

BQ25618/619 20mA 終了および 1A 昇圧機能搭載、I²C 制御、1 セル、1.5A バッテリ チャージャ

1 特長

- ウェアラブル アクセサリをアダプタまたはバッテリーから充電するためのシングル チップ ソリューション
- 高効率 1.5MHz 同期整流スイッチ モード降圧チャージャ
 - 0.5A で 95.5%、1A で 94.5% の充電効率
 - ±0.5% の充電電圧レギュレーション (10mV 刻み)
 - 充電電圧、電流、温度のスレッシュホールドの JEITA プロファイルで I²C でプログラム可能
 - 高精度の小さな終了電流、20mA±10mA
 - 2.5 x 2.0 x 1.0mm³ の小さなインダクタ フォーム ファクタ
- 4.6V~5.15V の可変出力電圧による昇圧モード動作
 - 1A 出力で 94% の昇圧効率
 - 充電モードと昇圧モードを切り換えるための制御回路を内蔵
 - PMID_GOOD ピンにより外部 PMOS FET を制御しフォルト条件から保護
- USB 入力、高電圧アダプタ、ワイヤレス電源をサポートするシングル入力
 - 4V~13.5V の入力電圧範囲に対応し、入力の絶対最大定格は 22V
 - I²C (100mA~3.2A、100mA 刻み) により入力電流制限値 (IINDPM) をプログラム可能
 - 最高 5.4V の入力電圧制限値 (VINDPM) による最大電力トラッキング
 - VINDPM スレッシュホールド () がバッテリー電圧に自動的に追従
- Narrow VDC (NVDC) パワー パス管理
 - バッテリー未接続または深放電状態でもシステムを即時オン
- I²C を使用した柔軟な構成と自律的充電によるシステム性能の最適化
- すべての MOSFET、電流センシング、ループ補償を含む高度な統合
- R_{DS(on)} が低い 19.5mΩ の BATFET により、充電損失を最小化し、バッテリー動作時間を延長
 - 出荷モード用の BATFET 制御、およびアダプタあり/なしでの完全なシステムリセット
- 出荷モードで 7μA の低いバッテリーリーク電流
- システム スタンバイ状態で 9.5μA の低いバッテリーリーク電流
- 高精度のバッテリー充電プロファイル
 - ±5% の充電電流レギュレーション
 - ±7.5% の入力電流レギュレーション
 - リモート バッテリ センシングによる高速充電

- バッテリーを完全に充電するためのプログラム可能な トップオフ タイマ

- 安全関連認証:
 - IEC 62368-1 CB 認証

2 アプリケーション

- 民生用のウェアラブルおよびスマートウォッチ
- パーソナル・ケアとフィットネス
- ヘッドセットとヘッドフォン
- イヤホン (完全ワイヤレス、TWS) 充電ケース
- 補聴器の充電ケース

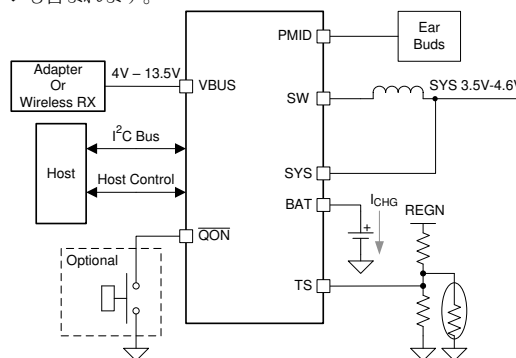
3 概要

BQ25618/619 は、充電、昇圧コンバータ、電圧保護の機能を 1 つのデバイスに統合しています。業界で最も終了電流が小さい本デバイスをスイッチング充電器として使用すると、ウェアラブル デバイスをバッテリーの最大容量まで充電できます。BQ25618/619 のクラス最小の静止電流はバッテリーのリーク電流を最小 6μA (シップ モード) に低減します。そのため、バッテリーのエネルギーを節約し、デバイスの保存可能期間を増加させることができます。BQ25619 は、レイアウトが簡単な 4mm × 4mm の QFN パッケージで供給されます。BQ25618 は、スペースの制約が厳しい設計に適した 2.0mm × 2.4mm WCSP パッケージで供給されます。

パッケージ情報

部品番号	パッケージ (1)	パッケージ サイズ (2)	本体サイズ (公称)
BQ25618	YFF (DSBGA, 30)	2.20mm × 2.60mm	2.00mm × 2.40mm
BQ25619	RTW (WQFN, 24)	4.00mm × 4.00mm	4.00mm × 4.00mm

- 供給されているすべてのパッケージについては、[セクション 13](#) を参照してください。
- パッケージ サイズ (長さ × 幅) は公称値であり、該当する場合はピンも含まれます。



アプリケーション概略図



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4 概要 (続き)

BQ25619/618 は、リチウムイオンおよびリチウムポリマ バッテリ用の高度に統合された 1.5A スイッチ モード バッテリ充電管理およびシステム パワー パス管理デバイスです。高い入力電圧による高速充電をサポートしているため、ウェアラブル、イヤホンの充電ケースなどの幅広いアプリケーションに対応できます。電力パスのインピーダンスが低いいため、スイッチモード動作効率の最適化、バッテリ充電時間の短縮、放電フェーズ中のバッテリ駆動時間の延長が実現できます。入力電圧および電流レギュレーション、小さい終了電流、バッテリのリモート センシングにより、バッテリに最大限の充電電力を供給できます。このソリューションは、入力逆電流ブロック FET (RBFET、Q1)、ハイサイド スイッチング FET (HSFET、Q2)、ローサイド スイッチング FET (LSFET、Q3)、およびシステムとバッテリの間にあるバッテリ FET (BATFET、Q4) を高度に統合しています。また、ハイサイド ゲートドライブ用のブートストラップ ダイオードを内蔵し、システム設計の簡素化を実現しています。充電とシステムの設定に I²C シリアル インターフェイスを使用できる非常に柔軟性が高いソリューションです。

このデバイスは、標準の USB ホスト ポート、USB 充電ポート、USB 対応高電圧アダプタ、ワイヤレス電源など、幅広い入力ソースをサポートしています。このデバイスは、入力電流および電圧のレギュレーションにより、USB 2.0 および USB 3.0 の電力仕様に準拠しています。本デバイスは、システム内の検出回路 (USB PHY デバイスなど) から結果を取得します。

本デバイスには降圧チャージャと昇圧レギュレータが統合されているため、1 つのインダクタを使ってソリューションが完成します。昇圧モードでは、PMID ピンに 5V (4.6V/4.75V/5V/5.15V で可変) を供給します。昇圧モードを使用して BOM を節約できます。また、PMID_GOOD を制御することにより、別のバッテリを充電できます。PMID_GOOD ピンを使用して外部 PMOS FET を駆動し、接続されたアクセサリから昇圧出力の PMID を切り離すことができます。

パワー パス管理により、システムは、アダプタが接続された状態でバッテリ電圧より少し高く、かつ最低システム電圧の 3.5V (プログラム可能) を下回らない電圧に安定化されます。この機能により、システムはバッテリが完全に消耗したとき、または取り外したときでも、動作を継続できます。入力電流制限値または電圧制限値に達すると、パワー パス管理機能が自動的に充電電流を低下させます。システム負荷が増加し続けるとバッテリは放電を開始し、システムの電力要件が満たされるまで放電を続けます。この補助モードにより入力ソースの過負荷を防止します。

このデバイスはソフトウェア制御なしに、充電サイクルを開始 完了できます。バッテリ電圧を感知し、プレコンディショニング、定電流、定電圧という 3 フェーズでバッテリを充電します。充電サイクルの終了時、充電電流があらかじめ設定された制限値を下回り、バッテリ電圧が再充電スレッシュホールドを上回ると、チャージャは自動的に処理を終了します。十分に充電されたバッテリが再充電スレッシュホールドを下回ると、チャージャは自動的に次の充電サイクルを開始します。

このチャージャは、バッテリの負温度係数サーミスタ監視、充電安全タイマ、過電圧および過電流保護など、バッテリ充電とシステム運用のための多様な安全機能を備えています。熱レギュレーションは、接合部温度が 110°Cを超えると充電電流を低減させます。ステータス レジスタは、充電ステータスとすべてのフォルト条件を報告します。I²C により、VBUS_GD ビットは良好な電源が存在するかどうかを示し、フォルトが発生すると INT 出力がただちにホストに通知します。

また、このデバイスは QON ピンを備えています。このピンを使用して BATFET のイネーブルとリセットを制御し、低消費電力の出荷モードの終了やシステム全体のリセットが可能です。

BQ25619 デバイスは、24 ピン、4mm × 4mm × 0.75mm の薄型 WQFN パッケージ、BQ25618 は、30 ボール、2.0mm × 2.4mm の WCSP パッケージで供給されます。

5 Pin Configuration and Functions

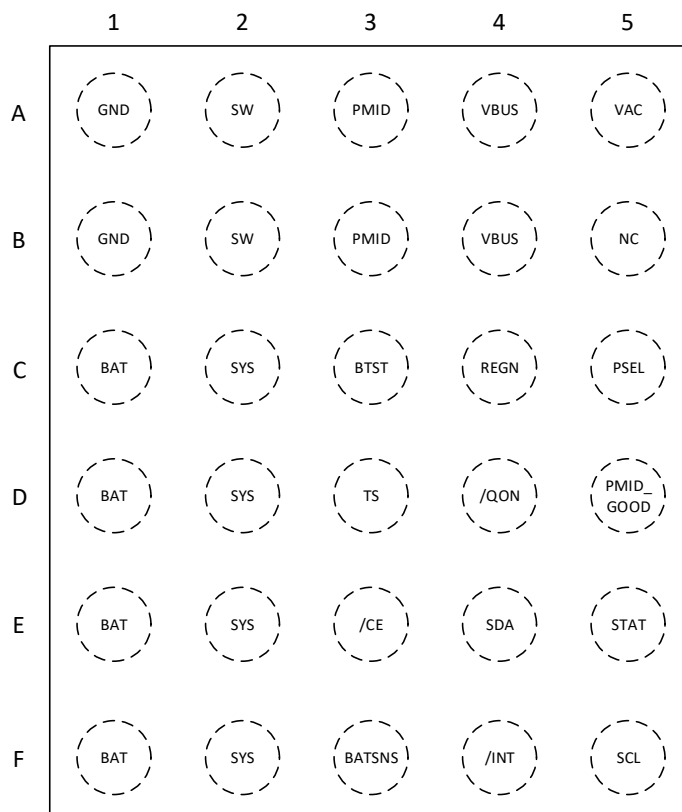


图 5-1. BQ25618 YFF Package 30-Pin WCSP Top View

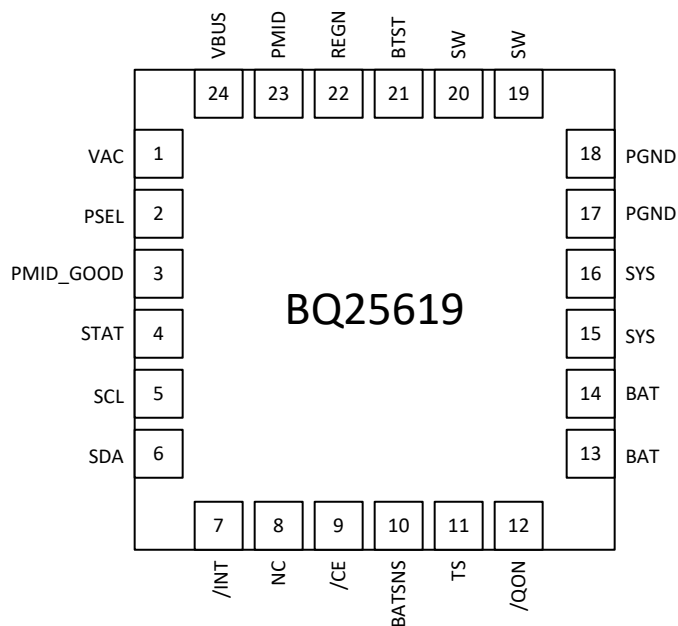


图 5-2. BQ25619 RTW Package 24-Pin WQFN Top View

表 5-1. Pin Functions

PIN			TYPE ⁽¹⁾	DESCRIPTION
NAME	BQ25618 NO.	BQ25619 NO.		
BAT	C1, D1, E1, F1	13	P	Battery connection point to the positive terminal of the battery pack. The internal current sensing resistor is connected between SYS and BAT. Connect a 10 μ F closely to the BAT pin.
		14		
BATSNS	F3	10	AIO	Battery voltage sensing pin for charge voltage regulation. In order to minimize the parasitic trace resistance during charging, BATSNS pin is connected to the positive terminal of battery pack as close as possible. If BATSNS pin is open or short to ground, BATSNS_STAT bit is set to 1 and charger regulates the battery voltage through BAT pin.
BTST	C3	21	P	PWM high-side driver positive supply. Internally, the BTST is connected to the cathode of the boot-strap diode. Connect the 0.047- μ F bootstrap capacitor from SW to BTST.
CE	E3	9	DI	Charge enable pin. When this pin is driven LOW, battery charging is enabled.
GND	A1, B1	17	P	Ground
		18		
INT	F4	7	DO	Open-drain interrupt output. Connect the INT to a logic rail through a 10-k Ω resistor. The INT pin sends an active low, 256- μ s pulse to the host to report charger device status and fault.
NC	B5	8		Not connected
PMID	A3, B3	23	DO	Boost mode output. Connected to the drain of the reverse blocking MOSFET (RBFET) and the drain of HSFET. Consider the total input capacitance, put 1 μ F on VBUS to GND, and the rest capacitance on PMID to GND (typical 2x4.7 μ F plus 1 nF).
PMID_GOOD	D5	3	DO	Open drain active high PMID good indicator. Connect to the pullup rail REGN through 10-k Ω resistor. HIGH indicates PMID voltage is below 5.2 V and the current through Q1 is below 110% of input current limit. This signal can be used to drive external PMOS FET to disconnect the PMID under charging load when Boost mode output voltage is too high or output current is too high.
PSEL	C5	2	DI	Power source selection input. HIGH indicates 500-mA input current limit. LOW indicates 2.4-A input current limit. Once the device gets into Host mode, the host can program a different input current limit to the IINDPM register.
$\overline{\text{QON}}$	D4	12	DI	BATFET enable/reset control input. When the BATFET is in Ship mode, a logic LOW of t_{SHIPMODE} duration turns on BATFET to exit Ship mode. When the BATFET is not in Ship mode, a logic LOW of $t_{\text{QON_RST}}$ (minimum 8 s) duration resets SYS (system power) by turning BATFET off for $t_{\text{BATFET_RST}}$ (minimum 250 ms) and then re-enables BATFET to provide full system power reset. The host chooses the BATFET reset function with VBUS unplug or not through I ² C bit BATFET_RST_WVBUS. The pin is pulled up to V _{BAT} through 200 k Ω to maintain default HIGH logic during Ship mode. It has an internal clamp to 6.5 V.
REGN	C4	22	P	PWM low-side driver positive supply output. Internally, REGN is connected to the anode of the bootstrap diode. Connect a 4.7- μ F (10-V rating) ceramic capacitor from REGN to analog GND. The capacitor should be placed close to the IC.
SCL	F5	5	DI	I ² C interface clock. Connect SCL to the logic rail through a 10-k Ω resistor.
SDA	E4	6	DIO	I ² C interface data. Connect SDA to the logic rail through a 10-k Ω resistor.
STAT	E5	4	DO	Open-drain interrupt output. Connect the STAT pin to a logic rail via 10-k Ω resistor. The STAT pin indicates charger status. Charge in progress: LOW Charge complete or charger in Sleep mode: HIGH Charge suspend (fault response): Blink at 1 Hz
SW	A2, B2	19	P	Switching node connecting to output inductor. Internally SW is connected to the source of the n-channel HSFET and the drain of the n-channel LSFET. Connect the 0.047- μ F bootstrap capacitor from SW to BTST.
		20		
SYS	C2, D2, E2, F2	15	P	Converter output connection point. The internal current sensing resistor is connected between SYS and BAT. Connect a 10 μ F (min) closely to the SYS pin.
		16		
TS	D3	11	AI	Battery temperature qualification voltage input. Connect a negative temperature coefficient thermistor (NTC). Program temperature window with a resistor divider from REGN to TS to GND. Charge and Boost mode suspend when TS pin voltage is out of range. When TS pin is not used, connect a 10-k Ω resistor from REGN to TS and a 10-k Ω resistor from TS to GND or set TS_IGNORE to HIGH to ignore TS pin. It is recommended to use a 103AT-2 thermistor.
VAC	A5	1	AI	Input voltage sensing. This pin must be tied to VBUS.

表 5-1. Pin Functions (続き)

PIN			TYPE ⁽¹⁾	DESCRIPTION
NAME	BQ25618 NO.	BQ25619 NO.		
VBUS	A4, B4	24	P	Charger input voltage. The internal n-channel reverse block MOSFET (RBFET) is connected between VBUS and PMID with VBUS on source. Place a 1- μ F ceramic capacitor from VBUS to GND and place it as close as possible to IC.
Thermal Pad	—	—	P	Ground reference for the device that is also the thermal pad used to conduct heat from the device. This connection serves two purposes. The first purpose is to provide an electrical ground connection for the device. The second purpose is to provide a low thermal-impedance path from the device die to the PCB. This pad should be tied externally to a ground plane.

(1) AI = Analog input, AO = Analog Output, AIO = Analog input Output, DI = Digital input, DO = Digital Output, DIO = Digital input Output, P = Power

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
Voltage	VAC (converter not switching)	–2	22	V
Voltage	VBUS (converter not switching)	–2	22	V
Voltage	PMID (converter not switching)	–0.3	22	V
Voltage	SW	–0.3	16	V
Voltage	BAT, SYS (converter not switching)	–0.3	7	V
Voltage	BTST	–0.3	22	V
Voltage	BATSNS (converter not switching)	–0.3	7	V
Voltage	PSEL, STAT, SCL, SDA, INT, PMID_GOOD, CE, TS, QON	–0.3	7	V
Output Sink Current	SDA, STAT, INT, PMID_GOOD		6	mA
T _J	Junction temperature	–40	150	°C
T _{stg}	Storage temperature	–55	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Rating* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Condition*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	±2000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±250	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V _{VBUS}	Input voltage	4		13.5	V
V _{BAT}	Battery voltage			4.52	V
I _{VBUS}	Input current			3.2	A
I _{SW}	Output current (SW)			1.8	A
I _{BAT}	Fast charging current			1.5	A
I _{BAT}	RMS discharge current			5	A
T _A	Ambient temperature	–40		85	°C
L	Inductance		1	2.2	μH
C _{VBUS}	VBUS capacitance		1		μF
C _{PMID}	PMID capacitance		10		μF
C _{SYS}	SYS capacitance		10		μF
C _{BAT}	BAT capacitance		10		μF
C _{REGN}	REGN capacitance		4.7		μF

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		BQ25618	UNIT
		YFF (DSBGA)	
		30 Balls	
$R_{\theta JA}$	Junction-to-ambient thermal resistance (JEDEC ⁽¹⁾)	58.8	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	0.2	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	8.3	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	1.4	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	8.3	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.5 Thermal Information

THERMAL METRIC ⁽¹⁾		BQ25619	UNIT
		RTW (WQFN)	
		24 Pins	
$R_{\theta JA}$	Junction-to-ambient thermal resistance (JEDEC ⁽¹⁾)	35.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	22.7	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	11.9	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	0.2	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	12	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	2.6	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.6 Electrical Characteristics

$V_{VBUS_UVLOZ} < V_{VBUS} < V_{VBUS_OV}$ and $V_{VBUS} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$, and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
QUIESCENT CURRENTS						
I _{Q_BAT}	Quiescent battery current (BATSNS, BAT, SYS, SW)	VBAT = 4.5V, VBUS floating, SCL, SDA = 0V or 1.8V, T _J < 85 °C, BATFET enabled		9.5	15	μA
I _{SHIP_BAT}	Shipmode battery current (BATSNS, BAT, SYS, SW)	VBAT = 4.5V, VBUS floating, SCL, SDA = 0V or 1.8V, T _J < 85 °C, BATFET disabled		7	9.5	μA
I _{VBUS}	Input current (VBUS) in buck mode when converter is switching	VBUS=5V, charge disabled, converter switching, ISYS = 0A		2.3		mA
I _{HIZ_VBUS}	Quiescent input current in HIZ	VAC/VBUS = 5V, HIZ mode, no battery		37	50	μA
		VAC/VBUS = 12V, HIZ mode, no battery		68	90	μA
I _{BST}	Quiescent battery current (BATSNS, BAT, SYS, SW) in boost mode when converter is switching	VBAT = 4.5V, VBUS = 5V, boost mode enabled, converter switching, I _{PMID} = 0A		2.4		mA
VBUS / VBAT SUPPLY						
V _{VBUS_OP}	VBUS operating range		4		13.5	V
V _{VBUS_UVLOZ}	VBUS rising for active I2C, no battery	VBUS rising		3.3	3.7	V
V _{VBUS_UVLO}	VBUS falling to turnoff I2C, no battery	VBUS falling		3	3.3	V
V _{VBUS_PRESENT}	VBUS to enable REGN	VBUS rising		3.65	3.9	V

6.6 Electrical Characteristics (続き)

$V_{VBUS_UVLOZ} < V_{VBUS} < V_{VBUS_OV}$ and $V_{VBUS} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$, and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{VBUS_PRESENTZ}	VBUS to disable REGN	VBUS falling	3.15		3.4	V
V _{SLEEP}	Enter Sleep mode threshold	VBUS falling, VBUS - VBAT, VBAT = 4V	15	60	110	mV
V _{SLEEPZ}	Exit Sleep mode threshold	VBUS rising, VBUS - VBAT, VBAT = 4V	115	220	340	mV
V _{ACOV}	VAC overvoltage rising threshold to turn of switching	VAC rising, OVP[1:0]=00	5.45	5.85	6.07	V
		VAC rising, OVP[1:0]=01	6.1	6.4	6.75	V
		VAC rising, OVP[1:0]=10	10.45	11	11.55	V
		VAC rising, OVP[1:0]=11 (default)	13.5	14.2	14.85	V
	VAC overvoltage falling threshold to resume switching	VAC falling, OVP[1:0]=00	5.2	5.6	5.8	V
		VAC falling, OVP[1:0]=01	5.8	6.2	6.45	V
		VAC falling, OVP[1:0]=10	10	10.7	11.1	V
		VAC falling, OVP[1:0]=11 (default)	13	13.9	14.5	V
V _{BAT_UVLOZ}	BAT voltage for active I2C, no VBUS	VBAT rising	2.5			V
V _{BAT_DPLZ}	BAT depletion rising threshold to turn on BATFET	VBAT rising	2.35		2.8	V
V _{BAT_DPL}	BAT depletion falling threshold to turn off BATFET	VBAT falling	2.18		2.62	V
V _{POORSRC}	Bad adapter detection threshold	VBUS falling	3.75	3.9	4.0	V

POWER PATH MANAGEMENT

V_{SYS_MIN}	Typical minimum system regulation voltage	$V_{BAT}=3.2\text{V} < V_{SYS_MIN} = 3.5\text{V}$, $I_{SYS} = 0\text{A}$	3.5	3.65		V
V_{SYS_OVP}	System overvoltage threshold	$V_{REG} = 4.35\text{V}$, Charge disabled, $I_{SYS} = 0\text{A}$		4.7		V
R_{ON_RBFET}	Blocking FET on-resistance (BQ25618)			35		mΩ
R_{ON_RBFET}	Blocking FET on-resistance (BQ25619)			45		mΩ
R_{ON_HSFET}	High-side switching FET on-resistance (BQ25618)			55		mΩ
R_{ON_HSFET}	High-side switching FET on-resistance (BQ25619)			62		mΩ
R_{ON_LSFET}	Low-side switching FET on-resistance (BQ25618)			60		mΩ
R_{ON_LSFET}	Low-side switching FET on-resistance (BQ25619)			71		mΩ
V_{BATFET_FWD}	BATFET forward voltage in supplement mode	BAT discharge current 10mA, converter running		30		mV

BATTERY CHARGER

V_{REG_RANGE}	Typical charge voltage regulation range		3.5		4.52	V
V_{REG_STEP}	Typical charge voltage step	$4.3\text{V} < V_{REG} < 4.52\text{V}$		10		mV
V_{REG_ACC}	Charge voltage accuracy	$V_{REG} = 4.2\text{V}$, $T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$	4.183	4.2	4.2168	V
V_{REG_ACC}	Charge voltage accuracy	$V_{REG} = 4.35\text{V}$, $T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$	4.333	4.35	4.367	V
V_{REG_ACC}	Charge voltage accuracy	$V_{REG} = 4.45\text{V}$, $T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$	4.432	4.45	4.468	V
I_{CHG_RANGE}	Typical charge current regulation range		0		1.5	A

6.6 Electrical Characteristics (続き)

$V_{VBUS_UVLOZ} < V_{VBUS} < V_{VBUS_OV}$ and $V_{VBUS} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$, and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{CHG_STEP}	Typical charge current regulation step			20		mA
I_{CHG_ACC}	Fast charge current regulation accuracy	$ICHG = 0.24\text{A}$, $V_{BAT} = 3.1\text{V}$ or 3.8V , $T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$	0.216	0.24	0.264	A
		$ICHG = 0.72\text{A}$, $V_{BAT} = 3.1\text{V}$ or 3.8V , $T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$	0.684	0.72	0.756	A
		$ICHG = 1.50\text{A}$, $V_{BAT} = 3.1\text{V}$ or 3.8V , $T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$	1.425	1.5	1.575	A
I_{PRECHG_RANGE}	Typical pre-charge current range		20		260	mA
I_{PRECHG_STEP}	Typical pre-charge current step			20		mA
I_{PRECHG_ACC}	Precharge current accuracy	$V_{BAT} = 2.6\text{V}$, $I_{PRECHG} = 40\text{mA}$	28	40	52	mA
		$V_{BAT} = 2.6\text{V}$, $I_{PRECHG} = 120\text{mA}$	84	120	156	mA
I_{TERM_RANGE}	Typical termination current range		20		260	mA
I_{TERM_STEP}	Typical termination current step			20		mA
I_{TERM_ACC}	Termination current accuracy	$I_{TERM}=40\text{mA}$, $ICHG>260\text{mA}$, $V_{REG}=4.35\text{V}$, $T_J = 0^{\circ}\text{C} - 85^{\circ}\text{C}$	30	40	50	mA
		$I_{TERM}=20\text{mA}$, $ICHG<260\text{mA}$, $V_{REG}=4.35\text{V}$, $T_J = 0^{\circ}\text{C} - 85^{\circ}\text{C}$	10	20	30	mA
V_{BAT_SHORTZ}	Battery short voltage rising threshold to start pre-charge	V_{BAT} rising	2.13	2.25	2.35	V
V_{BAT_SHORT}	Battery short voltage falling threshold to stop pre-charge	V_{BAT} falling	1.85	2	2.15	V
I_{BAT_SHORT}	Battery short trickle charging current	$V_{BAT} < V_{BAT_SHORTZ}$	15	25	30	mA
$V_{BATLOWV}$	Battery LOWV rising threshold to start fast-charge	V_{BAT} rising	3	3.12	3.24	V
	Battery LOWV falling threshold to stop fast-charge	V_{BAT} falling	2.7	2.8	2.9	V
V_{RECHG}	Battery recharge threshold	$V_{RECHG}=0$, V_{BAT} falling (default)	90	120	150	mV
		$V_{RECHG}=1$, V_{BAT} falling	185	210	245	mV
I_{SYS_LOAD}	System discharge load current during SYSOVP			30		mA
R_{ON_BATFET}	Battery FET on-resistance	$T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$		19.5	26	m Ω
		$T_J = -40^{\circ}\text{C} - 125^{\circ}\text{C}$		19.5	30	m Ω

BATTERY OVERVOLTAGE PROTECTION

V_{BAT_OVP}	Battery overvoltage rising threshold	V_{BAT} rising, as percentage of V_{REG}	103	104	105	%
	Battery overvoltage falling threshold	V_{BAT} falling, as percentage of V_{REG}	101	102	103	%

INPUT VOLTAGE / CURRENT REGULATION

V_{INDPM_RANGE}	Typical input voltage regulation range		3.9		5.4	V
V_{INDPM_STEP}	Typical input voltage regulation step			100		mV
V_{INDPM_ACC}	Typical input voltage regulation accuracy		4.365	4.5	4.635	V
V_{INDPM_TRACK}	V_{INDPM} threshold to track battery voltage	$V_{BAT} = 4.35\text{V}$, $V_{INDPM_BAT_TRACK} = V_{BAT}+200\text{mV}$	4.45	4.55	4.74	V
I_{INDPM_RANGE}	Typical input current regulation range		0.1		3.2	A
I_{INDPM_STEP}	Typical input current regulation step			100		mA
I_{INDPM_ACC}	Input current regulation accuracy	$I_{INDPM} = 500\text{mA}$ ($T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$)	450	465	500	mA
I_{INDPM_ACC}	Input current regulation accuracy	$I_{INDPM} = 900\text{mA}$ ($T_J = -40^{\circ}\text{C} - 85^{\circ}\text{C}$)	750	835	900	mA

6.6 Electrical Characteristics (続き)

$V_{VBUS_UVLOZ} < V_{VBUS} < V_{VBUS_OV}$ and $V_{VBUS} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$, and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{INDPM_ACC}	Input current regulation accuracy	$I_{INDPM} = 1500\text{mA}$ ($T_J = -40^{\circ}\text{C}$ - 85°C)	1300	1390	1500	mA
THERMAL REGULATION AND THERMAL SHUTDOWN						
T_{REG}	Junction temperature regulation accuracy	$T_{REG} = 90^{\circ}\text{C}$		90		$^{\circ}\text{C}$
		$T_{REG} = 110^{\circ}\text{C}$		110		$^{\circ}\text{C}$
T_{SHUT}	Thermal Shutdown Rising threshold	Temperature Increasing		150		$^{\circ}\text{C}$
	Thermal Shutdown Falling threshold	Temperature Decreasing		130		$^{\circ}\text{C}$
CHARGE MODE THERMISTOR COMPARATOR (JEITA 616J or HOT/COLD 616)						
$V_{T1_RISE\%}$	TS pin voltage rising threshold, Charge suspended above this voltage.	As Percentage to REGN (0°C w/ 103AT)	72.4	73.3	74.2	%
$V_{T1_FALL\%}$	TS pin voltage falling threshold. Charge re-enabled to 20% of ICHG and VREG below this voltage.	As Percentage to REGN	71.5	72	72.5	%
$V_{T2_RISE\%}$	TS pin voltage rising threshold, Charge back to 20% of ICHG and VREG above this voltage.	As Percentage to REGN, JEITA_T2= 5°C w/ 103AT	70.25	70.75	71.25	%
		As Percentage to REGN, JEITA_T2= 10°C w/ 103AT	67.75	68.25	68.75	%
		As Percentage to REGN, JEITA_T2= 15°C w/ 103AT	64.75	65.25	65.75	%
		As Percentage to REGN, JEITA_T2= 20°C w/ 103AT	61.75	62.25	62.75	%
$V_{T2_FALL\%}$	TS pin voltage falling threshold. Charge back to ICHG and VREG below this voltage.	As Percentage to REGN, JEITA_T2= 5°C w/ 103AT	68.7	69.2	69.7	%
		As Percentage to REGN, JEITA_T2= 10°C w/ 103AT	66.45	66.95	67.45	%
		As Percentage to REGN, JEITA_T2= 15°C w/ 103AT	63.7	64.2	64.7	%
		As Percentage to REGN, JEITA_T2= 20°C w/ 103AT	60.7	61.2	61.7	%
$V_{T3_FALL\%}$	TS pin voltage falling threshold. Charge to ICHG and 4.1V below this voltage.	As Percentage to REGN, JEITA_T3= 40°C w/ 103AT	47.75	48.25	48.75	%
		As Percentage to REGN, JEITA_T3= 45°C w/ 103AT	44.25	44.75	45.25	%
		As Percentage to REGN, JEITA_T3= 50°C w/ 103AT	40.2	40.7	41.2	%
		As Percentage to REGN, JEITA_T3= 55°C w/ 103AT	37.2	37.7	38.2	%
$V_{T3_RISE\%}$	TS pin voltage rising threshold. Charge back to ICHG and VREG above this voltage.	As Percentage to REGN, JEITA_T3= 40°C w/ 103AT	48.8	49.3	49.8	%
		As Percentage to REGN, JEITA_T3= 45°C w/ 103AT	45.3	45.8	46.3	%
		As Percentage to REGN, JEITA_T3= 50°C w/ 103AT	41.3	41.8	42.3	%
		As Percentage to REGN, JEITA_T3= 55°C w/ 103AT	38.5	39	39.5	%
$V_{T5_FALL\%}$	TS pin voltage falling threshold, charge suspended below this voltage.	As Percentage to REGN (60°C w/ 103AT)	33.7	34.2	35.1	%
$V_{T5_RISE\%}$	TS pin voltage rising threshold. Charge back to ICHG and 4.1V above this voltage.	As Percentage to REGN	35	35.5	36	%

6.6 Electrical Characteristics (続き)

$V_{VBUS_UVLOZ} < V_{VBUS} < V_{VBUS_OV}$ and $V_{VBUS} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$, and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
BOOST MODE THERMISTOR COMPARATOR (HOT/COLD)						
$V_{BCOLD_RISE\%}$	TS pin voltage rising threshold, boost mode is suspended above this voltage.	As Percentage to REGN (-19.5°C w/ 103AT)	79.5	80	80.5	%
$V_{BCOLD_FALL\%}$	TS pin voltage falling threshold	As Percentage to REGN (0°C w/ 103AT)		72		%
$V_{BHOT_FALL\%}$	TS pin voltage threshold. boost mode is suspended below this voltage.	As Percentage to REGN, (64°C w/ 103AT)	30.2	31.2	32.2	%
$V_{BHOT_RISE\%}$	TS pin voltage rising threshold	As Percentage to REGN, (55°C w/ 103AT), REG0C[1:0]=11		39		%
SWITCHING CONVERTER						
F_{SW}	PWM switching frequency	Oscillator frequency	1.32	1.5	1.68	MHz
D_{MAX}	Maximum PWM Duty Cycle			97		%
BOOST MODE CONVERTER						
$V_{BATLOWV_OTG}$	Battery voltage exiting boost mode	V_{VBAT} falling	2.6	2.8	2.9	V
	Battery voltage entering boost mode	V_{VBAT} rising	2.9	3.0	3.15	V
V_{BST_BAT}	Battery voltage exiting boost mode	BAT falling V_{VBAT}	2.4	2.5	2.6	V
V_{BST_RANGE}	Typical boost mode voltage regulation range		4.6		5.15	V
V_{BST_ACC}	Boost mode voltage regulation accuracy	$I_{VBUS} = 0\text{A}$, $BOOST_V = 5\text{V}$	4.85	5	5.15	V
$I_{SYS_OCP_Q4}$	Boost mode battery discharge current clamp on BATFET Q4		5	6		A
V_{BST_OVP}	Boost mode overvoltage threshold on PMID		5.55	5.8	6.15	V
REGN LDO						
V_{REGN}	REGN LDO output voltage	$V_{VBUS} = 5\text{V}$, $I_{REGN} = 20\text{mA}$	4.58	4.7	4.8	V
		$V_{VBUS} = 9\text{V}$, $I_{REGN} = 20\text{mA}$	5.6	6	6.5	V
I_{REGN}	REGN LDO current limit	$V_{VBUS} = 5\text{V}$, $V_{REGN} = 3.8\text{V}$	50			mA
I2C INTERFACE (SCL, SDA)						
V_{IH}	Input high threshold level, SDA and SCL	Pull up rail 1.8V	1.3			V
V_{IL}	Input low threshold level	Pull up rail 1.8V			0.4	V
V_{OL}	Output low threshold level	Sink current = 5mA			0.4	V
I_{BIAS}	High-level leakage current	Pull up rail 1.8V			1	μA
V_{IH_SDA}	Input high threshold level, SDA	Pull up rail 1.8V	1.3			V
V_{IL_SDA}	Input low threshold level	Pull up rail 1.8V			0.4	V
V_{OL_SDA}	Output low threshold level	Sink current = 5mA			0.4	V
I_{BIAS_SDA}	High-level leakage current	Pull up rail 1.8V			1	μA
V_{IH_SCL}	Input high threshold level, SDA	Pull up rail 1.8V	1.3			V
V_{IL_SCL}	Input low threshold level	Pull up rail 1.8V			0.4	V
V_{OL_SCL}	Output low threshold level	Sink current = 5mA			0.4	V
I_{BIAS_SCL}	High-level leakage current	Pull up rail 1.8V			1	μA
LOGIC INPUT PIN						
V_{IH}	Input high threshold level (/CE, PSEL)		1.3			V
V_{IL}	Input low threshold level (/CE, PSEL)				0.4	V

6.6 Electrical Characteristics (続き)

$V_{VBUS_UVLOZ} < V_{VBUS} < V_{VBUS_OV}$ and $V_{VBUS} > V_{BAT} + V_{SLEEP}$, $T_J = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$, and $T_J = 25^{\circ}\text{C}$ for typical values (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{IN_BIAS}	High-level leakage current (/CE, PSEL)	Pull up rail 1.8V			1	μA
LOGIC OUTPUT PIN						
V_{OL}	Output low threshold level (/INT, STAT, PMID_GOOD)	Sink current = 5mA			0.4	V
I_{OUT_BIAS}	High-level leakage current (/INT, STAT, PMID_GOOD)	Pull up rail 1.8V			1	μA

6.7 Timing Requirements

		MIN	NOM	MAX	UNIT
VBUS / VBAT POWER UP					
t_{VBUS_OV}	VBUS OVP Reaction-time		130		ns
$t_{POORSRC}$	Bad adapter detection duration		30		ms
$t_{POORSRC_RETRY}$	Bad adapter detection retry wait time		2		s
BATTERY CHARGER					
t_{TERM_DGL}	Deglitch time for charge termination		30		ms
t_{RECHG_DGL}	Deglitch time for recharge threshold		30		ms
t_{TOP_OFF}	Typical top-off timer accuracy TOP_OFF_TIMER[1:0]=10	24	30	36	min
t_{SAFETY}	Charge safety timer accuracy, CHG_TIMER = 20hr	17	20	24	hr
t_{SAFETY}	Charge safety timer accuracy, CHG_TIMER = 10hr	8	10	12	hr
QON Timing					
$t_{SHIPMODE}$	QON low time to turn on BATFET and exit shipmode ($-10^{\circ}\text{C} \leq T_J \leq 60^{\circ}\text{C}$)	0.9		1.3	s
t_{QON_RST}	QON low time before BATFET full system reset ($-10^{\circ}\text{C} \leq T_J \leq 60^{\circ}\text{C}$)	8		12	s
t_{BATFET_RST}	BATFET off time during full system reset ($-10^{\circ}\text{C} \leq T_J \leq 60^{\circ}\text{C}$)	250		400	ms
t_{BATFET_DLY}	Delay time before BATFET turn off in ship mode ($-10^{\circ}\text{C} \leq T_J \leq 60^{\circ}\text{C}$)	10		15	s
I2C INTERFACE					
f_{SCL}	SCL clock frequency			400	kHz
DIGITAL CLOCK AND WATCHDOG					
f_{LPDIG}	Digital low-power clock (REGN LDO is disabled)		30		kHz
f_{DIG}	Digital power clock		500		kHz
t_{LP_WDT}	Watchdog Reset time		160		s
t_{WDT}	Watchdog Reset time (WATCHDOG REG05[5:4] = 160s)		160		s

6.8 Typical Characteristics

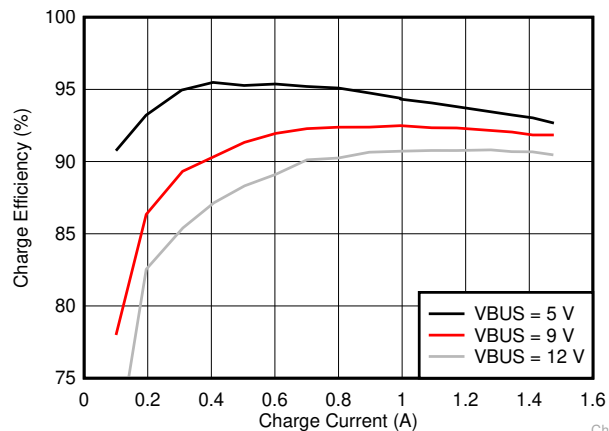
BQ25619EVM $V_{BAT} = 3.8V$ Inductor 2.2 μ H, DCR = 40 m Ω

Figure 6-1. Charge Efficiency

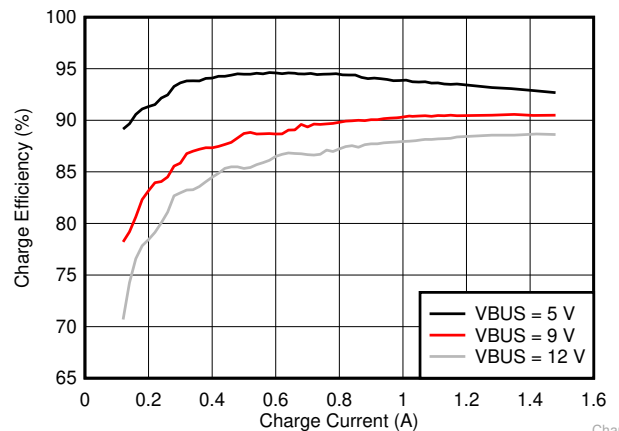
BQ25618EVM $V_{BAT} = 3.8V$ Inductor 1.0 μ H, DCR = 27 m Ω

Figure 6-2. Charge Efficiency

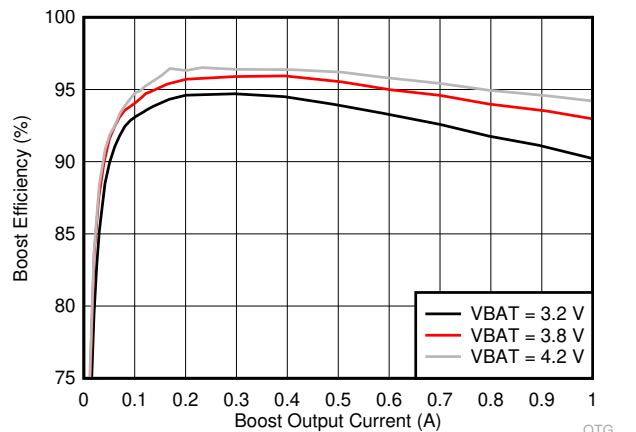
BQ25619EVM $V_{PMID} = 4.6V$ Inductor 2.2 μ H, DCR = 40 m Ω

Figure 6-3. Boost Efficiency

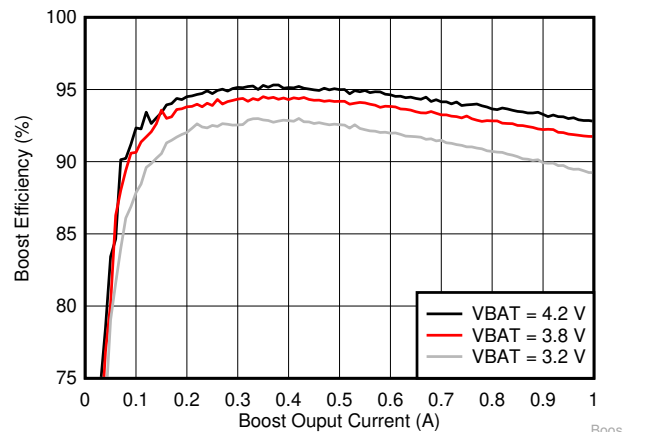
BQ25618EVM $V_{PMID} = 4.6V$ 1.0 μ H Inductor, DCR = 27 m Ω

Figure 6-4. Boost Efficiency

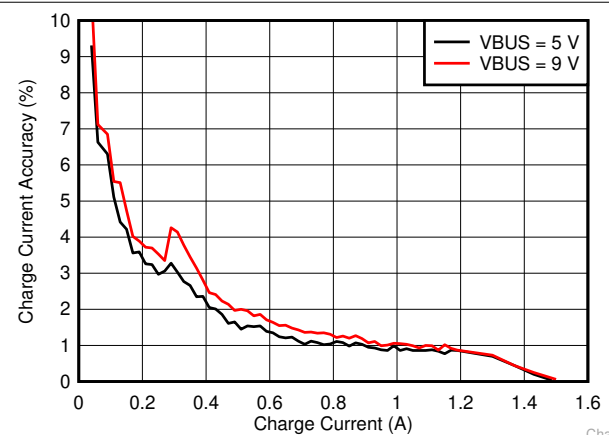


Figure 6-5. Charge Current Accuracy

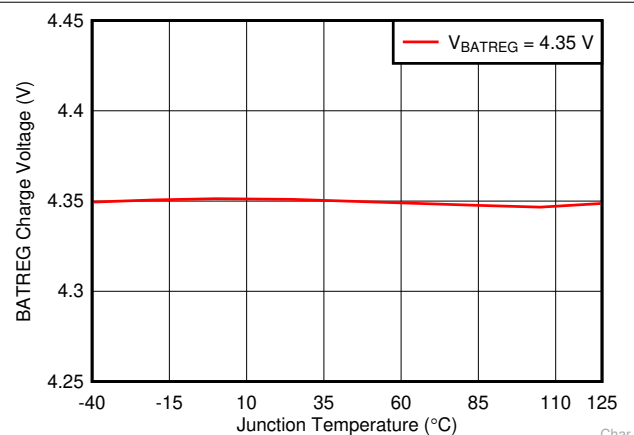


Figure 6-6. Battery Charge Voltage vs Junction Temperature

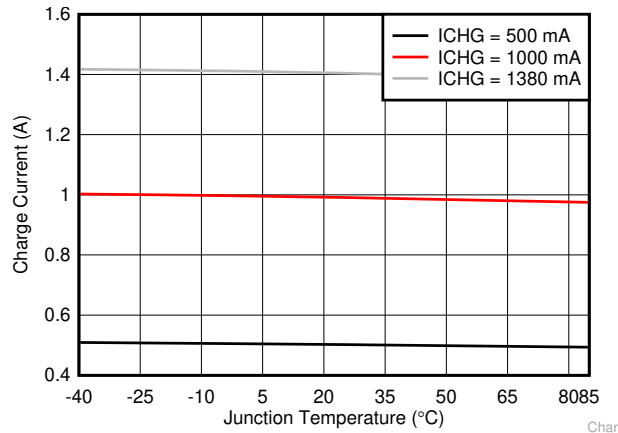


図 6-7. Charge Current vs Junction Temperature

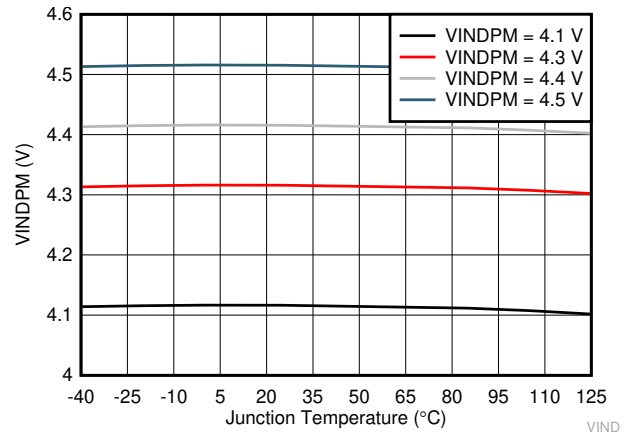


図 6-8. VINDPM vs Junction Temperature

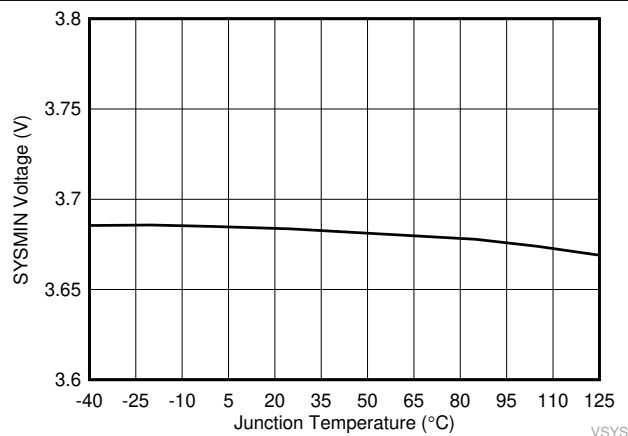


図 6-9. SYSMIN Voltage vs Junction Temperature (VSYS set at 3.5 V)

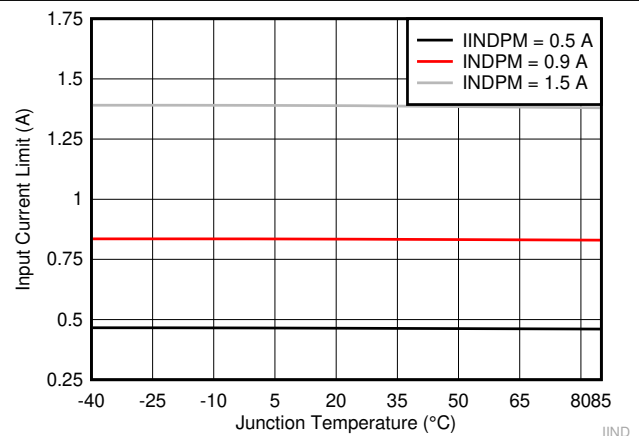


図 6-10. Input Current Limit vs Junction Temperature

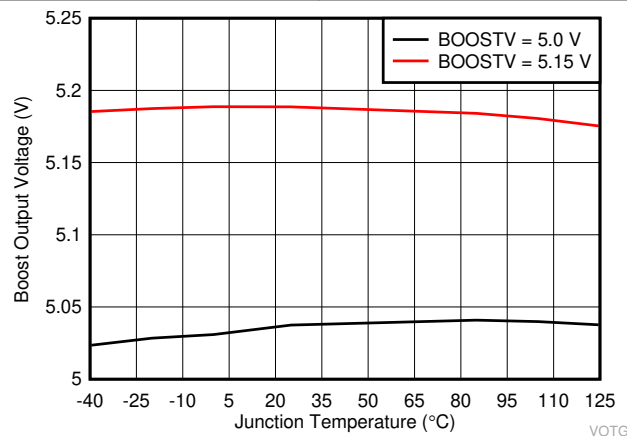


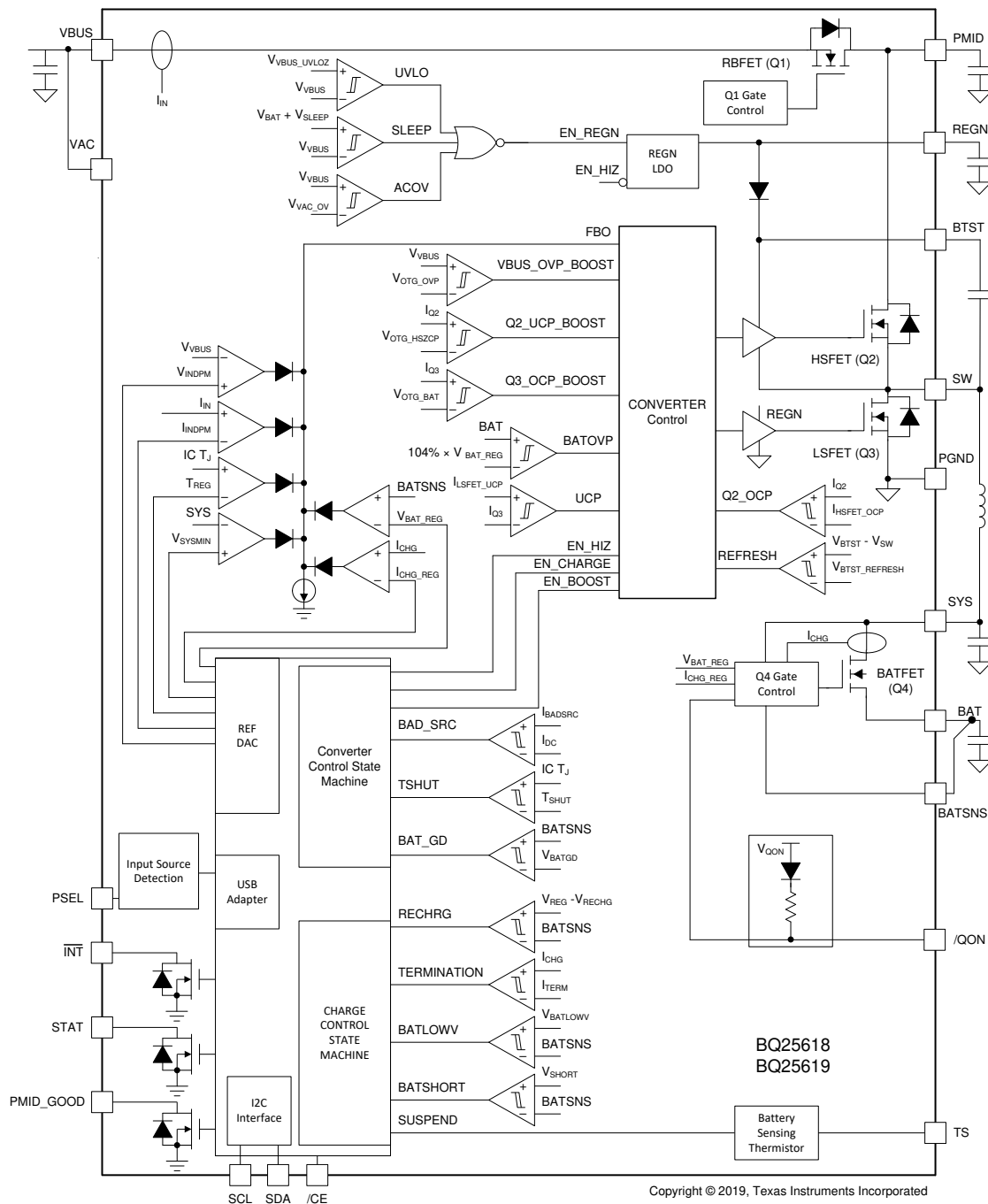
図 6-11. Boost Output Voltage vs Junction Temperature

7 Detailed Description

7.1 Overview

The BQ25619/618 device is a highly integrated 1.5-A switch-mode battery charger for single cell Li-ion and Li-polymer battery. It includes an input reverse-blocking FET (RBFET, Q1), high-side switching FET (HSFET, Q2), low-side switching FET (LSFET, Q3), and battery FET (BATFET, Q4), and bootstrap diode for the high-side gate drive.

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 Power-On-Reset (POR)

The device powers internal bias circuits from the higher voltage of VBUS and BAT. When V_{VBUS} rises above V_{VBUS_UVLOZ} or V_{BAT} rises above V_{BAT_UVLOZ} , the sleep comparator, battery depletion comparator, and BATFET driver are active. The I²C interface is ready for communication and all registers are reset to default values. The host can access all registers after POR.

7.3.2 Device Power Up From Battery Without Input Source

If only the battery is present and the voltage is above depletion threshold (V_{BAT_DPLZ}), the BATFET turns on and connects the battery to the system. The REGN stays off to minimize the quiescent current. The low $R_{DS(on)}$ of BATFET and the low quiescent current on BAT minimize the conduction loss and maximize the battery run time.

The device always monitors the discharge current through the BATFET. When the system is overloaded or shorted ($I_{BAT} > I_{SYS_OCP_Q4}$), the device turns off BATFET immediately.

With I²C, when the BATFET turns off due to overcurrent, the device sets the BATFET_DIS bit to indicate the BATFET is disabled until the input source plugs in again or one of the methods described in [セクション 7.3.7.2](#) is applied to re-enable BATFET.

7.3.3 Power Up From Input Source

When an input source is plugged in, the device checks the input source voltage to turn on the REGN LDO and all the bias circuits. It detects and sets the input current limit before the buck converter is started. The power-up sequence from input source is as listed:

1. Power Up REGN LDO, see [セクション 7.3.3.1](#)
2. Poor Source Qualification, see [セクション 7.3.3.2](#)
3. Input Source Type Detection is based on PSEL to set default input current limit (IINDPM threshold), see [セクション 7.3.3.3](#)
4. Input Voltage Limit Threshold Setting (VINDPM threshold), see [セクション 7.3.3.4](#)
5. Power Up Converter, see [セクション 7.3.3.5](#)

7.3.3.1 Power Up REGN LDO

The REGN LDO supplies internal bias circuits as well as the HSFET and LSFET gate drive. It also provides the bias rail to TS external resistors. The pull-up rail of STAT can be connected to REGN as well. The REGN LDO is enabled when all the below conditions are valid:

- $V_{VBUS} > V_{VBUS_UVLOZ}$
- In buck mode, $V_{VBUS} > V_{BAT} + V_{SLEEPZ}$
- In boost mode, $V_{VBUS} < V_{BAT} + V_{SLEEPZ}$
- After 220-ms delay is completed

During high impedance mode when EN_HIZ bit is 1, REGN LDO turns off. The battery powers up the system.

7.3.3.2 Poor Source Qualification

After the REGN LDO powers up, the device starts to check current capability of the input source. The first step is poor source detection.

- VBUS voltage above $V_{POORSRC}$ when pulling I_{BADSRC} (typical 30 mA)

With I²C, once the input source passes poor source detection, the status register bit VBUS_GD is set to 1 and the INT pin is pulsed to signal to the host.

If the device fails the poor source detection, it repeats poor source qualification every 2 seconds.

7.3.3.3 Input Source Type Detection (IINDPM Threshold)

After poor source detection, the device runs input source detection through the PSEL pin. The PSEL pin sets input current limit 0.5 A (HIGH) or 2.4 A (LOW). After input source type detection is completed, the PMID_GOOD pin is asserted to HIGH and the PG_STAT bit goes to 1.

With I²C, after input source type detection is completed, an $\overline{\text{INT}}$ pulse is asserted to the host. In addition, the following register bits are updated:

1. Input Current Limit (IINDPM) register is updated from detection result
2. VBUS_STAT bit is updated to indicate USB or other input source
3. PG_STAT bit is updated to indicate good adapter plugs in

The host can overwrite the IINDPM register to change the input current limit if needed.

7.3.3.3.1 PSEL Pins Sets Input Current Limit

The device with the PSEL pin directly takes the USB PHY device output to decide whether the input is a USB host or charging port. When the device operates in host-control mode, the host needs the INDET_EN bit set to 1 to update the IINDPM register. When the device is in default mode, the PSEL value updates IINDPM in real time.

表 7-1. Input Current Limit Setting from PSEL

INPUT DETECTION	PSEL PIN	INPUT CURRENT LIMIT (ILIM)	VBUS_STAT
USB SDP	HIGH	500 mA	001
Adapter	LOW	2.4 A	011

7.3.3.4 Input Voltage Limit Threshold Setting (VINDPM Threshold)

The device has two modes to set the VINDPM threshold.

- Fixed VINDPM threshold. VINDPM is in default set at 4.5 V (programmable from 3.9 V to 5.4 V).
- VINDPM threshold tracks the battery voltage to optimize the converter headroom between input and output. When it is enabled in REG07[1:0], the actual input voltage limit is the higher of the VINDPM setting in register and $V_{\text{BAT}} + \text{offset voltage in VINDPM_BAT_TRACK}[1:0]$.

7.3.3.5 Power Up Converter in Buck Mode

After the input current limit is set, the converter is enabled and the HSFET and LSFET start switching. The system voltage is powered from the converter instead of the battery. If battery charging is disabled, the BATFET turns off. Otherwise, the BATFET stays on to charge the battery.

The device provides soft start when the system rail is ramping up. When the system rail is below $V_{\text{BAT_SHORT}}$, the input current is limited to the lower of 200 mA or IINDPM register setting. The system load should be appropriately planned not to exceed the 200-mA IINDPM limit. After the system rises above $V_{\text{BAT_SHORTZ}}$, the device input current limit is the value set by the IINDPM register.

As a battery charger, the device deploys a highly efficient 1.5-MHz step-down switching regulator. The fixed frequency oscillator keeps tight control of the switching frequency under all conditions of input voltage, battery voltage, charge current, and temperature simplifying output filter design.

The converter supports PFM operation by default for fast transient response during system voltage regulation and better light load efficiency. The PFM_DIS bit disables PFM operation if system voltage is not in regulation.

7.3.3.6 HIZ Mode with Adapter Present

By setting the EN_HIZ bit to 1 with adapter, the device enters a high impedance state (HIZ). In HIZ mode, the system is powered from the battery even with good adapter present. The device is in the low input quiescent current state with Q1 RBFET, REGN LDO, and the bias circuits off.

7.3.4 Boost Mode Operation From Battery

The device supports boost converter operation to deliver power from the battery to other portable devices through a USB port. The output voltage is regulated at 5 V (programmable 4.6/4.75/5.0/5.15 V) and output current is up to 1 A. The user needs to have at least 350 mV between V_{BAT} and Boost mode regulation voltage (V_{BST}) to power up Boost mode reliably. For example, the BOOSTV[1:0] setting is recommended to be 4.75 V or higher if the battery voltage is 4.4 V.

Boost operation is enabled if the conditions below are valid:

1. Register setting: BATFET_DIS = 0, CHG_CONFIG = 0 and BST_CONFIG = 1
2. BAT above V_{BST_BAT} set by MIN_VBAT_SEL bit,
3. VBUS less than $V_{BAT} + V_{SLEEP}$ (in sleep mode) before converter starts.
4. Voltage at TS (thermistor) pin, as a percentage of V_{REGN} , is within acceptable range ($V_{BHOT_RISE\%} < V_{TS\%} < V_{BCOLD_FALL\%}$)

During Boost mode, the status register VBUS_STAT bits are set to 111.

The converter supports PFM operation at light load in Boost mode. The PFM_DIS bit can be used to disable PFM operation in boost configuration.

The BQ25619/618 keeps the Q1 FET off during Boost mode. During adapter plug-in or removal, the charger automatically transitions between charging mode and Boost mode by setting the BST_CONFIG bit and CHG_CONFIG bit both to 1. When the adapter plugs in and the conditions to start a new charge cycle are valid, the device is in charging mode. If the adapter is removed and the boost enable conditions are valid, the device transits to Boost mode to power the accessories connected to PMID automatically.

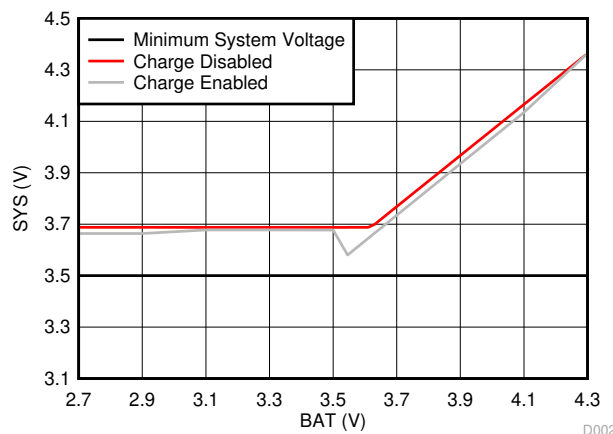
7.3.5 Power Path Management

The device accommodates a wide range of input sources such as USB, wall adapter, or car charger. The device provides automatic power path selection to supply the system (SYS) from the input source (VBUS), battery (BAT), or both.

7.3.5.1 Narrow VDC Architecture

When the battery is below the minimum system voltage setting, the BATFET operates in linear mode (LDO mode), and the system is typically 180 mV above the minimum system voltage setting. As the battery voltage rises above the minimum system voltage, the BATFET is fully on and the voltage difference between the system and battery is the V_{DS} of the BATFET.

When battery charging is disabled and above the minimum system voltage setting or charging is terminated, the system is always regulated at typically 50 mV above the battery voltage. The status register VSYS_STAT bit goes to 1 when the system is in minimum system voltage regulation.



7-1. System Voltage vs Battery Voltage

7.3.5.2 Dynamic Power Management

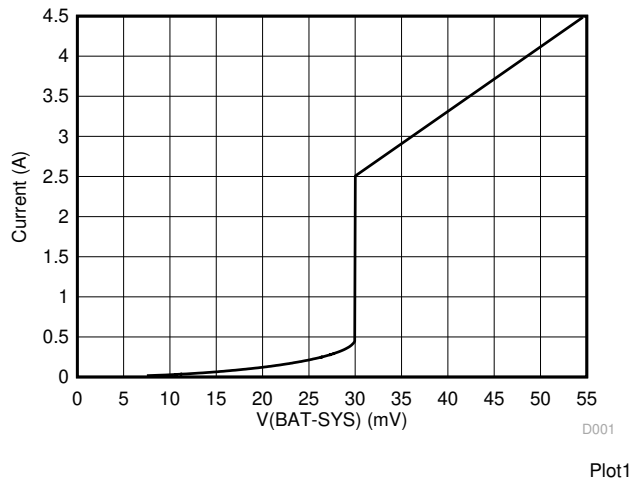
To meet the maximum current limit in the USB specification and avoid overloading the adapter, the device features Dynamic Power Management (DPM), which continuously monitors the input current and input voltage. When input source is overloaded, either the current exceeds the input current limit (IINDPM) or the voltage falls below the input voltage limit (VINDPM). The device then reduces the charge current until the input current falls below the input current limit or the input voltage rises above the input voltage limit.

When the charge current is reduced to zero, but the input source is still overloaded, the system voltage starts to drop. Once the system voltage falls below the battery voltage, the device automatically enters the supplement mode where the BATFET turns on and the battery starts discharging so that the system is supported from both the input source and battery.

During DPM mode, the status register bits VINDPM_STAT or IINDPM_STAT go to 1.

7.3.5.3 Supplement Mode

When the system voltage falls below the battery voltage, the BATFET turns on and the BATFET gate is regulated so that the minimum BATFET V_{DS} stays at 30 mV when the current is low. This prevents oscillation from entering and exiting the supplement mode.

As the discharge current increases, the BATFET gate is regulated with a higher voltage to reduce $R_{DS(on)}$ until the BATFET is in full conduction. At this point onwards, the BATFET V_{DS} linearly increases with discharge current.  shows the V-I curve of the BATFET gate regulation operation. The BATFET turns off to exit supplement mode when the battery is below battery depletion threshold.

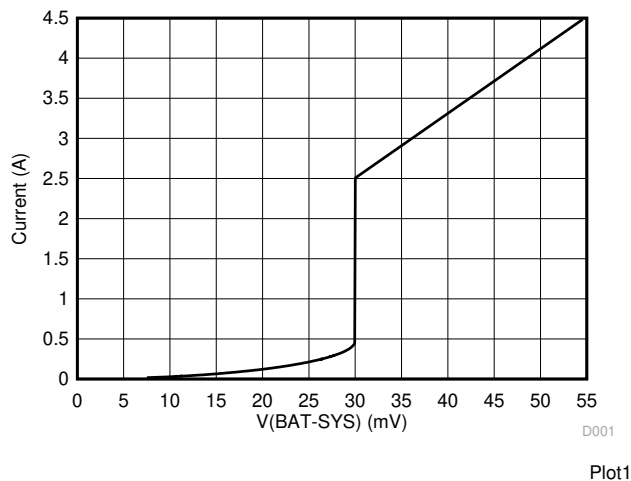


图 7-2. BATFET V-I Curve

7.3.6 Battery Charging Management

The device charges a 1-cell Li-ion battery with up to 1.5-A charge current for a high capacity tablet battery. The 19.5-m Ω BATFET improves charging efficiency and minimizes the voltage drop during discharging.

7.3.6.1 Autonomous Charging Cycle

When battery charging is enabled (CHG_CONFIG bit = 1 and $\overline{CE}$ pin is LOW), the device autonomously completes a charging cycle without host involvement. The device default charging parameters are listed in [表 7-2](#). The host configures the power path and charging parameters by writing to the corresponding registers through I²C.

表 7-2. Charging Parameter Default Settings

DEFAULT MODE	BQ25619/618
Charging voltage	4.20 V
Charging current	340 mA

表 7-2. Charging Parameter Default Settings (続き)

DEFAULT MODE	BQ25619/618
Pre-charge current	40 mA
Termination current	60 mA
Temperature profile	JEITA
Safety timer	10 hours

A new charge cycle starts when the following conditions are valid:

- Converter starts
- Battery charging is enabled (CHG_CONFIG bit = 1 and I_{CHG} register is not 0 mA and $\overline{CE}$ is low)
- No thermistor fault on TS. (TS pin can be ignored by setting TS_IGNORE bit to 1)
- No safety timer fault
- BATFET is not forced to turn off (BATFET_DIS bit = 0)

The device automatically terminates the charging cycle when the charging current is below the termination threshold, the battery voltage is above the recharge threshold, and the device is not in DPM mode or thermal regulation. When a fully charged battery is discharged below recharge threshold (selectable through VRECHG bit), the device automatically starts a new charging cycle. After the charge is done, a toggle of the $\overline{CE}$ pin or CHG_CONFIG bit initiates a new charging cycle. Adapter removal and replug will also restart a charging cycle.

The STAT output indicates charging status: charging (LOW), charging complete or charge disable (HIGH), or charging fault (blinking). The status register (CHRG_STAT) indicates the different charging phases: 00-charging disable, 01-precharge, 10-fast charge (CC) and constant voltage (CV), 11-charging done. Once a charging cycle is completed, an $\overline{INT}$ pulse is asserted to notify the host.

7.3.6.2 Battery Charging Profile

The device charges the battery in five phases: battery short, preconditioning, constant current, constant voltage, and top-off trickle charging (optional). At the beginning of a charging cycle, the device checks the battery voltage and regulates current and voltage accordingly.

Resistance between charger output and battery cell terminal such as board routing, connector, MOSFETs, and sense resistor can force the charging process to move from constant current to constant voltage too early and increase charge time. To speed up the charging cycle, the device provides the BATSNS pin to extend the constant current charge time to deliver maximum power to battery. The BATSNS pin is connected directly to the battery cell terminal to remotely sense battery cell voltage. BATSNS is by default enabled and can be disabled through the BATSNS_DIS bit. If BATSNS is connected to GND or left floating, the charger regulates the BAT pin instead.

表 7-3. Charging Current Setting

V_{BAT}	CHARGING CURRENT	DEFAULT SETTING	CHRG_STAT
< 2.2 V	I_{BAT_SHORT}	25 mA	01
2.2 V to 3 V	I_{PRECHG}	40 mA	01
> 3 V	I_{CHG}	340 mA	10

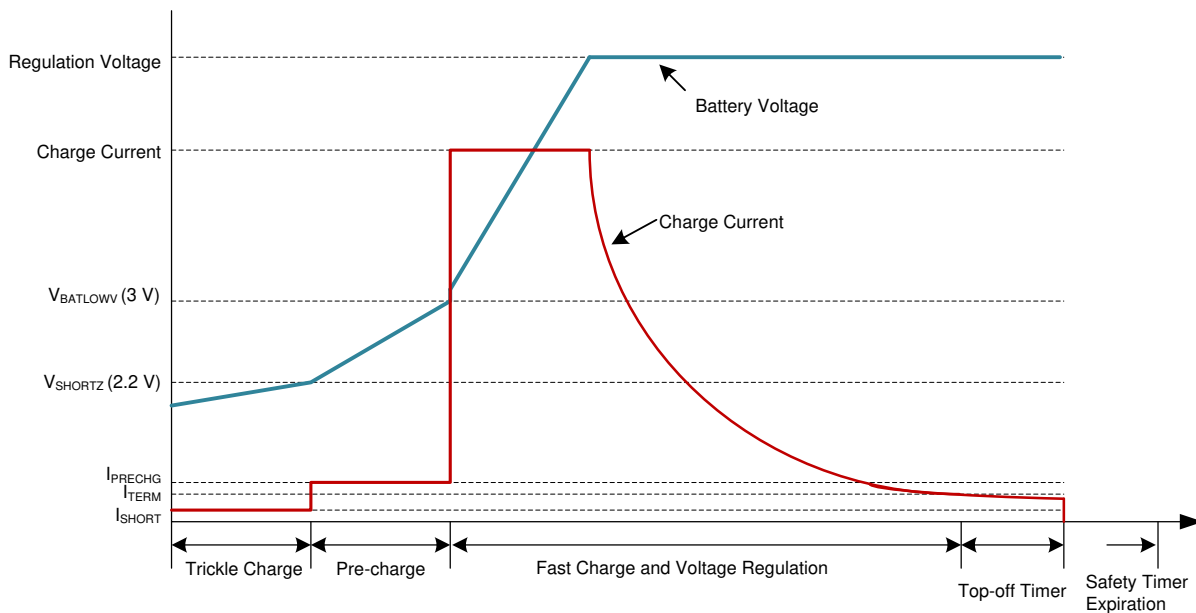


図 7-3. Battery Charging Profile

7.3.6.3 Charging Termination

The device terminates a charge cycle when the battery voltage is above the recharge threshold, and the current is below termination current. After the charging cycle has completed, the BATFET turns off. STAT is asserted HIGH to indicate charging is done. The converter keeps running to power the system, and BATFET can turn on again to engage セクション 7.3.5.3.

If the device is in IINDPM/VINDPM regulation, or thermal regulation, the actual charging current will be less than the termination value. In this case, termination is temporarily disabled.

When termination occurs, the STAT pin goes HIGH. The status register CHRG_STAT is set to 11, and an $\overline{\text{INT}}$ pulse is asserted to the host. Termination can be disabled by writing 0 to the EN_TERM bit prior to charge termination.

Termination current is set in REG03[3:0]. For a small capacity battery, the termination current can be set as low as 20 mA for full charge. Due to the termination current accuracy, the actual termination current may be higher than the termination target. In order to compensate for termination accuracy, a programmable top-off timer can be applied after termination is detected. The top-off timer will follow safety timer constraints, such that if the safety timer is suspended, so will the top-off timer. Similarly, if the safety timer is doubled, so will the termination top-off timer. The TOPOFF_ACTIVE bit reports whether the top-off timer is active or not. The host can read CHRG_STAT and TOPOFF_ACTIVE to find out the termination status. The STAT pin stays HIGH during a top-off timer counting cycle.

Top-off timer settings are read in once termination is detected by the charger. Programming a top-off timer value (01, 10, 11) after termination has no effect unless a recharge cycle is initiated. The top-off timer immediately stops if it is disabled (00). An $\overline{\text{INT}}$ is asserted to the host when entering a top-off timer segment as well as when the top-off timer expires.

7.3.6.4 Thermistor Qualification

The device provides a single thermistor input for battery temperature monitoring.

7.3.6.4.1 JEITA Guideline Compliance During Charging Mode

To improve the safety of charging Li-ion batteries, the JEITA guideline was released on April 20, 2007. The guideline emphasized the importance of avoiding a high charge current and high charge voltage at certain low and high temperature ranges.

To initiate a charge cycle, the voltage on TS pin, as a percentage of V_{REGN} , must be within the $V_{T1_FALL\%}$ to $V_{T5_RISE\%}$ thresholds. If the TS voltage percentage exceeds the T1-T5 range, the controller suspends charging, a TS fault is reported and waits until the battery temperature is within the T1-T5 range.

At cool temperature (T1-T2), the charge current is reduced to a programmable fast charge current (0%, 20% default, 50%, 100% of I_{CHG} , by JEITA_ISET). At warm temperature (T3-T5), the charge voltage is reduced to 4.1 V or kept at V_{REG} (JEITA_VSET), and the charge current can be reduced to a programmable level (0%, 20%, 50%, 100% default). Battery termination is disabled in T3-T5. The charger provides more flexible settings on a T2 and T3 threshold as well to program the temperature profile beyond JEITA. When T1 is set to 0°C and T5 is set to 60°C, T2 can be programmed to 5.5°C/10°C (default)/15°C/20°C, and T3 can be programmed to 40°C/45.5°C (default)/50.5°C/54.5°C.

When the charger does not need to monitor the NTC, the host sets the TS_IGNORE bit to 1 to ignore the TS pin condition during charging and Boost mode. If the TS_IGNORE bit is set to 1, the TS pin is ignored and the charger ignores the TS pin input. In this case, the NTC_FAULT bits are 000 to report normal TS status.

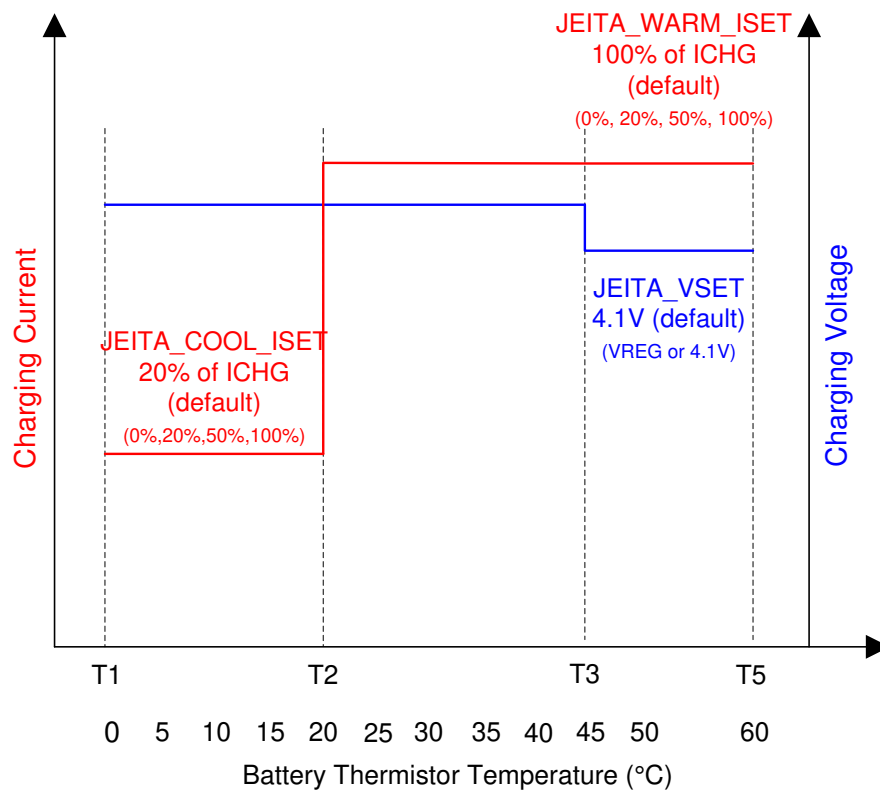


図 7-4. JEITA Profile

式 1 through 式 2 describe how to calculate resistor divider values on the TS pin.

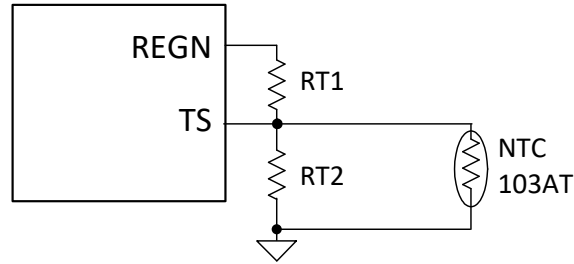


図 7-5. TS Pin Resistor Network

$$RT1 = \frac{\frac{1}{V_{T1}\%} - 1}{\frac{1}{R_{T2}} + \frac{1}{R_{NTC,T1}}} \quad (1)$$

$$RT2 = \frac{R_{NTC,T1} \times R_{NTC,T5} \times \left(\frac{1}{V_{T5}\%} - \frac{1}{V_{T1}\%} \right)}{R_{NTC,T1} \times \left(\frac{1}{V_{T1}\%} - 1 \right) - R_{NTC,T5} \times \left(\frac{1}{V_{T5}\%} - 1 \right)} \quad (2)$$

In the equations above, $R_{NTC,T1}$ is the NTC thermistor resistance value at temperature $T1$ and $R_{NTC,T5}$ is the NTC thermistor resistance value at temperature $T5$. Selecting a 0°C to 60°C range for a Li-ion or Li-polymer battery then:

- $R_{NTC,T1} = 27.28 \text{ k}\Omega$ (0°C)
- $R_{NTC,T5} = 3.02 \text{ k}\Omega$ (60°C)
- $RT1 = 5.3 \text{ k}\Omega$
- $RT2 = 31.14 \text{ k}\Omega$

7.3.6.4.2 Boost Mode Thermistor Monitor During Battery Discharge Mode

For battery protection during Boost mode, the device monitors battery temperature to be within the V_{BCOLD} and V_{BHOT} thresholds. When $RT1$ is $5.3 \text{ k}\Omega$ and $RT2$ is $31.14 \text{ k}\Omega$, T_{BCOLD} default is -19.5°C and T_{BHOT} default is 64°C . When the temperature is outside of the temperature thresholds, Boost mode is suspended. In addition, the $VBUS_STAT$ bits are set to 000 and NTC_FAULT is reported. Once the temperature returns within the thresholds, Boost mode is recovered and NTC_FAULT is cleared.

7.3.6.5 Charging Safety Timer

The device has a built-in safety timer to prevent an extended charging cycle due to abnormal battery conditions. The safety timer is 2 hours when the battery is below the $V_{BATLOWV}$ threshold and 10 hours (10/20 hours in $REG05[2]$) when the battery is higher than the $V_{BATLOWV}$ threshold. When the safety timer expires, the $STAT$ pin is blinking at 1 Hz to report a safety timer expiration fault.

The user can program the fast charge safety timer through I²C (CHG_TIMER bit $REG05[2]$). When the safety timer expires, the fault register $CHRG_FAULT$ bits ($REG09[5:4]$) are set to 11 and an $\overline{INT}$ is asserted to the host. The safety timer (both fast charge and precharge) can be disabled through I²C by setting the EN_TIMER bit.

During IINDPM/VINDPM regulation, thermal regulation, or JEITA cool/warm when fast charge current is reduced, the safety timer counts at a half clock rate, because the actual charge current is likely below the setting. For example, if the charger is in input current regulation ($IINDPM_STAT = 1$) throughout the whole charging cycle, and the safety time is set to 10 hours, the safety timer will expire in 20 hours. This half clock rate feature can be disabled by writing 0 to the $TMR2X_EN$ bit.

During faults of BAT_FAULT, NTC_FAULT that lead to charging suspend, the safety timer is suspended as well. Once the fault goes away, the timer resumes. If the user stops the current charging cycle, and starts it again, the timer gets reset (toggle of CE pin or CHG_CONFIG bit).

7.3.7 Ship Mode and QON Pin

7.3.7.1 BATFET Disable (Enter Ship Mode)

To extend battery life and minimize power when the system is powered off during system idle, shipping, or storage, the device turns off BATFET so that the system voltage is floating to minimize the battery leakage current. When the host sets the BATFET_DIS bit, the charger can turn off the BATFET immediately or delay by t_{BATFET_DLY} as configured by the BATFET_DLY bit. To set the device into ship mode with the adapter present, the host has to first set BATFET_RST_VBUS to 1 and then BATFET_DIS to 1. The charger will turn off the BATFET (no charging, no supplement) while the adapter is still attached. When the adapter is removed, the charger will enter ship mode.

7.3.7.2 BATFET Enable (Exit Ship Mode)

When the BATFET is disabled (in ship mode) as indicated by setting BATFET_DIS, one of the following events can enable the BATFET to restore system power:

1. Plug in adapter
2. Clear BATFET_DIS bit
3. Set REG_RST bit to reset all registers including BATFET_DIS bit to default (0)
4. A logic high to low transition on QON pin with $t_{SHIPMODE}$ deglitch time to enable BATFET to exit ship mode. EN_HIZ bit is set to 1 (regardless of adapter present or not). Host has to set EN_HIZ bit to 0 before boost mode enable. Once adapter plugs in, EN_HIZ will be cleared.

7.3.7.3 BATFET Full System Reset

The BATFET functions as a load switch between the battery and system when an input source is not plugged in. When BATFET_RST_EN = 1 and BATFET_DIS = 0, the BATFET full system reset function is enabled. By changing the state of BATFET from on to off, systems connected to SYS can be effectively forced to have a power-on-reset. After the reset is complete, the device is in the POR state, and all registers are in POR default settings. The QON pin supports a push-button interface to reset system power without the host by changing the state of BATFET. Internally, it is pulled up to the V_{QON} voltage through a 200-k Ω resistor.

When the QON pin is driven to logic low for t_{QON_RST} , the BATFET reset process starts. The BATFET is turned off for t_{BATFET_RST} and then it is re-enabled to reset system power. This function can be disabled by setting the BATFET_RST_EN bit to 0.

The BATFET full system reset functions either with or without an adapter present. If BATFET_RST_WVBUS = 1, the system reset function starts after t_{QON_RST} when the QON pin is pushed to LOW. Once the reset process starts, the device first goes into HIZ mode to turn off the converter, and then power cycles BATFET. If BATFET_RST_WVBUS = 0, the system reset function does not start until t_{QON_RST} after the QON pin is pushed to LOW and the adapter is removed.

After the BATFET full system reset is complete, the device powers up again if EN_HIZ is not set to 1 before the system reset.

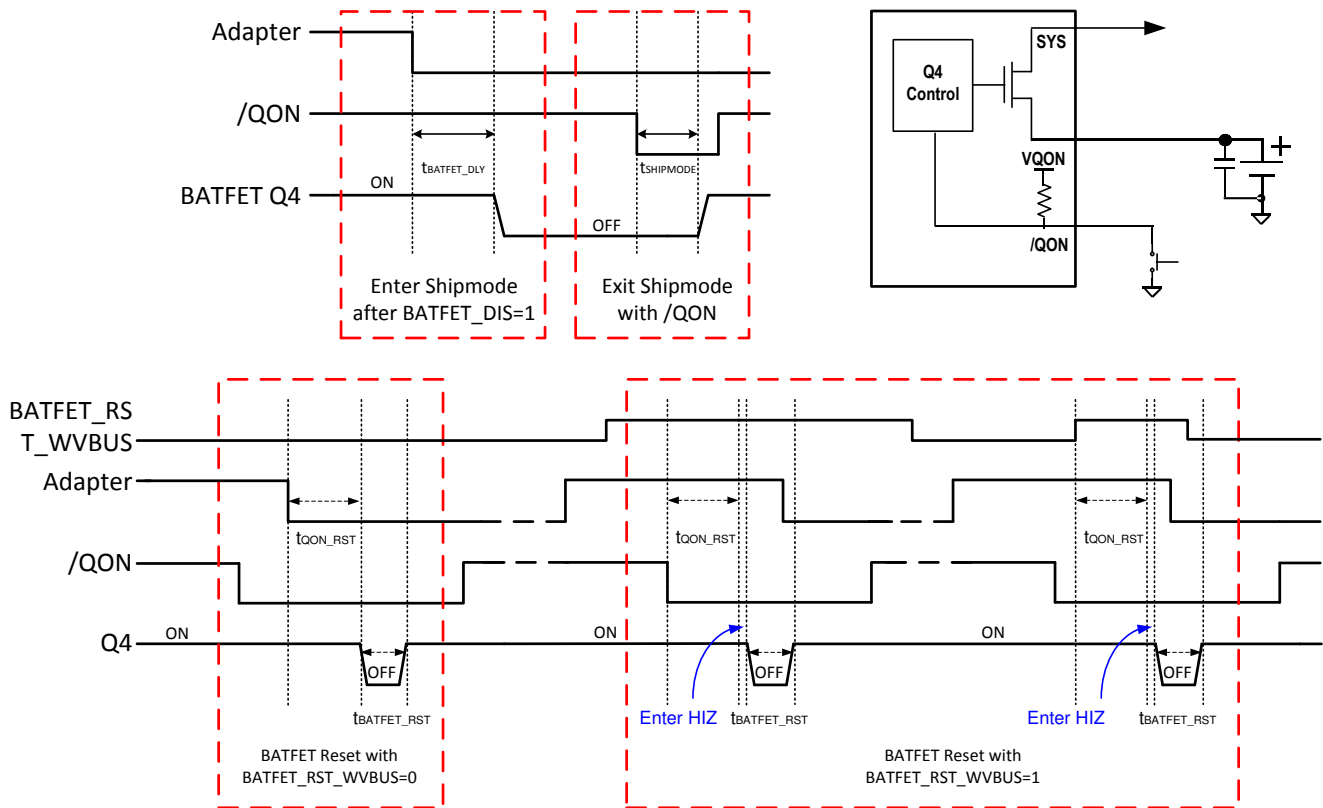


図 7-6. QON Timing

7.3.8 Status Outputs (STAT, $\overline{INT}$, PMID_GOOD)

7.3.8.1 Power Good Indicator (PG_STAT Bit)

The PG_STAT bit goes to 1 to indicate a good input source when:

- V_{VBUS} above V_{VBUS_UVLO}
- V_{VBUS} above battery (not in sleep)
- V_{VBUS} below V_{ACOV} threshold
- V_{VBUS} above $V_{POORSRC}$ (typical 3.8 V) when I_{BADSRC} (typical 30 mA) current is applied (not a poor source)
- Completed [セクション 7.3.3.3](#)

7.3.8.2 Charging Status Indicator (STAT)

The device indicates the charging state on the open drain STAT pin. The STAT pin can drive an LED.

表 7-4. STAT Pin State

CHARGING STATE	STAT INDICATOR
Charging in progress (including recharge)	LOW
Charging termination (top off timer may be running)	HIGH
Sleep mode, charge disable, Boost mode	HIGH
Charge suspend (input overvoltage, TS fault, safety timer fault, or system overvoltage)	Blinking at 1 Hz

7.3.8.3 Interrupt to Host ($\overline{INT}$)

In some applications, the host does not always monitor charger operation. The $\overline{INT}$ pulse notifies the host on device operation. The following events generate a 256- μ s $\overline{INT}$ pulse.

- Good input source detected:
 - V_{VBUS} above battery (not in sleep)

- V_{VBUS} below V_{ACOV} threshold
- V_{VBUS} above $V_{POORSRC}$ (typical 3.8 V) when I_{BADSRC} (typical 30 mA) current is applied (not a poor source)
- Input adapter removed
- USB/adapter source identified during セクション 7.3.3.3.
- Charge complete
- Any FAULT event in REG09
- VINDPM / IINDPM event detected (REG0A[1:0], maskable)
- Top-off timer starts and expires

REG09[7:0] and REG0A[6:4] report charger operation faults and status change to the host. When a fault/status change occurs, the charger sends out an $\overline{INT}$ pulse and keeps the state in REG09[7:0]/REG0A[6:4] until the host reads the registers. Before the host reads REG09[7:0]/REG0A[6:4] and all the ones are cleared, the charger does not send any $\overline{INT}$ upon new fault/status change. To read the current status, the host has to read REG09/REG0A two times consecutively. The first read reports the pre-existing register status and the second read reports the current register status.

7.3.8.4 PMID Voltage Indicator (PMID_GOOD)

In the BQ25619/618, the accessory devices can be connected to the charger PMID pin to get power either from the adapter through the Q1 direct path or from battery Boost mode. An optional external PMOS FET can be placed between the charger PMID pin and accessory input to disconnect the power path during overcurrent and overvoltage conditions. PMID_GOOD is used to drive an external PMOS FET through an inverter. PMID_GOOD HIGH turns on an inverter to pull the PMOS FET gate low to turn on the PMOS FET, and PMID_GOOD LOW turns off the PMOS FET.

Upon adapter plug-in, PMID_GOOD goes from LOW to HIGH when VBUS rises above the battery but below V_{ACOV} , and passes poor source detection. During the operation, PMID_GOOD goes from HIGH to LOW if Q1 current exceeds 115% of the IINDPM threshold, (I_{BLK_OCP}), or adapter voltage rises above 5.8 V (V_{BST_OVP}).

The high-voltage adapter over V_{BST_OVP} keeps charging the battery if all conditions are valid. The external PMOS FET stays off to protect the accessory from an overvoltage fault.

When the adapter is removed, PMID_GOOD goes LOW before battery Boost mode starts.

In battery Boost mode, the device regulates PMID voltage between 4.6 V to 5.15 V as a stable power supply to the accessory devices. PMID_GOOD goes from LOW to HIGH when PMID voltage rises above 3.8 V ($V_{POORSRC}$). Similar to the adapter present scenario, the PMID valid voltage range is between $V_{POORSRC}$ and V_{BST_OVP} . Once PMID voltage is out of this range, PMID_GOOD goes LOW to disconnect the accessory device from PMID. During Boost mode, all of the conditions to exit Boost mode will drive PMID_GOOD from HIGH to LOW, including Boost mode disable in register, ACOV, TS fault, battery depleted (V_{BAT_DPL}), BATFET overcurrent, ($I_{SYS_OCP_Q4}$), etc.

7.3.9 Protections

7.3.9.1 Voltage and Current Monitoring in Buck Mode

7.3.9.1.1 Input Overvoltage Protection (ACOV)

The input voltage is sensed via the VAC pin. The default OVP threshold is 14.2 V, and can be programmed at 5.7 V/6.4 V/11 V/14.2 V via OVP[1:0] register bits. ACOV event immediately stops converter switching whether in buck or Boost mode. The device automatically resumes normal operation once the input voltage drops back below the OVP threshold. During ACOV, REGN LDO is on, and the device does not enter HIZ mode.

During ACOV, the fault register CHRG_FAULT bits are set to 01. An $\overline{INT}$ pulse is asserted to the host.

7.3.9.1.2 System Overvoltage Protection (SYSOVP)

The charger device clamps the system voltage during a load transient so that the components connected to the system are not damaged due to high voltage. The V_{SYS_OVP} threshold is about 300 mV above battery regulation voltage when battery charging is terminated. Upon SYSOVP, the converter stops switching immediately to clamp the overshoot. The charger pulls 30-mA I_{SYS_LOAD} discharge current to bring down the system voltage.

7.3.9.2 Voltage and Current Monitoring in Boost Mode

7.3.9.2.1 Boost Mode Overvoltage Protection

When PMID voltage rises above the regulation target and exceeds V_{BST_OVP} , the device stops switching immediately and the device exits Boost mode and PMID_GOOD is pulled low as well. The BST_CONFIG bit is set to 0. During Boost mode overvoltage, the fault register bit BOOST_FAULT is set to 1 to indicate a fault in boost operation. An $\overline{INT}$ is asserted to the host.

7.3.9.2.2 PMID Overcurrent Protection

The BQ25619/618 closely monitors the battery discharge current through BATFET (Q4) to ensure safe Boost mode operation. During an overcurrent condition when boost input current exceeds $I_{SYS_OCP_Q4}$, the device latches off in 100 μ s. When an overcurrent condition is detected, the fault register bit BOOST_FAULT is set high to indicate a fault in boost operation. An $\overline{INT}$ is asserted to the host.

7.3.9.3 Thermal Regulation and Thermal Shutdown

7.3.9.3.1 Thermal Protection in Buck Mode

Besides the battery temperature monitor on the TS pin, the device monitors the internal junction temperature T_J to avoid overheating the chip and limits the IC junction temperature in buck mode. When the internal junction temperature exceeds the thermal regulation limit (110°C), the device lowers down the charge current. During thermal regulation, the actual charging current is usually below the programmed battery charging current. Therefore, termination is disabled, the safety timer runs at half the clock rate, and the status register THERM_STAT bit goes high.

Additionally, the device has thermal shutdown to turn off the converter and the BATFET when the IC surface temperature exceeds T_{SHUT} 150°C. The BATFET and converter are enabled to recover when IC temperature is 130°C. The fault register CHRГ_FAULT is set to 10 during thermal shutdown and an $\overline{INT}$ is asserted to the host.

7.3.9.3.2 Thermal Protection in Boost Mode

Besides the battery temperature monitor on the TS pin, the device monitors the internal junction temperature to provide thermal shutdown during Boost mode. When the IC junction temperature exceeds T_{SHUT} 150°C, Boost mode is disabled by setting the BST_CONFIG bit low. When the IC junction temperature is below 145°C, the host can re-enable Boost mode.

7.3.9.4 Battery Protection

7.3.9.4.1 Battery Overvoltage Protection (BATOVP)

The battery overvoltage limit is clamped at 4% above battery regulation voltage. When battery overvoltage occurs, the charger device immediately stops switching. The fault register BAT_FAULT bit goes high and an $\overline{INT}$ is asserted to the host.

7.3.9.4.2 Battery Overdischarge Protection

When the battery is discharged below $V_{BAT_DPL_FALL}$, the BATFET latches off to protect the battery from overdischarge. To recover from overdischarge latch-off, an input source plug-in is required at VAC/VBUS.

7.3.9.4.3 System Overcurrent Protection

$I_{SYS_OCP_Q4}$ sets the battery discharge current limit. Once $I_{BAT} > I_{SYS_OCP_Q4}$, the charger latches off Q4 and puts the device into Ship mode. All methods to exit Ship mode are valid to bring the part out of Q4 latch-off.

7.3.10 Serial Interface

The device uses an I²C compatible interface for flexible charging parameter programming and instantaneous device status reporting. I²C™ is a bi-directional 2-wire serial interface developed by Philips Semiconductor (now NXP Semiconductors). Only two bus lines are required: a serial data line (SDA) and a serial clock line (SCL). Devices can be considered as masters or slaves when performing data transfers. A master is the device which initiates a data transfer on the bus and generates the clock signals to permit that transfer. At that time, any device addressed is considered a slave.

The device operates as a slave device with address 6AH, receiving control inputs from the master device like a microcontroller or a digital signal processor through REG00 to REG0C. A register read beyond REG0C returns 0xFF. The I²C interface supports both standard mode (up to 100 kbits), and fast mode (up to 400 kbits), connecting to the positive supply voltage via a current source or pullup resistor. When the bus is free, both lines are HIGH. The SDA and SCL pins are open drain.

7.3.10.1 Data Validity

The data on the SDA line must be stable during the HIGH period of the clock. The HIGH or LOW state of the data line can only change when the clock signal on the SCL line is LOW. One clock pulse is generated for each data bit transferred.

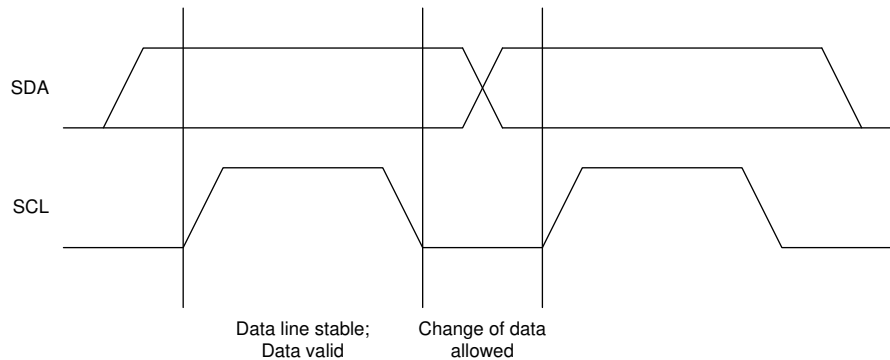


図 7-7. Bit Transfer on the I²C Bus

7.3.10.2 START and STOP Conditions

All transactions begin with a START (S) and can be terminated by a STOP (P). A HIGH to LOW transition on the SDA line while SCL is HIGH defines a START condition. A LOW to HIGH transition on the SDA line when the SCL is HIGH defines a STOP condition. START and STOP conditions are always generated by the master. The bus is considered busy after the START condition, and free after the STOP condition.

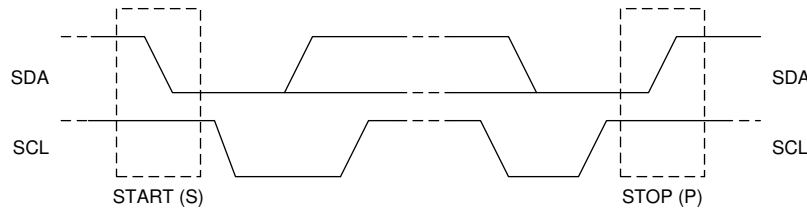


図 7-8. TS START and STOP conditions

7.3.10.3 Byte Format

Every byte on the SDA line must be 8 bits long. The number of bytes to be transmitted per transfer is unrestricted. Each byte has to be followed by an Acknowledge bit. Data is transferred with the Most Significant Bit (MSB) first. If a slave cannot receive or transmit another complete byte of data until it has performed some other function, it can hold the clock line SCL low to force the master into a wait state (clock stretching). Data transfer then continues when the slave is ready for another byte of data and release the clock line SCL.

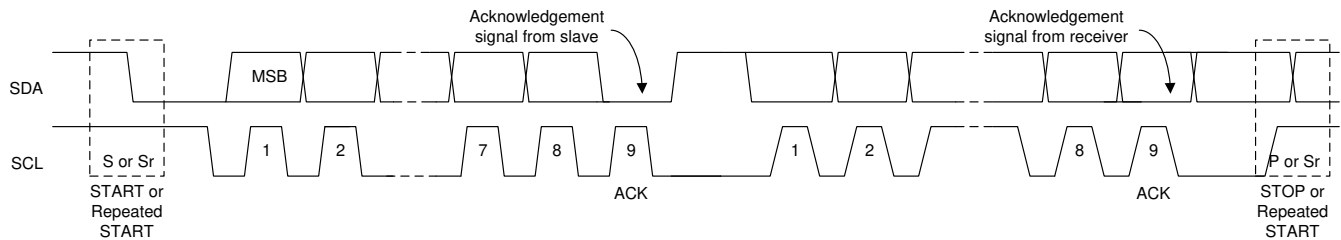


図 7-9. Data Transfer on the I²C Bus

7.3.10.4 Acknowledge (ACK) and Not Acknowledge (NACK)

The acknowledge takes place after every byte. The acknowledge bit allows the receiver to signal the transmitter that the byte was successfully received and another byte may be sent. All clock pulses, including the acknowledge ninth clock pulse, are generated by the master. The transmitter releases the SDA line during the acknowledge clock pulse so the receiver can pull the SDA line LOW and it remains stable LOW during the HIGH period of this clock pulse.

When SDA remains HIGH during the ninth clock pulse, this is the Not Acknowledge signal. The master can then generate either a STOP to abort the transfer or a repeated START to start a new transfer.

7.3.10.5 Slave Address and Data Direction Bit

After the START, a slave address is sent. This address is 7 bits long followed by the eighth bit as a data direction bit (bit R/W). A zero indicates a transmission (WRITE) and a one indicates a request for data (READ).

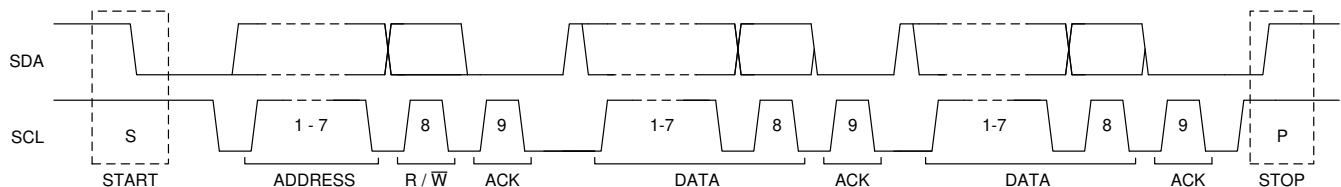


図 7-10. Complete Data Transfer

7.3.10.6 Single Read and Write

If the register address is not defined, the charger IC sends back NACK and goes back to the idle state.

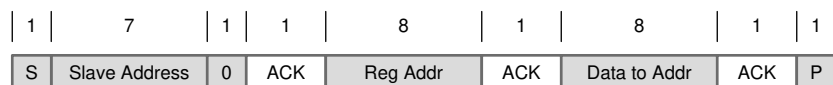


図 7-11. Single Write

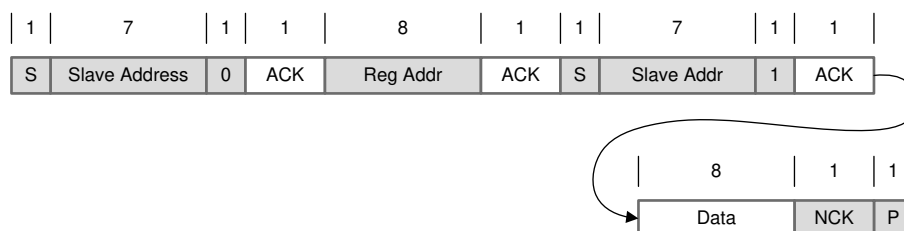


図 7-12. Single Read

7.3.10.7 Multi-Read and Multi-Write

The charger device supports multi-read and multi-write on REG00 through REG0C.

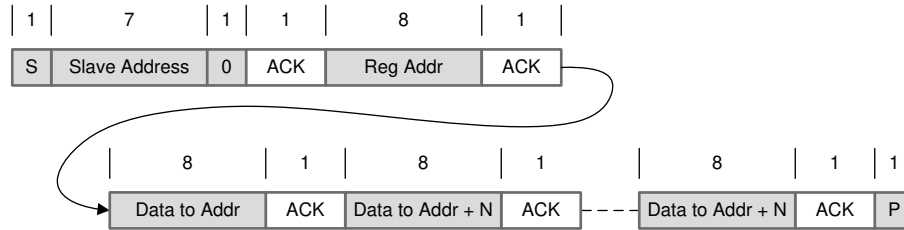


図 7-13. Multi-Write

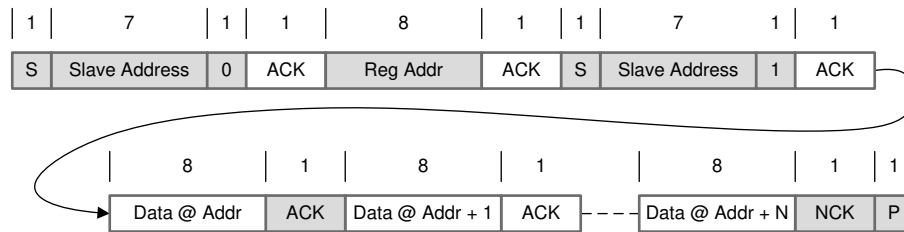


図 7-14. Multi-Read

REG09[7:0]/REG0A[6:4] are fault/status change registers. They keep all of the fault/status information from the last read until the host issues a new read. For example, if a Charge Safety Timer Expiration fault occurs but recovers later, the fault register REG09 reports the fault when it is read the first time, but returns to normal when it is read the second time. In order to get the fault information at present, the host has to read REG09/REG0A for the second time.

7.4 Device Functional Modes

7.4.1 Host Mode and Default Mode

The device is a host controlled charger, but it can operate in default mode without host management. In default mode, the device can be used as an autonomous charger with no host or while the host is in sleep mode. When the charger is in default mode, the WATCHDOG_FAULT bit is HIGH. When the charger is in host mode, the WATCHDOG_FAULT bit is LOW.

After power-on-reset, the device starts in default mode with watchdog timer expired, or default mode. All registers are in the default settings.

In default mode, the device keeps charging the battery with the 10-hour fast charging safety timer. At the end of the 10-hour, charging is stopped and the buck converter continues to operate to supply system load. Any write command to the device transitions the charger from default mode to host mode. All device parameters can be programmed by the host. To keep the device in host mode, the host has to reset the watchdog timer by writing a 1 to the WD_RST bit before the watchdog timer expires (WATCHDOG_FAULT bit is set), or disable the watchdog timer by setting the WATCHDOG bits = 00.

All device parameters can be programmed by the host. To keep the device in host mode, the host has to reset the watchdog timer by writing a 1 to the WD_RST bit before the watchdog timer expires (WATCHDOG_FAULT bit is set), or disable the watchdog timer by setting the WATCHDOG bits = 00.

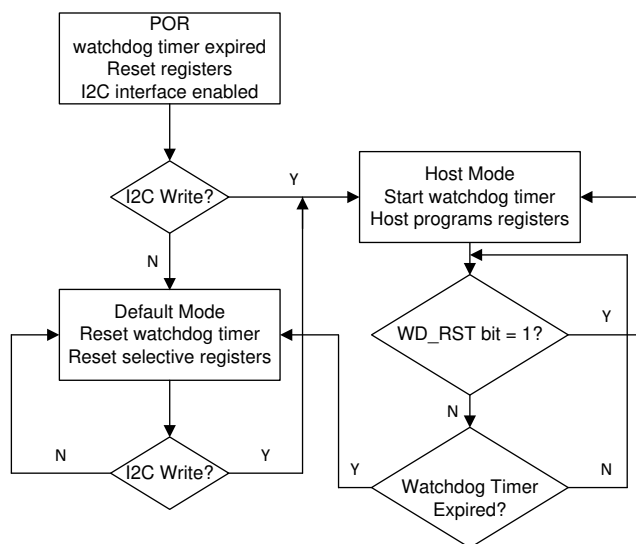


图 7-15. Watchdog Timer Flow Chart

7.5 Register Maps

I²C Slave Address: 6AH

Default I²C Slave Address: 0x6A (1101 010B + R/ $\overline{W}$)

表 7-5. I²C Registers

Address	Access Type	Acronym	Register Name	Section
00h	R/W	REG00	Input Current Limit	Go
01h	R/W	REG01	Charger Control 0	Go
02h	R/W	REG02	Charge Current Limit	Go
03h	R/W	REG03	Precharge and Termination Current Limit	Go
04h	R/W	REG04	Battery Voltage Limit	Go
05h	R/W	REG05	Charger Control 1	Go
06h	R/W	REG06	Charger Control 2	Go
07h	R/W	REG07	Charger Control 3	Go
08h	R	REG08	Charger Status 0	Go
09h	R	REG09	Charger Status 1	Go
0Ah	R	REG0A	Charger Status 2	Go
0Bh	R	REG0B	Part Information	Go
0Ch	R/W	REG0C	Charger Control 4	Go

Complex bit access types are encoded to fit into small table cells. 表 7-6 shows the codes that are used for access types in this section.

表 7-6. I²C Access Type Codes

Access Type	Code	Description
Read Type		
R	R	Read
Write Type		
W	W	Write
Reset Value		
-n		Value after reset
-X		Undefined value

7.5.1 Input Current Limit Register (Address = 00h) [Reset = 17h]

図 7-16. REG00 Register

7	6	5	4	3	2	1	0
0	0	0	1	0	1	1	1
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-7. REG00 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	EN_HIZ	0	R/W	by REG_RST by Watchdog	HIZ mode enable in buck mode. 0 – Disable (default) 1 – Enable
6	TS_IGNORE	0	R/W	by REG_RST	When charger does not monitor the NTC, host sets this bit to 1 to ignore the TS pin condition during charging and boost mode. 0 – Include TS pin into charge and boost mode enable conditions. (default) 1 – Ignore TS pin. Always consider TS is good to allow charging and boost mode. NTC_FAULT bits are 000 to report normal status.
5	BATSNS_DIS	0	R/W	by REG_RST	Select either BATSNS pin or BAT pin to regulate battery voltage. 0 – Enable BATSNS in battery CV regulation. If the device fails BATSNS open/short detection (BATSNS_STAT = 1). Battery voltage is regulated through BAT pin. (default) 1 – Disable BATSNS. Use BAT pin in battery CV regulation.
4	IINDPM[4]	1	R/W	by REG_RST	1600 mA
3	IINDPM[3]	0	R/W	by REG_RST	800 mA
2	IINDPM[2]	1	R/W	by REG_RST	400 mA
1	IINDPM[1]	1	R/W	by REG_RST	200 mA
0	IINDPM[0]	1	R/W	by REG_RST	100 mA

LEGEND: R/W = Read/Write; R = Read only

7.5.2 Charger Control 0 Register (Address = 01h) [Reset = 1Ah]

図 7-17. REG01 Register

7	6	5	4	3	2	1	0
0	0	0	1	1	0	1	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-8. REG01 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	PFM_DIS	0	R/W	by REG_RST	PFM disable in both buck and boost mode. 0 – PFM enable (default) 1 – PFM disable
6	WD_RST	0	R/W	by REG_RST by Watchdog	I ² C Watchdog timer reset. Back to 0 after watchdog timer reset 0 – Normal (default) 1 – Reset
5	BST_CONFIG	0	R/W	by REG_RST by Watchdog	Boost mode enable. In charging case application, based on adapter plug-in or removal, the charger will automatically transit between charging mode and boost mode by setting BST_CONFIG bit and CHG_CONFIG bit both to 1. 0 – Boost mode disable (default) 1 – Boost mode enable
4	CHG_CONFIG	1	R/W	by REG_RST by Watchdog	Battery charging buck mode enable. Charging is enabled when $\overline{CE}$ pin is pulled low, CHG_CONFIG bit is 1 and charge current is not zero. 0 – Charge Disable 1 – Charge Enable (default)
3	SYS_MIN[2]	1	R/W	by REG_RST	System minimum voltage setting. 000 – 2.6 V 001 – 2.8 V 010 – 3 V 011 – 3.2 V 100 – 3.4 V 101 – 3.5 V (default) 110 – 3.6 V 111 – 3.7 V
2	SYS_MIN[1]	0	R/W	by REG_RST	
1	SYS_MIN[0]	1	R/W	by REG_RST	
0	MIN_VBAT_SEL	0	R/W	by REG_RST	Minimum battery voltage when exiting boost mode. The rising threshold allows the device to start boost mode if other conditions are valid. 0 – 2.8 V V _{BAT} falling, 3 V rising (default) 1 – 2.5 V V _{BAT} falling, 2.8 V rising

LEGEND: R/W = Read/Write; R = Read only

7.5.3 Charge Current Limit Register (Address = 02h) [Reset = 91h]

☒ 7-18. REG02 Register

7	6	5	4	3	2	1	0
1	0	0	1	0	0	0	1
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-9. REG02 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	Reserved	1	R/W		
6	Q1_FULLON	0	R/W	by REG_RST	In buck mode, charger will fully turn on Q1 RBFET according to this bit setting when IINDPM is below 700 mA. When IINDPM is over 700 mA, Q1 is always fully on. 0 – Partially turn on Q1 for better regulation accuracy when IINDPM is below 700 mA. (default) 1 – Fully turn on Q1 for better efficiency when IINDPM is below 700 mA.
5	ICHG[5]	0	R/W	by REG_RST by Watchdog	640 mA
4	ICHG[4]	1	R/W	by REG_RST by Watchdog	320 mA
3	ICHG[3]	0	R/W	by REG_RST by Watchdog	160 mA
2	ICHG[2]	0	R/W	by REG_RST by Watchdog	80 mA
1	ICHG[1]	0	R/W	by REG_RST by Watchdog	40 mA
0	ICHG[0]	1	R/W	by REG_RST by Watchdog	20 mA

LEGEND: R/W = Read/Write; R = Read only

7.5.4 Precharge and Termination Current Limit Register (Address = 03h) [Reset = 12h]

図 7-19. REG03 Register

7	6	5	4	3	2	1	0
0	0	0	1	0	0	1	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-10. REG03 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	IPRECHG[3]	0	R/W	by REG_RST by Watchdog	160 mA
6	IPRECHG[2]	0	R/W	by REG_RST by Watchdog	80 mA
5	IPRECHG[1]	0	R/W	by REG_RST by Watchdog	40 mA
4	IPRECHG[0]	1	R/W	by REG_RST by Watchdog	20 mA
3	ITERM[3]	0	R/W	by REG_RST by Watchdog	160 mA
2	ITERM[2]	0	R/W	by REG_RST by Watchdog	80 mA
1	ITERM[1]	1	R/W	by REG_RST by Watchdog	40 mA
0	ITERM[0]	0	R/W	by REG_RST by Watchdog	20 mA

LEGEND: R/W = Read/Write; R = Read only

7.5.5 Battery Voltage Limit Register (Address = 04h) [Reset = 40h]

図 7-20. REG04 Register

7	6	5	4	3	2	1	0
0	1	0	0	0	0	0	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-11. REG04 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	VBATREG[4]	0	R/W	by REG_RST by Watchdog	Battery voltage setting, also called V_{REG} . Default: 4.200 V (01000)
6	VBATREG[3]	1	R/W	by REG_RST by Watchdog	00000 – 3.504 V 00001 – 3.600 V 00010 – 3.696 V
5	VBATREG[2]	0	R/W	by REG_RST by Watchdog	00011 – 3.800 V 00100 – 3.904 V
4	VBATREG[1]	0	R/W	by REG_RST by Watchdog	00101 – 4.000 V 00110 – 4.100 V 00111 – 4.150 V
3	VBATREG[0]	0	R/W	by REG_RST by Watchdog	01000 – 4.200 V 01001 - 11111 – 4.300 V - 4.520 V, 10 mV/step 01110 4.350 V, 10011 4.400 V, 11000 4.450 V, 11101 4.500 V
2	TOPOFF_TIMER[1]	0	R/W	by REG_RST by Watchdog	Top-off timer setting. 00 – Disabled (default)
1	TOPOFF_TIMER[0]	0	R/W	by REG_RST by Watchdog	01 – 15 minutes 10 – 30 minutes 11 – 45 minutes
0	VRECHG	0	R/W	by REG_RST by Watchdog	Battery recharge threshold setting. 0 – 120 mV (default) 1 – 210 mV

LEGEND: R/W = Read/Write; R = Read only

7.5.6 Charger Control 1 Register (Address = 05h) [Reset = 9Eh]

☒ 7-21. REG05 Register

7	6	5	4	3	2	1	0
1	0	0	1	1	1	1	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-12. REG05 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	EN_TERM	1	R/W	by REG_RST by Watchdog	Battery charging termination enable. 0 – Disable 1 – Enable (default)
6	Reserved	0	R/W	by REG_RST by Watchdog	Reserved
5	WATCHDOG[1]	0	R/W	by REG_RST by Watchdog	Watchdog timer setting. 00 – Disable timer 01 – 40 s (default)
4	WATCHDOG[0]	1	R/W	by REG_RST by Watchdog	10 – 80 s 11 – 160 s
3	EN_TIMER	1	R/W	by REG_RST by Watchdog	Battery charging safety timer enable, including both fast charge and precharge timers. Precharge timer is 2 hours. Fast charge timer is set by REG05[2] 0 – Disable 1 – Enable timer (default)
2	CHG_TIMER	1	R/W	by REG_RST by Watchdog	Battery fast charging safety timer setting. 0 – 20 hrs 1 – 10 hrs (default)
1	TREG	1	R/W	by REG_RST by Watchdog	Thermal regulation threshold: 0 – 90°C 1 – 110°C (default)
0	JEITA_VSET (45C-60C)	0	R/W	by REG_RST by Watchdog	Battery voltage setting during JEITA warm (T3 - T5, typically 45C - 60C) 0 – Set charge voltage to 4.1 V (max) (default) 1 – Set charge voltage to V _{REG}

LEGEND: R/W = Read/Write; R = Read only

7.5.7 Charger Control 2 Register (Address = 06h) [Reset = E6h]

図 7-22. REG06 Register

7	6	5	4	3	2	1	0
1	1	1	0	0	1	1	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-13. REG06 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	OVP[1]	1	R/W	by REG_RST	V_{ACOV} threshold during buck mode and boost mode. 00 – 5.85 V 01 – 6.4 V (5-V input) 10 – 11 V (9-V input) 11 – 14.2 V (12-V input) (default)
6	OVP[0]	1	R/W	by REG_RST	
5	BOOSTV[1]	1	R/W	by REG_RST	
4	BOOSTV[0]	0	R/W	by REG_RST	
3	VINDPM[3]	0	R/W	by REG_RST	Boost regulation voltage setting 00 – 4.6 V 01 – 4.75 V 10 – 5.0 V (default) 11 – 5.15 V
2	VINDPM[2]	1	R/W	by REG_RST	
1	VINDPM[1]	1	R/W	by REG_RST	
0	VINDPM[0]	0	R/W	by REG_RST	
					VINDPM threshold setting Default: 4.5 V (0110) Range: 3.9 V (0000) – 5.4 V (1111) Offset: 3.9 V

LEGEND: R/W = Read/Write; R = Read only

7.5.8 Charger Control 3 Register (Address = 07h) [Reset = 4Ch]

図 7-23. REG07 Register

7	6	5	4	3	2	1	0
0	1	0	0	1	1	0	0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-14. REG07 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	IINDET_EN	0	R/W	by REG_RST by Watchdog	Force input source type detection. After the detection is complete, this bit returns to 0. 0 – Not in input current limit detection. (default) 1 – Force input current limit detection when adapter is present.
6	TMR2X_EN	1	R/W	by REG_RST by Watchdog	Safety timer is slowed by 2X during input DPM, JEITA cool/warm or thermal regulation. 0 – Disable. Safety timer duration is set by REG05[2]. 1 – Safety timer slowed by 2X during input DPM (both V and I) or JEITA cool/warm (except I _{CHG} =100%), or thermal regulation. (default)
5	BATFET_DIS	0	R/W	by REG_RST	BATFET Q4 ON/OFF control. Set this bit to 1 to enter ship mode. To reset the device with adapter present, the host shall set BATFET_RST_WVBUS to 1 and then BATFET_DIS to 1. 0 – Turn on Q4. (default) 1 – Turn off Q4 after t _{BATFET_DLY} delay time (REG07[3])
4	BATFET_RST_WVBUS	0	R/W	by REG_RST	Start BATFET full system reset with or without adapter present. 0 – Start BATFET full system reset after adapter is removed from VBUS. (default) 1 – Start BATFET full system reset when adapter is present on VBUS.
3	BATFET_DLY	1	R/W	by REG_RST	Delay from BATFET_DIS (REG07[5]) set to 1 to BATFET turn off during ship mode. 0 – Turn off BATFET immediately when BATFET_DIS bit is set. 1 – Turn off BATFET after t _{BATFET_DLY} (typ 10 s) when BATFET_DIS bit is set. (default)
2	BATFET_RST_EN	1	R/W	by REG_RST by Watchdog	Enable BATFET full system reset. The time to start of BATFET full system reset is based on the setting of BATFET_RST_WVBUS bit. 0 – Disable BATFET reset function 1 – Enable BATFET reset function when REG07[5] is also 1. (default)
1	VINDPM_BAT_TRACK[1]	0	R/W	by REG_RST	Sets VINDPM to track BAT voltage. Actual VINDPM is higher of register value and V _{BAT} + VINDPM_BAT_TRACK. 00 – Disable function (VINDPM set by register) (default) 01 – V _{BAT} + 200 mV 10 – V _{BAT} + 250 mV 11 – V _{BAT} + 300 mV
0	VINDPM_BAT_TRACK[0]	0	R/W	by REG_RST	

LEGEND: R/W = Read/Write; R = Read only

7.5.9 Charger Status 0 Register (Address = 08h)

 7-24. REG08

7	6	5	4	3	2	1	0
x	x	x	x	x	x	x	x
R	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-15. REG08 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	VBUS_STAT[2]	x	R	NA	VBUS Status register 000 – No input 001 – USB host SDP (500 mA) → PSEL pin HIGH 011 – Adapter 2.4 A → PSEL pin LOW 111 – Boost mode Software current limit is reported in IINDPM register
6	VBUS_STAT[1]	x	R	NA	
5	VBUS_STAT[0]	x	R	NA	
4	CHRG_STAT[1]	x	R	NA	Charging status: 00 – Not charging 01 – Precharge or trickle charge ($< V_{BATLOWV}$) 10 – Fast charging 11 – Charge termination
3	CHRG_STAT[0]	x	R	NA	
2	PG_STAT	x	R	NA	Power Good status: 0 – Power Not Good 1 – Power Good
1	THERM_STAT	x	R	NA	0 – Not in thermal regulation 1 – In thermal regulation
0	VSYS_STAT	x	R	NA	0 – Not in SYS_MIN regulation ($V_{BAT} > V_{SYS_MIN}$) 1 – In SYS_MIN regulation ($V_{BAT} < V_{SYS_MIN}$)

LEGEND: R/W = Read/Write; R = Read only

7.5.10 Charger Status 1 Register (Address = 09h)

図 7-25. REG09 Register

7	6	5	4	3	2	1	0
1	x	x	x	x	x	x	x
R	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-16. REG09 Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	WATCHDOG_FAULT	1	R	NA	0 – Normal, device is in host mode, 1 – Watchdog timer expiration, device is in default mode.
6	BOOST_FAULT	x	R	NA	0 – Normal 1 – Fault detected in boost mode (any conditions that are not valid for boost operation), including VBUS overloaded (BST_OVP) or battery is too low (BST_BAT)
5	CHRG_FAULT[1]	x	R	NA	00 – Normal 01 – Input fault
4	CHRG_FAULT[0]	x	R	NA	10 – Thermal shutdown 11 – Charge safety timer expiration
3	BAT_FAULT	x	R	NA	0 – Normal, 1 – Battery overvoltage.
2	NTC_FAULT[2]	x	R	NA	TS fault in buck mode
1	NTC_FAULT[1]	x	R	NA	000 – Normal 010 – Warm 011 – Cool 101 – Cold 110 – Hot
0	NTC_FAULT[0]	x	R	NA	TS fault in boost mode 000 – Normal 101 – Cold 110 – Hot

LEGEND: R/W = Read/Write; R = Read only

7.5.11 Charger Status 2 Register (Address = 0Ah)

7-26. REG0A Register

7	6	5	4	3	2	1	0
x	x	x	x	x	x	0	0
R	R	R	R	R	R	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-17. REG0A Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	VBUS_GD	x	R	NA	0 – VBUS does not pass poor source detection 1 – VBUS passes poor source detection
6	VINDPM_STAT	x	R	NA	0 – Not in VINDPM 1 – In VINDPM
5	IINDPM_STAT	x	R	NA	0 – Not in IINDPM 1 – In IINDPM
4	BATSNS_STAT	x	R	NA	0 – BATSNS pin is in good connection. Regulation battery voltage through BATSNS pin. 1 – BATSNS pin is open/short. Regulate battery voltage through BAT pin.
3	TOPOFF_ACTIVE	x	R	NA	0 – Top off timer not counting. 1 – Top off timer counting
2	ACOV_STAT	x	R	NA	0 – Not in ACOV 1 – In ACOV
1	VINDPM_INT_MASK	0	R/W	by REG_RST	Allow or block $\overline{\text{INT}}$ pulse assertion to host during VINDPM. 0 – $\overline{\text{INT}}$ is asserted to host during VINDPM (default). 1 – No $\overline{\text{INT}}$ pulse asserted to host during VINDPM.
0	IINDPM_INT_MASK	0	R/W	by REG_RST	Allow or block $\overline{\text{INT}}$ pulse assertion to host during IINDPM. 0 – $\overline{\text{INT}}$ is asserted to host during IINDPM (default). 1 – No $\overline{\text{INT}}$ pulse asserted to host during IINDPM.

LEGEND: R/W = Read/Write; R = Read only

7.5.12 Part Information Register (Address = 0Bh)

7-27. REG0B Register

7	6	5	4	3	2	1	0
0	0	1	0	1	1	0	0
R/W	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-18. REG0B Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	REG_RST	0	R/W	NA	Register reset 0 – Keep current register setting (default). 1 – Reset to default register value and reset safety timer. This bit returns to 0 after register reset is completed.
6	PN[3]	0	R	NA	Reserved
5	PN[2]	1	R	NA	
4	PN[1]	0	R	NA	
3	PN[0]	1	R	NA	
2	Reserved	1	R	NA	Reserved
1	Reserved	0	R	NA	Reserved
0	Reserved	0	R	NA	

LEGEND: R/W = Read/Write; R = Read only

7.5.13 Charger Control 4 Register (Address = 0Ch) [Reset = 75h]

7-28. REG0C

7	6	5	4	3	2	1	0
0	1	1	1	0	1	0	1
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

表 7-19. REG0C Field Descriptions

Bit	Field	POR	Type	Reset	Description
7	JEITA_COOL_ISET [1]	0	R/W	by REG_RST by Watchdog	Fast charge current setting during cool temperature range (T1 - T2), as percentage of I_{CHG} in REG02[5:0]. 00 – No Charge
6	JEITA_COOL_ISET [0]	1	R/W	by REG_RST by Watchdog	01 – 20% of I_{CHG} (default) 10 – 50% of I_{CHG} 11 – 100% of I_{CHG} (safety timer does not become 2X)
5	JEITA_WARM_ISET [1]	1	R/W	by REG_RST by Watchdog	Fast charge current setting during warm temperature range (T3 - T5), as percentage of I_{CHG} in REG02[5:0]. 00 – No Charge
4	JEITA_WARM_ISET [0]	1	R/W	by REG_RST by Watchdog	01 – 20% of I_{CHG} 10 – 50% of I_{CHG} 11 – 100% of I_{CHG} (safety timer does not become 2X) (default)
3	JEITA_VT2 [1]	0	R/W	by REG_RST by Watchdog	00 – VT2% = 70.75% (5.5°C) 01 – VT2% = 68.25% (10°C) (default)
2	JEITA_VT2 [0]	1	R/W	by REG_RST by Watchdog	10 – VT2% = 65.25% (15°C) 11 – VT2% = 62.25% (20°C)
1	JEITA_VT3 [1]	0	R/W	by REG_RST by Watchdog	00 – VT3% = 48.25% (40°C) 01 – VT3% = 44.75% (44.5°C) (default)
0	JEITA_VT3 [0]	1	R/W	by REG_RST by Watchdog	10 – VT3% = 40.75% (50.5°C) 11 – VT3% = 37.75% (54.5°C)

LEGEND: R/W = Read/Write; R = Read only

8 Application and Implementation

注

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8.1 Application Information

A typical application consists of the device configured as an I²C controlled power path management device and a single cell battery charger for Li-ion and Li-polymer batteries used in a wide range of smart phones and other portable devices. It integrates an input reverse-block FET (RBFET, Q1), high-side switching FET (HSFET, Q2), low-side switching FET (LSFET, Q3), and battery FET (BATFET Q4) between the system and battery. The device also integrates a bootstrap diode for the high-side gate drive.

8.2 Typical Application

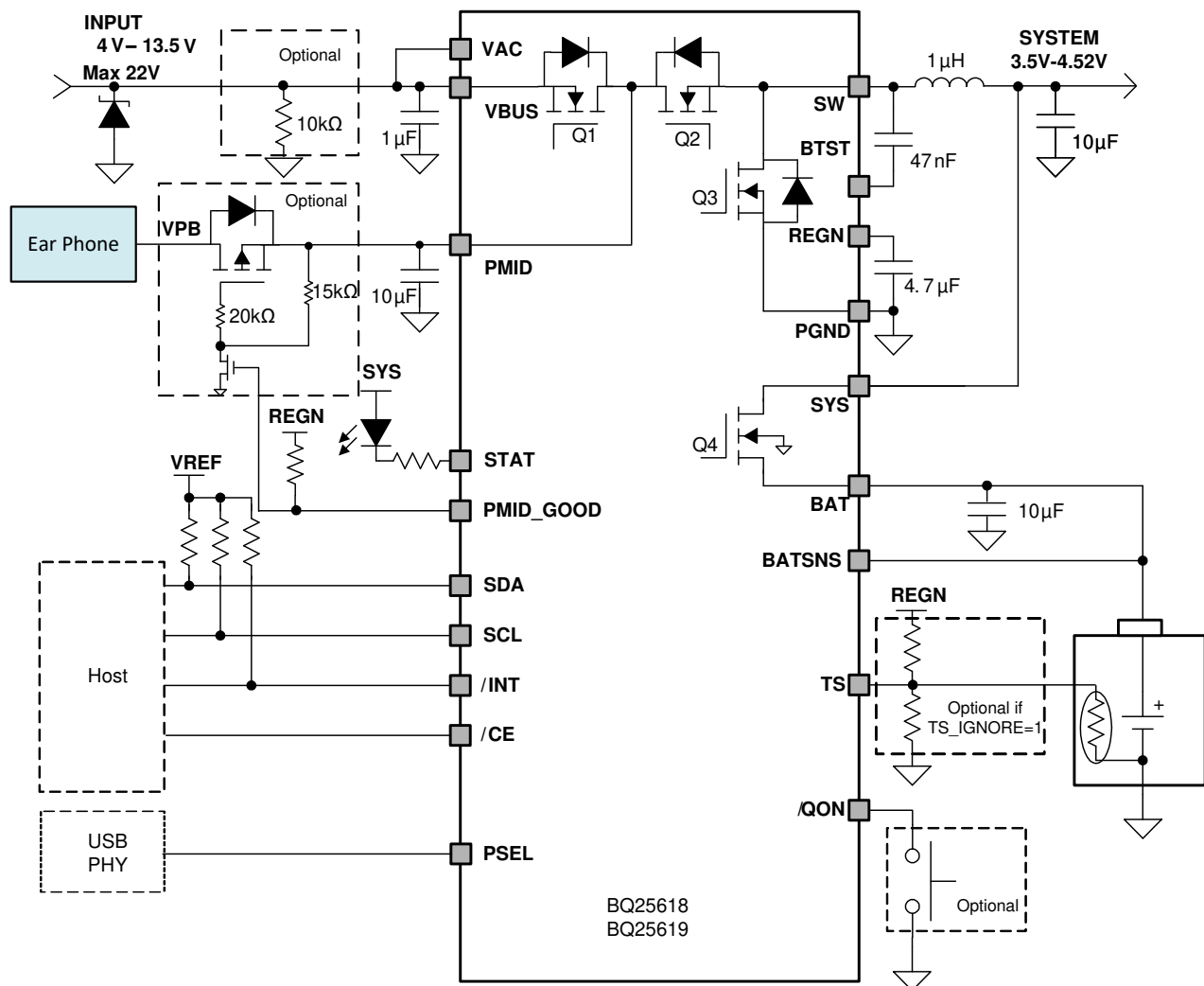


図 8-1. BQ25619 Application Diagram with Optional PMOS

See the [BQ25618 BMS024 Evaluation Module User's Guide](#) and [BQ25619 BMS025 Evaluation Module EVM User's Guide](#) for complete schematic and component placement with trace and via locations.

8.2.1 Design Requirements

For this design example, use the parameters shown in the table below.

表 8-1. Design Parameters

PARAMETER	VALUE
V _{VBUS} voltage range	4 V to 13.5 V
Input current limit (REG00[4:0])	2.4 A
Fast charge current limit (REG02[5:0])	1.024 A
Minimum system voltage (REG01[3:1])	3.5 V
Battery regulation voltage (REG04[7:3])	4.2 V

8.2.2 Detailed Design Procedure

8.2.2.1 Inductor Selection

The 1.5-MHz switching frequency allows the use of small inductor and capacitor values to maintain an inductor saturation current higher than the charging current (I_{CHG}) plus half the ripple current (I_{RIPPLE}):

$$I_{SAT} \geq I_{CHG} + (1/2) I_{RIPPLE} \quad (3)$$

The inductor ripple current depends on the input voltage (V_{VBUS}), the duty cycle ($D = V_{BAT}/V_{VBUS}$), the switching frequency (f_S) and the inductance (L).

$$I_{RIPPLE} = \frac{V_{IN} \times D \times (1 - D)}{f_S \times L} \quad (4)$$

The maximum inductor ripple current occurs when the duty cycle (D) is 0.5 or approximately 0.5. Usually inductor ripple is designed in the range between 20% and 40% maximum charging current as a trade-off between inductor size and efficiency for a practical design.

For compact solution size and efficiency at high current, a 1-μH inductor is recommended. To achieve better light load efficiency during boost mode (output current below 500 mA), the device also supports a 2.2-μH inductor with a 10-μF (min) cap on the system.

8.2.2.2 Input Capacitor and Resistor

Design the input capacitance to provide enough ripple current rating to absorb the input switching ripple current. Worst case RMS ripple current is half of the charging current when the duty cycle is 0.5. If the converter does not operate at 50% duty cycle, then the worst case capacitor RMS current I_{CIN} occurs where the duty cycle is closest to 50% and can be estimated using 式 5.

$$I_{CIN} = I_{CHG} \times \sqrt{D \times (1 - D)} \quad (5)$$

A low ESR ceramic capacitor such as X7R or X5R is preferred for the input decoupling capacitor and should be placed as close as possible to the drain of the high-side MOSFET and source of the low-side MOSFET. The voltage rating of the capacitor must be higher than the normal input voltage level. A 25-V or higher rated capacitor is preferred for a 12-V input voltage. Minimum capacitance of 10 μF is suggested for typical of 1.5-A charging current.

During high current output over 700 mA in boost mode, a 10-kΩ pull-down resistor on VBUS is recommended to keep VBUS low in case Q1 RBFET leakage gets high.

8.2.2.3 Output Capacitor

Ensure that the output capacitance has enough ripple current rating to absorb the output switching ripple current. 式 6 shows the output capacitor RMS current I_{COU}T calculation.

$$I_{\text{COUT}} = \frac{I_{\text{RIPPLE}}}{2 \times \sqrt{3}} \approx 0.29 \times I_{\text{RIPPLE}} \quad (6)$$

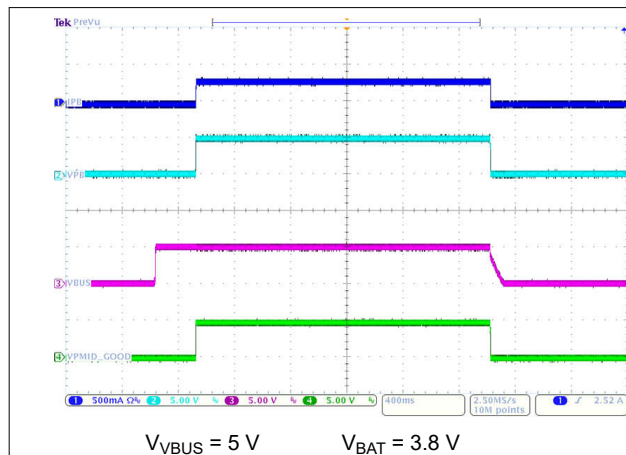
The output capacitor voltage ripple can be calculated as follows:

$$\Delta V_O = \frac{V_{\text{OUT}}}{8LCfs^2} \left(1 - \frac{V_{\text{OUT}}}{V_{\text{IN}}} \right) \quad (7)$$

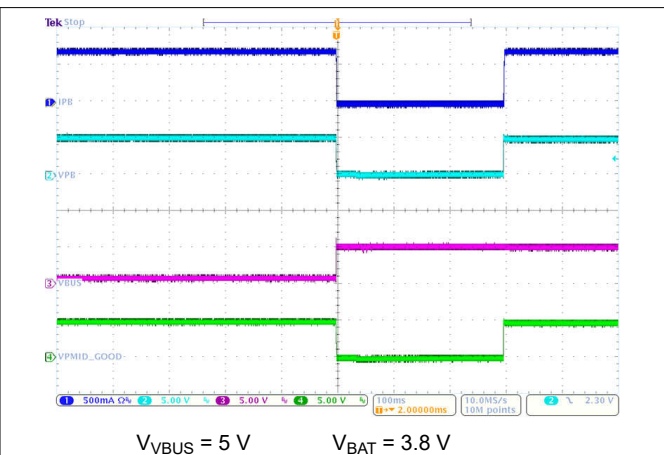
At certain input and output voltages and switching frequency, the voltage ripple can be reduced by increasing the output filter LC.

The charger device has internal loop compensation optimized for >10-μF ceramic output capacitance. The preferred ceramic capacitor is 10-V rating, X7R or X5R.

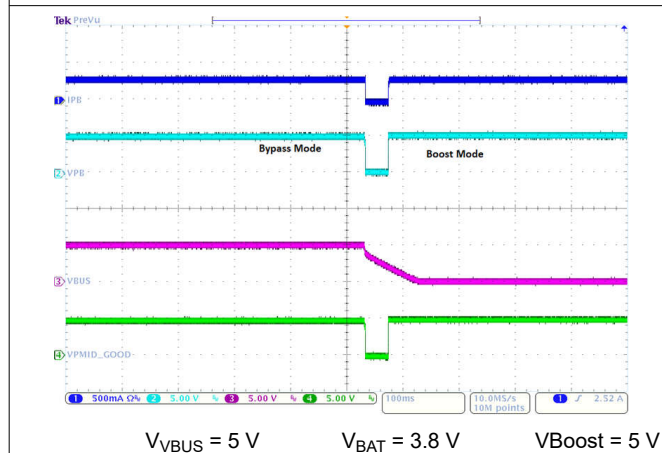
8.2.3 Application Curves



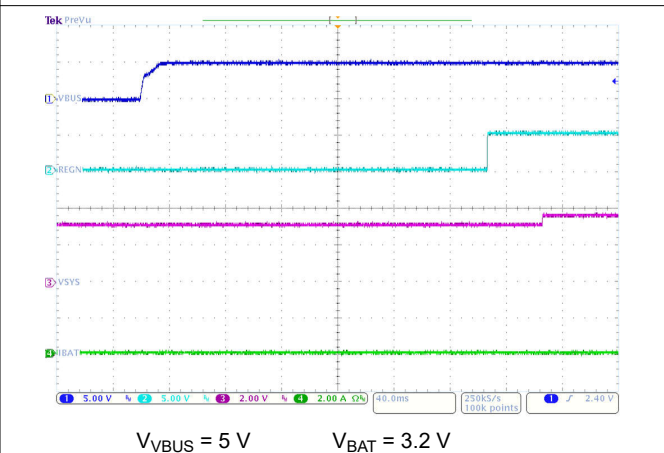
8-2. VBUS Plugged In and Unplugged with Boost Mode Disabled



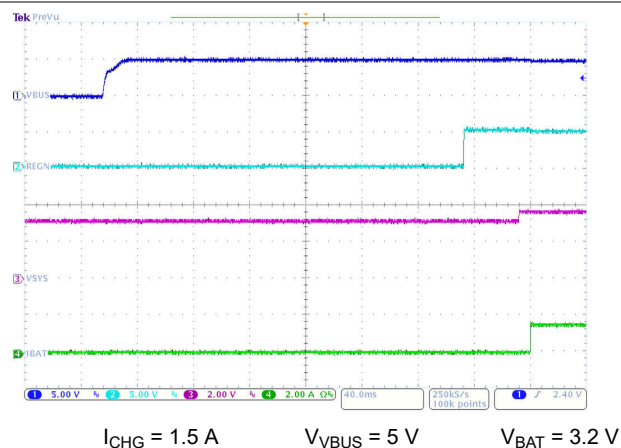
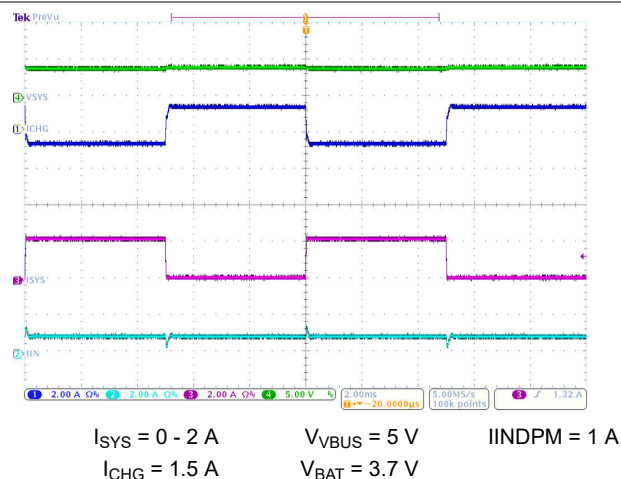
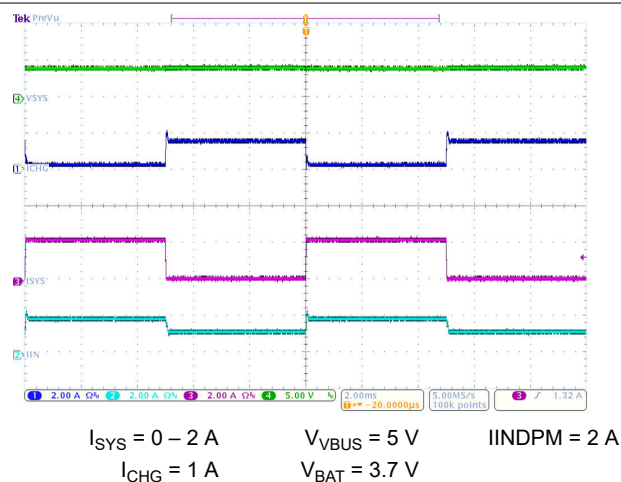
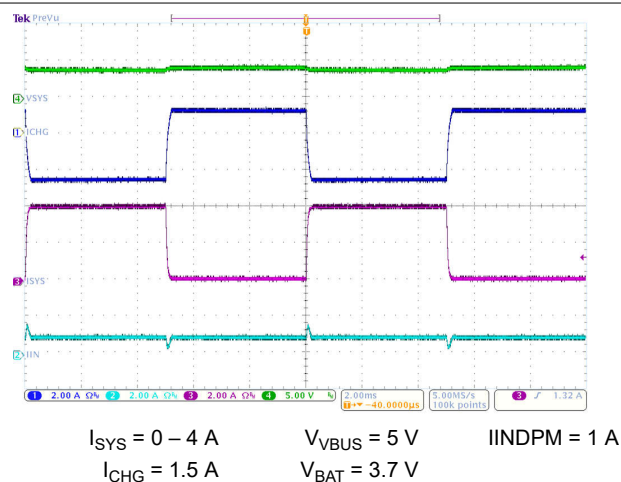
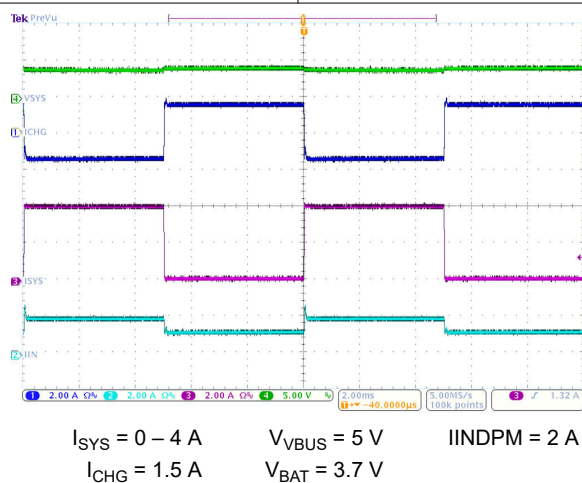
8-3. VBUS Plugged In with Boost Mode Enabled



8-4. VBUS Unplugged with Boost Mode Enabled



8-5. Power Up with Charge Disabled


 **8-6. Power Up with Charge Enabled**

 **8-7. System Load Transient Response**

 **8-8. System Load Transient Response**

 **8-9. System Load Transient Response**

 **8-10. System Load Transient Response**

9 Power Supply Recommendations

In order to provide an output voltage on SYS, the battery charger requires a power supply between 4 V and 13.5 V input with at least a 100-mA current rating connected to VBUS and a single-cell Li-ion battery with battery voltage greater than V_{BAT_UVLOZ} connected to BAT. The source current rating needs to be at least 3 A in order for the buck converter of the charger to provide maximum output power to SYS.

10 Layout

10.1 Layout Guidelines

The switching node rise and fall times should be minimized for minimum switching loss. Proper layout of the components to minimize the high frequency current path loop (see [Figure 10-1](#)) is important to prevent electrical and magnetic field radiation and high frequency resonant problems. Follow this specific order carefully to achieve the proper layout.

1. Place an input capacitor as close as possible to the PMID pin and GND pin connections and use the shortest copper trace connection or GND plane. Add a 1-nF small size (such as 0402 or 0201) decoupling cap for the high frequency noise filter and EMI improvement.
2. Place the inductor input pin as close as possible to SW pin. Minimize the copper area of this trace to lower electrical and magnetic field radiation but make the trace wide enough to carry the charging current. Do not use multiple layers in parallel for this connection. Minimize parasitic capacitance from this area to any other trace or plane.
3. Put the output capacitor near to the inductor and the device. Ground connections need to be tied to the IC ground with a short copper trace connection or GND plane.
4. Route the analog ground separately from power ground. Connect the analog ground and connect power ground separately. Connect the analog ground and power ground together using the thermal pad as the single ground connection point. Or use a 0-Ω resistor to tie the analog ground to power ground.
5. Use a single ground connection to tie the charger power ground to the charger analog ground just beneath the device. Use ground copper pour but avoid power pins to reduce inductive and capacitive noise coupling.
6. Place the decoupling capacitors next to the IC pins and make the trace connection as short as possible.
7. It is critical that the exposed thermal pad on the backside of the device package be soldered to the PCB ground. Ensure that there are sufficient thermal vias directly under the IC, connecting to the ground plane on the other layers.
8. Ensure that the number and sizes of vias allow enough copper for a given current path.

See the [BQ25618 BMS024 Evaluation Module User's Guide](#) and [BQ25619 BMS025 Evaluation Module EVM User's Guide](#) for the recommended component placement with trace and via locations. For the VQFN information, refer to [Quad Flatpack No-Lead Logic Packages Application Report](#) and [QFN and SON PCB Attachment Application Report](#).

10.2 Layout Example

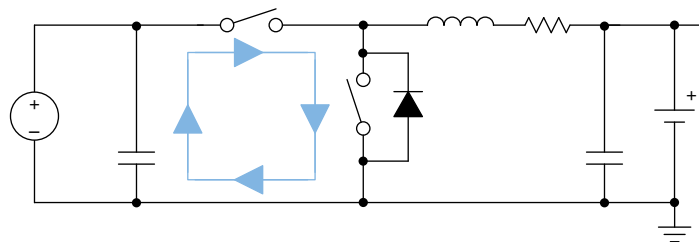
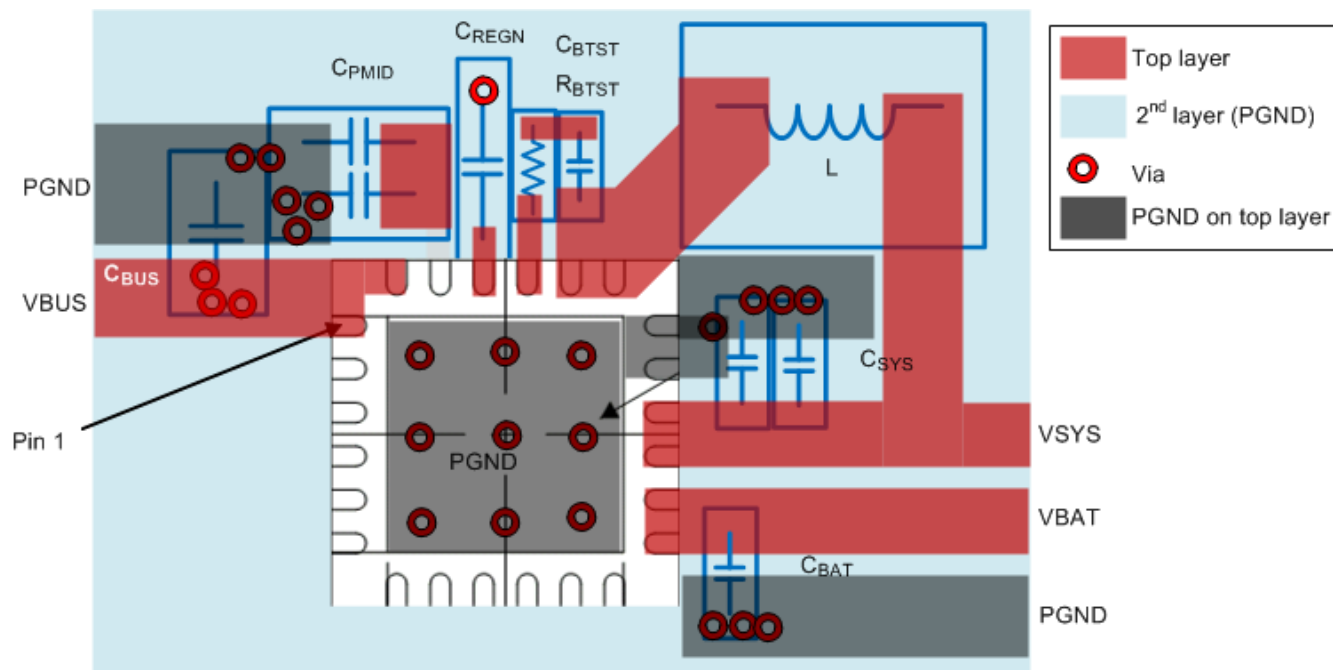


Figure 10-1. High Frequency Current Path




10-2. Layout Example

11 Device and Documentation Support

11.1 Device Support

11.1.1 サード・パーティ製品に関する免責事項

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11.2 Documentation Support

11.2.1 Related Documentation

For related documentation see the following:

- [BQ25619 BMS025 Evaluation Module User's Guide](#)
- [BQ25618 BMS024 Evaluation Module User's Guide](#)

11.3 ドキュメントの更新通知を受け取る方法

ドキュメントの更新についての通知を受け取るには、www.tij.co.jp のデバイス製品フォルダを開いてください。[通知] をクリックして登録すると、変更されたすべての製品情報に関するダイジェストを毎週受け取ることができます。変更の詳細については、改訂されたドキュメントに含まれている改訂履歴をご覧ください。

11.4 サポート・リソース

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11.7 用語集

[テキサス・インスツルメンツ用語集](#) この用語集には、用語や略語の一覧および定義が記載されています。

12 Revision History

資料番号末尾の英字は改訂を表しています。その改訂履歴は英語版に準じています。

Changes from Revision E (May 2024) to Revision F (February 2025)	Page
• 特長 の充電電流レギュレーションを $\pm 5\%$ に更新.....	1
• Updated fast charge current regulation to $\pm 5\%$ in Electrical Characteristics	8

Changes from Revision D (December 2021) to Revision E (July 2024)	Page
• Changed BAT, SYS (converter not switching) MAX value from 17V to 7V in Absolute Maximum Ratings	7
• Changed I _{Q_BAT} and I _{SHIP_BAT} test conditions in Electrical Characteristics	8

Changes from Revision C (June 2021) to Revision D (December 2021)	Page
• Deleted t _{SU_STA} , t _{HD_DAT} , t _{rDA} , and t _{fDA} from Timing Requirements.....	13

Changes from Revision B (September 2019) to Revision C (June 2021)	Page
• 安全関連認証を追加:IEC 62368-1 CB 認証.....	1
• Changed REG2 ICHG[5:0] 111100: 1430 mA to 111110: 1430 mA and 111100: 1500 mA to 111111: 1500 mA.....	33

Changes from Revision A (July 2019) to Revision B (September 2019)	Page
• BQ25618 を追加.....	1

Changes from Revision * (June 2019) to Revision A (July 2019)	Page
• 「事前情報」から「量産データ」に変更.....	1

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
BQ25618YFFR	Active	Production	DSBGA (YFF) 30	3000 LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 85	BQ25618
BQ25618YFFR.A	Active	Production	DSBGA (YFF) 30	3000 LARGE T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 85	BQ25618
BQ25618YFFT	Active	Production	DSBGA (YFF) 30	250 SMALL T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 85	BQ25618
BQ25618YFFT.A	Active	Production	DSBGA (YFF) 30	250 SMALL T&R	Yes	SNAGCU	Level-1-260C-UNLIM	-40 to 85	BQ25618
BQ25619RTWR	Active	Production	WQFN (RTW) 24	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ25619
BQ25619RTWR.A	Active	Production	WQFN (RTW) 24	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ25619
BQ25619RTWT	Active	Production	WQFN (RTW) 24	250 SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ25619
BQ25619RTWT.A	Active	Production	WQFN (RTW) 24	250 SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ25619

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "-" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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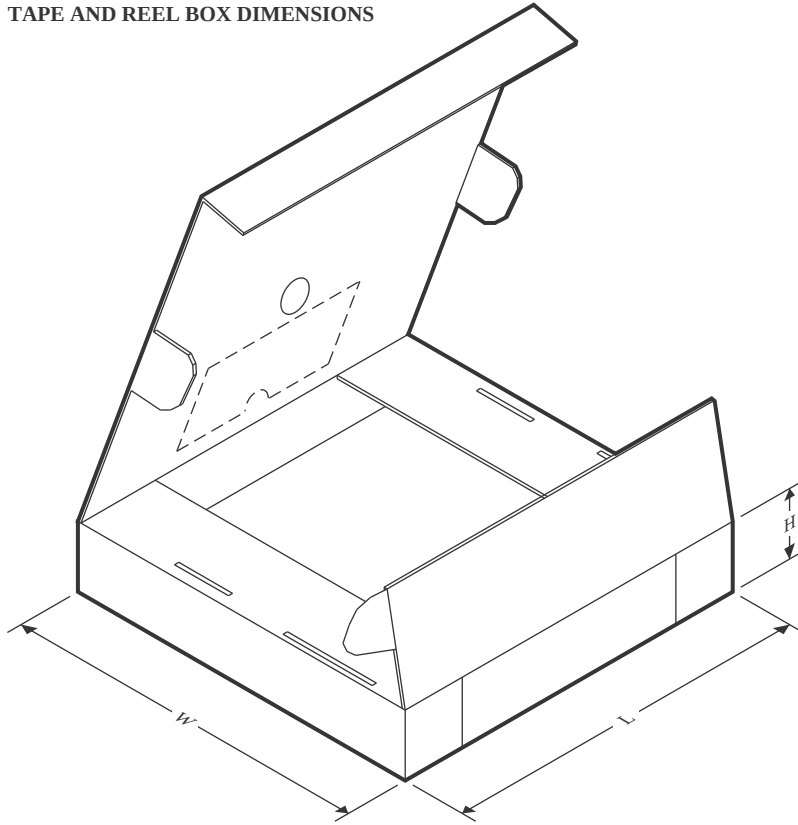
TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
BQ25618YFFR	DSBGA	YFF	30	3000	180.0	8.4	2.09	2.59	0.78	4.0	8.0	Q1
BQ25618YFFT	DSBGA	YFF	30	250	180.0	8.4	2.09	2.59	0.78	4.0	8.0	Q1
BQ25619RTWR	WQFN	RTW	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
BQ25619RTWT	WQFN	RTW	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

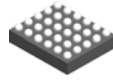
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
BQ25618YFFR	DSBGA	YFF	30	3000	182.0	182.0	20.0
BQ25618YFFT	DSBGA	YFF	30	250	182.0	182.0	20.0
BQ25619RTWR	WQFN	RTW	24	3000	367.0	367.0	35.0
BQ25619RTWT	WQFN	RTW	24	250	210.0	185.0	35.0

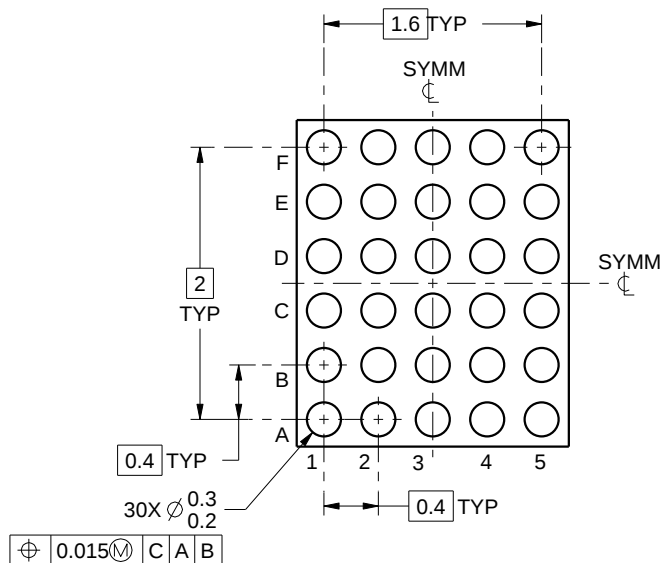
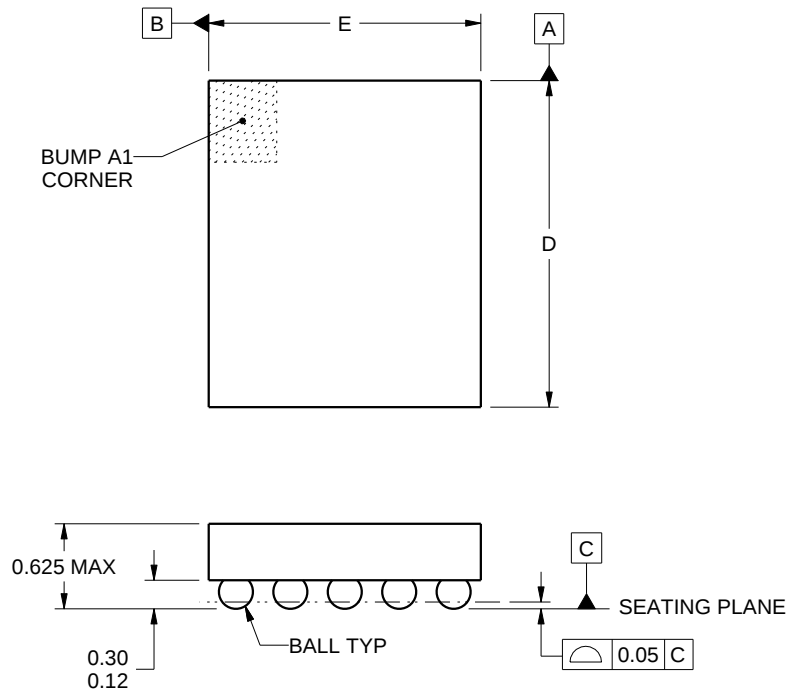
YFF0030



PACKAGE OUTLINE

DSBGA - 0.625 mm max height

DIE SIZE BALL GRID ARRAY



D: Max = 2.392 mm, Min = 2.332 mm

E: Max = 1.992 mm, Min = 1.931 mm

4219433/A 03/2016

NOTES:

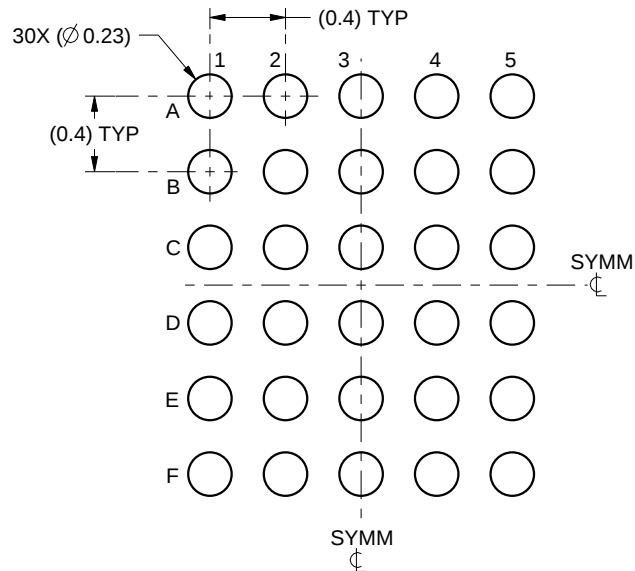
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.

EXAMPLE BOARD LAYOUT

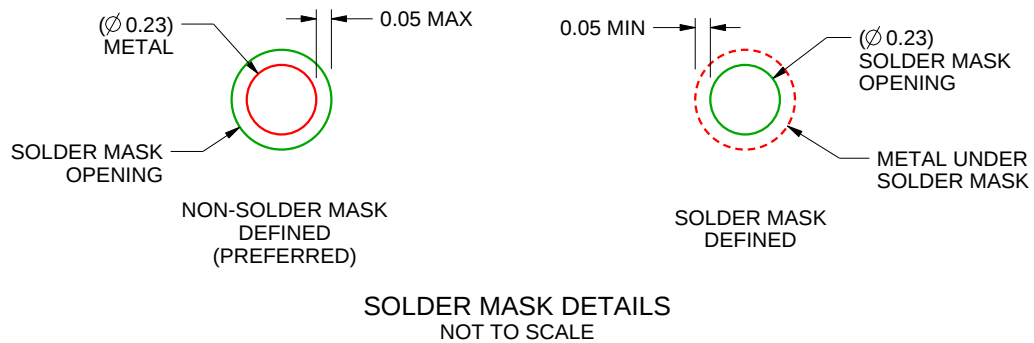
YFF0030

DSBGA - 0.625 mm max height

DIE SIZE BALL GRID ARRAY



LAND PATTERN EXAMPLE
SCALE:25X



4219433/A 03/2016

NOTES: (continued)

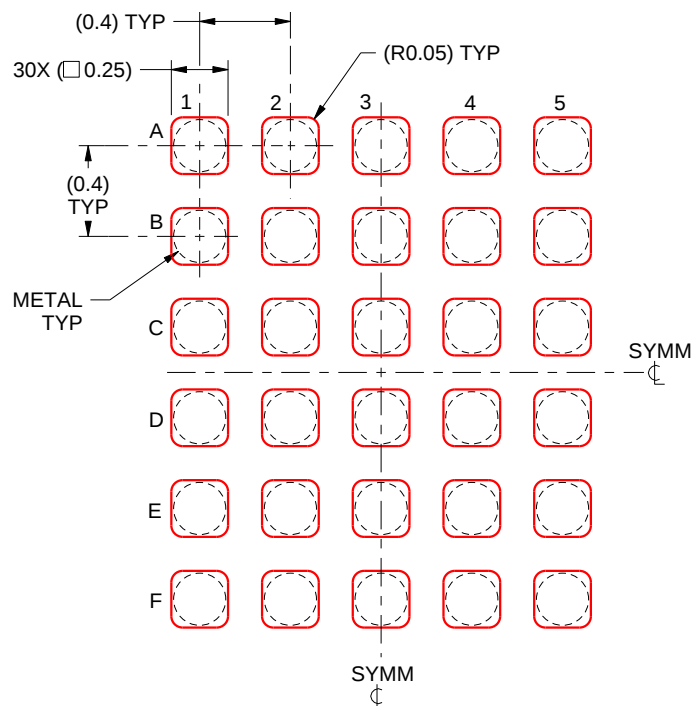
- Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints. For more information, see Texas Instruments literature number SNVA009 (www.ti.com/lit/snva009).

EXAMPLE STENCIL DESIGN

YFF0030

DSBGA - 0.625 mm max height

DIE SIZE BALL GRID ARRAY



SOLDER PASTE EXAMPLE
BASED ON 0.1 mm THICK STENCIL
SCALE:30X

4219433/A 03/2016

NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.

GENERIC PACKAGE VIEW

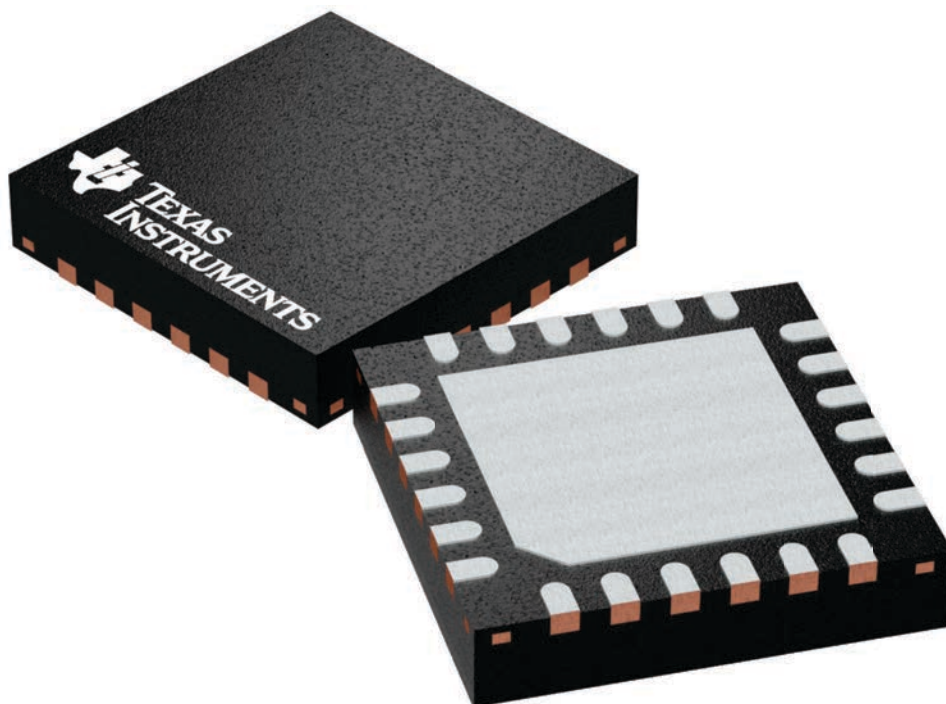
RTW 24

WQFN - 0.8 mm max height

4 x 4, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

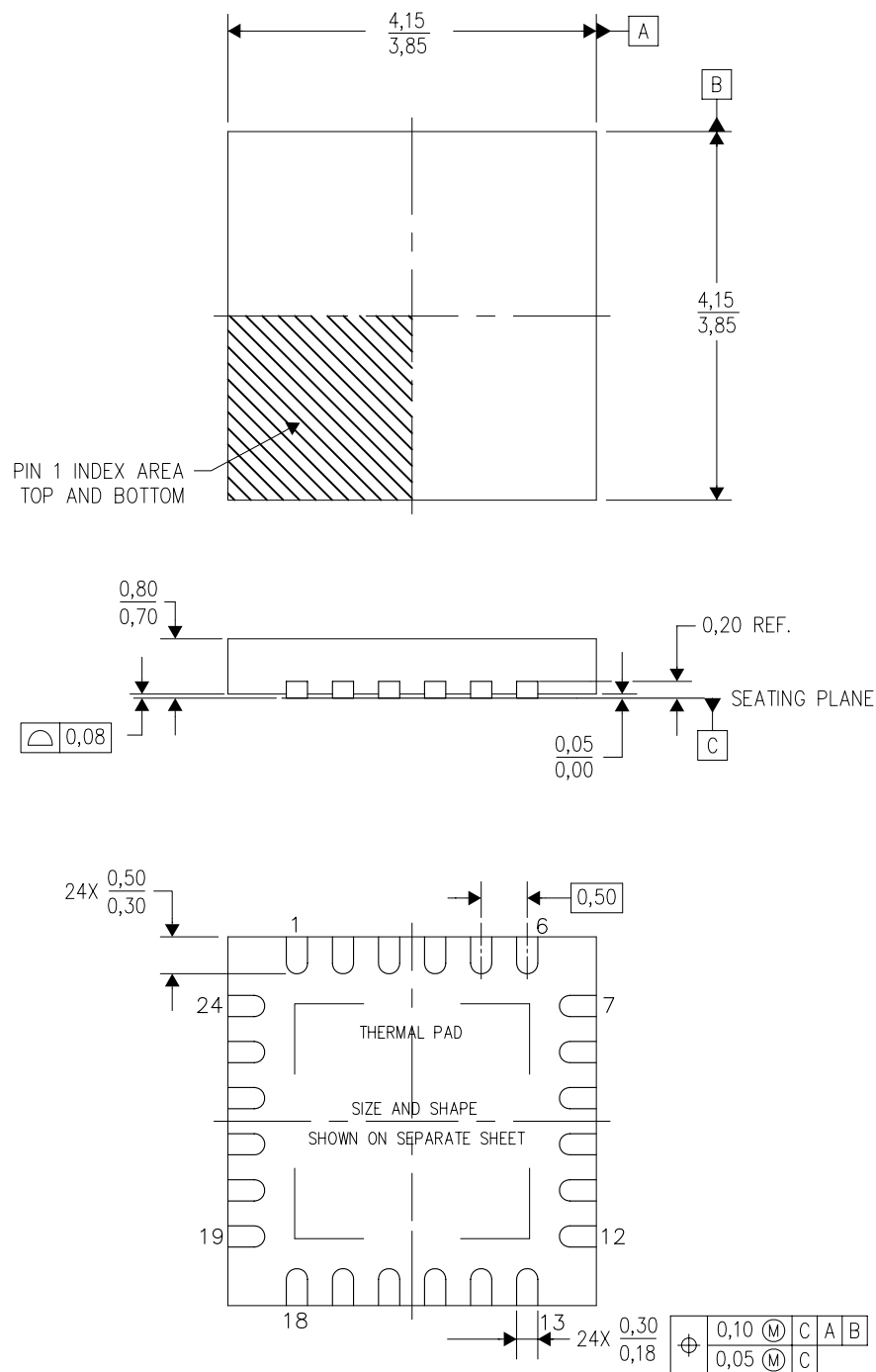
This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4224801/A

RTW (S-PWQFN-N24)

PLASTIC QUAD FLATPACK NO-LEAD



4206244/C 07/11

- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 - C. Quad Flatpack, No-Leads (QFN) package configuration.
 - D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
 - E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
 - F. Falls within JEDEC MO-220.

RTW (S-PWQFN-N24)

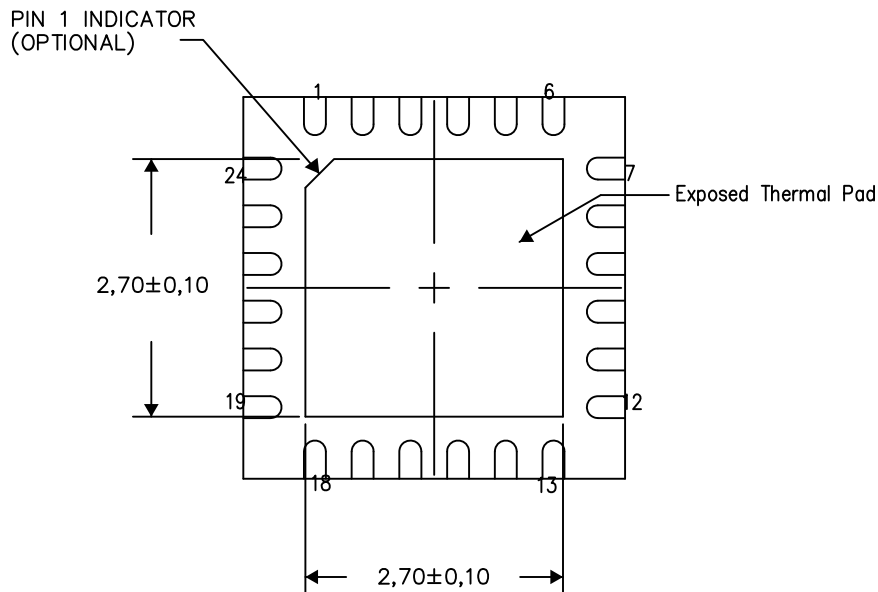
PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

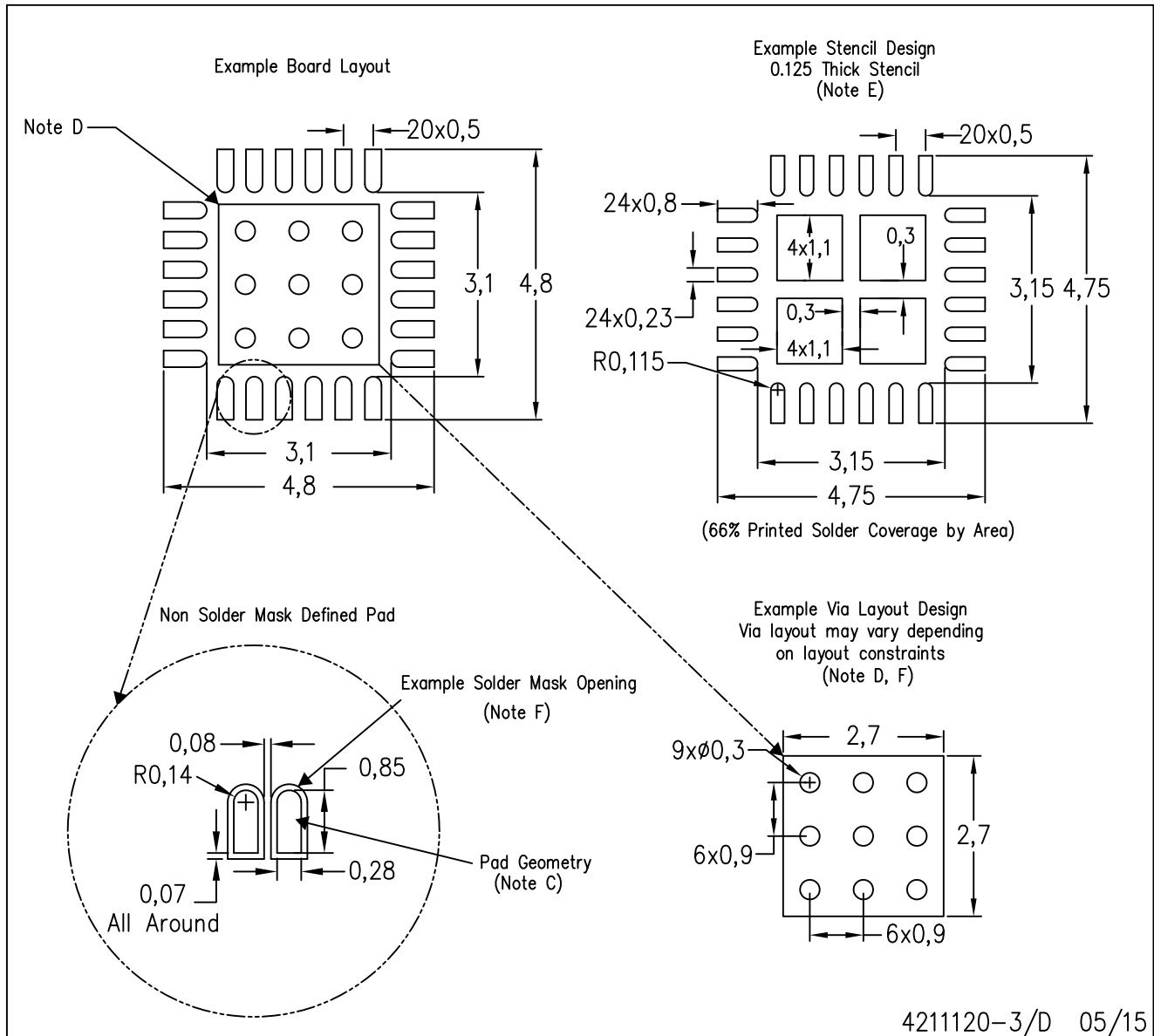
Exposed Thermal Pad Dimensions

4206249-5/P 05/15

NOTES: A. All linear dimensions are in millimeters

RTW (S-PWQFN-N24)

PLASTIC QUAD FLATPACK NO-LEAD



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.

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