By checking the voltage of PACK–, the load removal can be detected to release the protection. When the load is present and protection works, the discharging path for the PACK– causes somewhat larger power consumption; Q16 and Q17 can be used to control the detection path. If the detection is finished, the discharging path can be cut off to reduce the power consumption. If Q16 and Q17 are not used, the circuits are simpler but R29 must be larger to limit the discharging current. Moreover, the relevant detection interval must be selected accordingly.

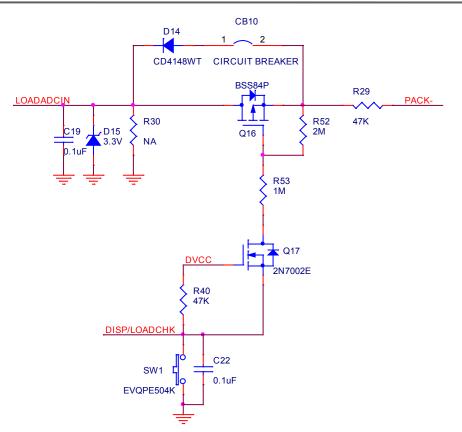


Figure 8. Load Removal Detection

## 3.2 MCU Software Description

The MSP430G2xx2 can monitor individual cell voltages, pack current, battery alert, and temperature, in order to achieve battery charge and discharge management. Basically, the user can set user.h to adjust these parameters of BMS. See Appendix A for more detail on user.h. User can also refer to the following sections for software flow.

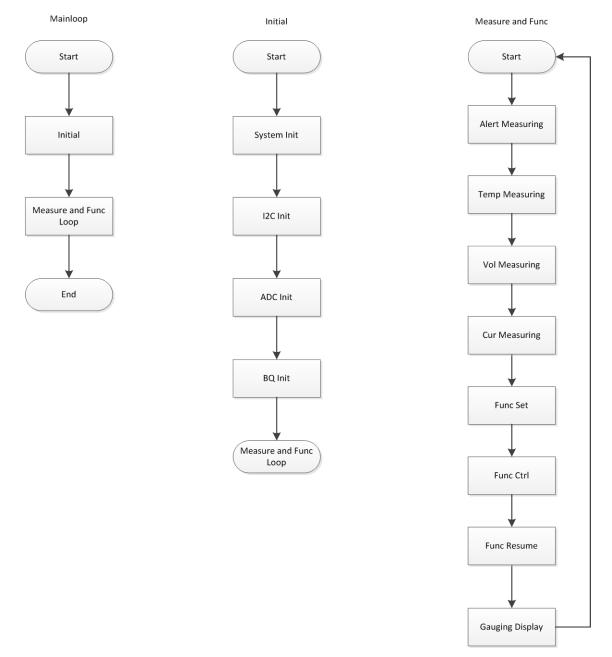
## 3.2.1 MSP430G2xx2 Software Flow

#### 3.2.1.1 Mainloop and Initial

After the Initial loop, the MSP430 keeps running the Measure and Func loop. In the Initial loop, BQ Init is to get the correct factor, which is used for the ADC calibrated in Measuring. There are eight sections for the Measure and Func loop: Alert Measuring, Temp Measuring, Vol Measuring, Cur Measuring, Func Set, Func Ctrl, Func Resume, and Gauging Display. These sections can be defined in two main sections (Measuring and Func); in other words, the MSP430 must monitor the status of cells by Measuring, and then do the charge and discharge management by Func.

As for the interrupt-enables in this system, TimerA0 INT is used for the counter to meet the required function delay time, such as LPM keep time, Balance keep time, OC keep time, and so on. Port1 INT is for Alert monitor. ADC10 INT is for ADC sample conversion trigger. USI INT is for I<sup>2</sup>C communication control.

Figure 9 shows the Mainloop flow.





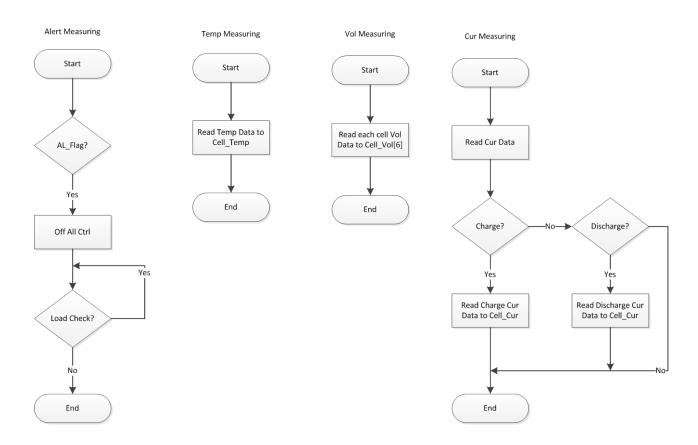
#### 3.2.1.2 Measure and Function Loop

#### **Measuring Section**

- Alert Measuring monitors whether there is short circuit among the batteries of pack.
- Temp Measuring monitors the temperature for pack.
- Vol Measuring monitors the voltage for each cell.
- Cur Measuring monitors the current for pack.

In the Measuring section, although it is only 10-bit ADC result for MSP430G2xx2, it can be calibrated into a high-accuracy value with the particular compensated algorithmic of correction factors in bq76925. TI does not recommend that users modify the particular compensated algorithmic in Measuring.

Figure 10 shows the Measuring flow.







### **Function Section**

- Func Set sets the flag for each cell status.
- Func Ctrl controls the cell in the right status by the flag of cells.
- Temp Ctrl protects the pack from OT and UT.
- Cur Ctrl protects the pack from OC and UC.
- Vol Ctrl protects the pack from OV and UV.
- Balance Ctrl balances the cell current when the voltage and offset are both over range.
- Sleep Check Ctrl, lets BQ go to sleep by checking the Min-voltage of cells.
- Open Cell Check Ctrl checks if the cell pack was open in connect.
- LPM Check Ctrl lets the MSP430 go to sleep by checking the idle time.
- Func Resume reset s the flag by the cells status.
- Gauging Display shows the capacity of the pack (increased number of LEDs indicates greater capacity).

To match the different condition of protected mode, the user can also customize the parameter which is defined in user.h (see Appendix A). For example, the threshold value.

Figure 11 shows the Function flow.

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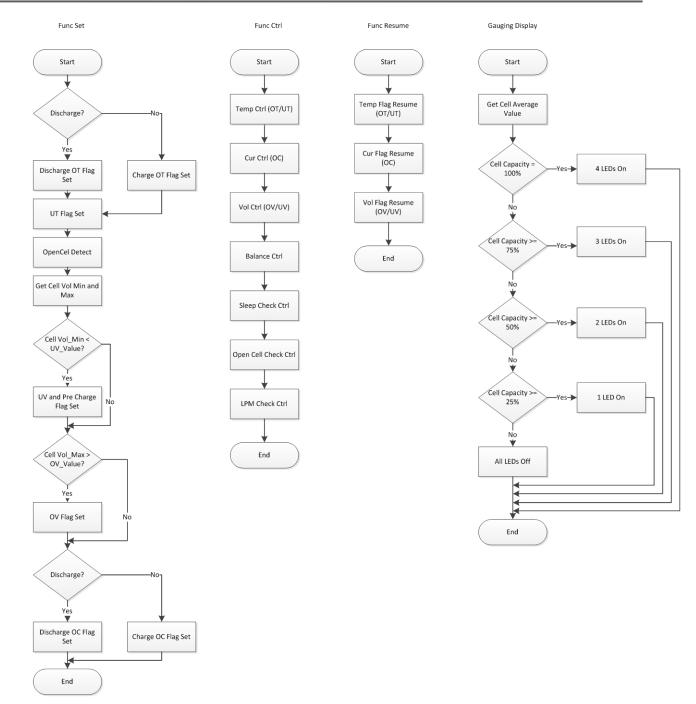


Figure 11. Function Flow

## 3.2.2 Accuracy

A test platform was set up with battery pack to test the measurement accuracy of cell voltage, current, and temperature. We determined the Calculated Value by the MSP430G2 software algorithm. See the following tables.

## **Cell Voltage**

Cell	ADC Value (mV)	Calculated Value (mV)	Measured Value (mV)	Delta (mV)
1	2179	3631.7	3636.6	4.9
2	2180	3633.3	3638.2	4.9
3	2179	3631.7	3633.9	2.2
4	2182	3636.7	3640.6	3.9
5	2162	3603.3	3606.8	3.5
6	2188	3646.7	3650.0	3.3

## Current

ADC Value (mV)	Calculated Value (mV)	Measure Value (mV)	Delta (mV)
111	13.88	14.67	0.80
366	45.75	46.14	0.39

### Temperature

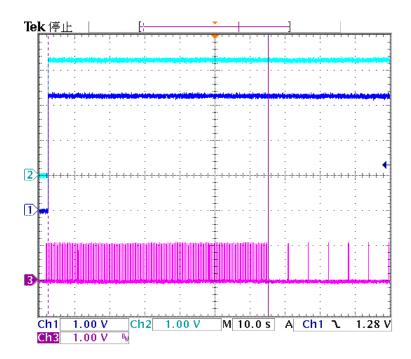
ADC Value	Measured Value	
(mV)	(mV)	Delta (mV)
1068	1065.1	-2.9



### 3.2.3 Low-power Mode

Set up the test platform with battery pack. When the system continues to remain in normal mode (there is no OV, UV, COC, DOC, SC, charge current, discharge current, and balance) for longer than a user-defined time, MSP430G2 enters low-power mode (LPM). In LPM, MSP430G2 only keeps the current comparator enabled for SC detection, while turning off bq76925 current amplifier, cell amplifier, thermistor bias, and voltage reference. Then MSP430G2 enter s LPM to save power consumption. Also, LPM is kept for a user-defined time. MSP430G2 does not detect anything during this time and does not wake up until time-out. After wakeup, if there is still no OV, UV, COC, DOC, SC, charge current, discharge current, and balance, MSP430G2 re-enters LPM. If any of these events occur, the system wakes up from LPM.

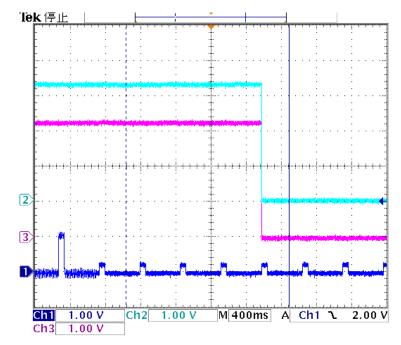
Current	Total Current (uA)	MCU Current (uA)	925 Current (uA)
Normal	165	115	50
LMP	55	43	12



Ch1 = MSP430 DSG pin, Ch2 = MSP430 CHG pin, Ch3 = Voltage on NTC

## 3.2.4 Over Temperature (OT)

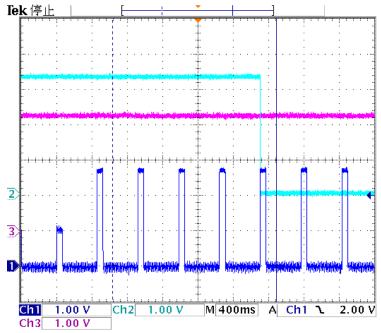
Set up the test platform with battery pack. The system detects the voltage on NTC once in main loop. If the voltage is lower than the OT protection threshold, the system will keep detecting it; after voltage detection exceeds the user-defined times, both CHG and DSG will be off.



Ch1 = Voltage on NTC, Ch2 = MSP430 CHG pin, CH3 = MSP430 DSG pin

## 3.2.5 Under Temperature (UT)

Set up the test platform with battery pack. The system detects the voltage on NTC once in main loop. If the voltage is higher than the UT protection threshold, the system will keep detecting it; after voltage detection exceeds the user-defined times, DSG will be on, but CHG will be off.



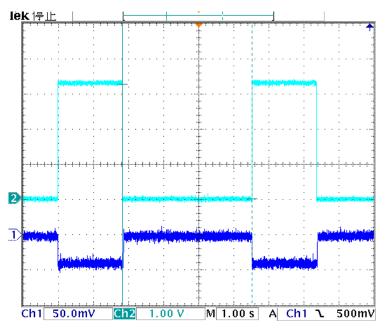
Ch1 = Voltage on NTC, Ch2 = MSP430 CHG pin, Ch3 = MSP430 DSG pin

## 3.2.6 Over Current (OC)

OC occurs when battery is in discharge (DOC) and in charge (COC), and the protection is different.

• COC: OC in charge

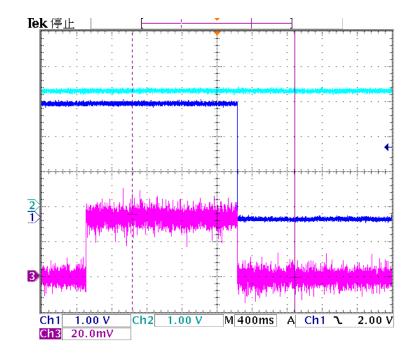
Set up the test platform with a DC source to charge the battery. The system detects the charge current once in main loop. If the charge current is higher than the COC protection threshold, the system will keep detecting it; after charge current detection exceeds the user-defined times, CHG will be off. After 4 seconds in COC protection, CHG will be on again.



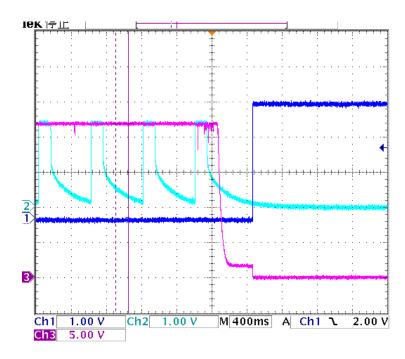
*Ch1* = *Voltage on sense resistor tranferred from Current, Ch2* = *MSP430 CHG pin* 

• DOC: OC in battery discharge

Set up the test platform with battery pack. The system detects the discharge current once in main loop. If the discharge current is higher than the DOC protection threshold, the system will keep detecting it; after discharge current detection exceeds the user-defined times, DSG will be off. If the load is removed in DOC protection, DSG will be on again.



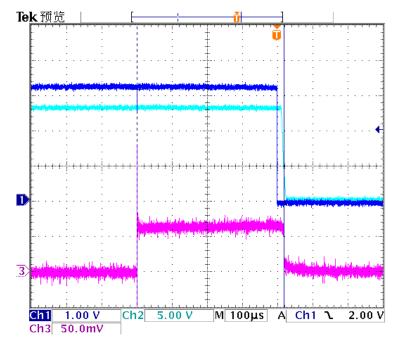
*Ch1* = *MSP430 DSG pin*, *Ch2* = *MSP430 CHG pin*, *Ch3* = *Voltage on sense resistor* 



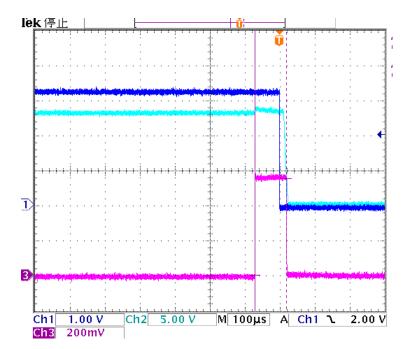
Ch1 = MSP430 DSG pin, Ch2 = MSP430 load detection pin, Ch3 = Pack-

## 3.2.7 Short Current (SC)

Set up the test platform with the real battery. System will detect the discharge current once in main loop, if it is higher than the SC protection threshold, system will keep detecting it and after it exceeds 80  $\mu$ s, DSG will be off. The SC delay time depends on the voltage on the sense resistor. The higher for the voltage, the less for the SC delay time.

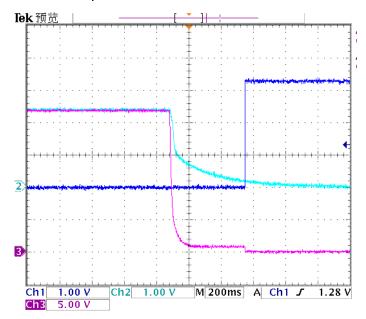


SC threshold is 50 mV, SC delay time is 418  $\mu$ s, Ch1 = MSP430 DSG pin, Ch2 = Gate of MOS, Ch3 = Voltage on sense resistor current transferred by current



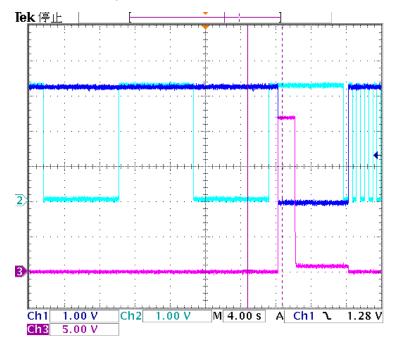
SC threshold is 50 mV, SC delay time is 90  $\mu$ s, Ch1 = MSP430 DSG pin, Ch2 = Gate of MOS, Ch3 = Voltage on sense resistor transferred by current

During SC protection, MSP430G2 simply waits for the release condition. If the load is removed, DSG will be on after one main loop time.



Ch1 = MSP430 DSG pin, Ch2 = MSP430 load detection pin, Ch3 = Pack-

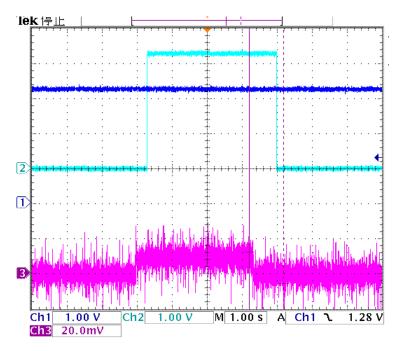
In LPM, DSG will be off immediately. MSP430G2 will not wake up from LPM until timer-out.



*Ch1* = *MSP430 DSG pin*, *Ch2* = *Main loop count*, *Ch3* = *Pack*-

## 3.2.8 Over Voltage (OV)

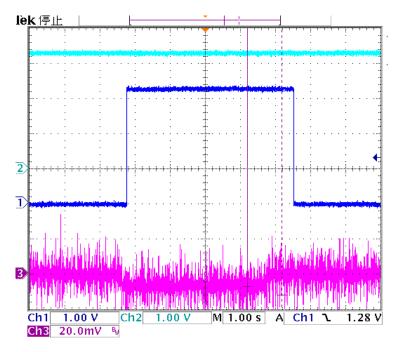
Set up the test platform with battery emulator of resistance. Adjust the DC source output voltage to higher than the OV protection threshold. The system detects the cell voltage once in main loop. If any cell voltage is higher than the OV protection threshold, the system will keep detecting it; after cell voltage detection exceeds the user-defined times, the OV protection will be enabled to turn off CHG. After OV protection, CHG will remain off. The OV protection, however, will turn on CHG if there is discharge current, and then CHG will turn off again if there is no discharge current.



*Ch1* = *MSP430 DSG pin, Ch2* = *MSP430 CHG pin, Ch3* = *Voltage on sense resistor* 

## 3.2.9 Under Voltage (UV)

Set up the test platform with battery emulator of resistance. Adjust the DC source to less than the UV protection threshold. Use another DC on PACK+ and PACK– to act as the charger. The system detects the cell voltage once in main loop. If any cell voltage is less than the UV protection threshold, the system will keep detecting it; after cell voltage detection exceeds the user-defined times, the UV protection will be enabled to turn off DSG. After UV protection, DSG will remain off. The UV protection, however, will turn on DSG if there is charge current, and then DSG will turn off again if there is no charge current.



*Ch1* = *MSP430 DSG pin*, *Ch2* = *MSP430 CHG pin*, *Ch3* = *Voltage on sense resistor* 

## Appendix A. user.h

This appendix provides easy-to-understand explanations of the data in user.h.

// Define the operation function for choosing // Defined => Chosen, Undefined => Not chosen //#define MSP430 ADC Cal 1 //Enabled if ADC value need calculated by Gain and Offset //#define ADC MultSamples 1 //Enabled if ADC need sample for multiple times to get the average value //#define PRE RELEASE SAMPLES 1 //Enabled if it is the prereleased samples #define OpenCell ADCDet 1 //Enabled if open cell need to be detected // Define references for measurement calculations 11 // VREF\_NOM : Nominal reference voltage. Set by writing REF\_SEL in the CONFIG\_2 register (REF\_SEL = 0, 1 => VREF\_NOM = 1.5, 3.0). // // GVCOUT : Nominal cell voltage gain. Also set by writing REF\_SEL in the CONFIG 2 register (REF SEL = 0, 1 => GVCOUT = 0.3, 0.6) . // // GVSENSE : Current amplifier gain. Set by writing the I GAIN bit in the CONFIG\_1 register (I\_GAIN = 0, 1 => GVSENSE = 4, 8).  $\Pi$ // FS CNT : ADC full-scale count. For an 'n' bit ADC, FS CNT = pow(2,n) - 1. #define VREF NOM 3 #define GVCOUT 0.6 #define GVSENSE 8 #define FS CNT 1023L // Define fault thresholds //// The fault thresholds are transformed using function like macros. The trans-// formation reduces code size by avoiding floating-point arithmetic and // division. //// VC TRANS(x) : Transforms voltage threshold. 'x' is in Volts. // VI TRANS(x) : Transforms current threshold. 'x' is in Volts. // VT\_TRANS(x) : Transforms temperature threshold. 'x' is in Volts. #define VC TRANS(x) ((unsigned int) (x \* GVCOUT \* 1000)) //unit mV #define OV THRESHValue // OV threshold in Volts/V 4.250 #define OV\_ResetValue 4.050 // OV Reset threshold in Volts #define UV\_THRESHValue 2.800 // UV threshold in Volts #define UV ResetValue 3.000 // UV Reset threshold in Volts #define PreCHG Value 2.500 //Cell voltage for Prechange in Volts //Cell reset voltage for Prechange in Volts #define PreCHG ResetValue 2.600 #define CELL BalMax //Cell max voltage value for balance in Volts 3.500 #define CELL BalDif 0.100 //Cell difference voltage value for balance in Volts #define BQSleep THRESHValue 1.500 // BQSleep threshold in Volts/V

#define BQSleep\_ResetValue

1.600

define VI_TRANS(x) ((unsigned int) (x * define DOC_THRESHValue 20.000//1 define COC_THRESHValue 20.000//1 define OC_ResetValue 19.000//1	// OC threshold in Volts //2mv 1mohm => 2*8 = 16(A // OC threshold in Volts //2mv 1mohm => 2*8 = 16(A // OC threshold in Volts //2mv 1mohm => 2*8 = 16(A)
	* 1000)) //unit mV
define DOT_THRESHValue 0.469 define DOT_ResetValue 0.547	<pre>// DOT threshold in Volts/V, 0.469V-&gt;70 degree // DOT Reset in Volts, 0.547V-&gt;65 degree</pre>
define COT_THRESHValue 0.855	// COT threshold in Volts/V, 0.855V->50 degree
define COT ResetValue 0.992	// COT Reset in Volts, 0.992V->45 degree
define UT_THRESHValue 2.475	// UT threshold in Volts/V, 2.475->0 degree
define UT_ResetValue 2.360	// UT Reset in Volts, 2.360V->5 degree
	f 1000)) //unit mV
define OL_THRESHValue 2.000	// OverLoad Reset in Volts/V
/ =====================================	
, / Define Protect condition enable times.	
/ Detect multiple times to confirm the pro	
	ct times to delay confirm
—	ct times to delay confirm
	arge OC detect times to delay confirm
	ge OC detect times to delay confirm
	DC detect times to delay confirm
	ge OC detect times to delay confirm
_	t times to delay confirm
_	narge detect times to delay confirm ect times to delay confirm
—	pen detect times to delay confirm
	en detect once in mainLoop times
	ar once in mainLoop times when keep cell open
	ar once in main200p times when keep cen open
/ =====================================	
/ Define operation parameter.	
/ Set it for the battery-management syste	
•	ent value offset set for charge and discharge confirm
—	clear and time for resume from OC, 255*8ms = 2s
	r cell balance time set, 255*8ms = 2s
	eeping time in LPM, $125*32$ ms = 4s
	me to go to LPM when idle, 2000 -> 16S t times for load check to confirm
—	argin set for open load check, 60->0.6 mV
/======================================	=======================================
/ ====================================	

// BQ Reset threshold in Volts



#define V_SOC100 4.180	//Cell voltage set when in 100% coulometer
#define V_SOC75 3.800	//Cell voltage set when in 75% coulometer
#define V_SOC50 3.400	//Cell voltage set when in 50% coulometer
#define V_SOC25 2.800	//Cell voltage set when in 25% coulometer

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