LP8758-B0 4相DC/DC降圧型コンバータ

1 特長

- 高効率の降圧型4相DC/DCコンバータ・コア
  - 最大出力電流: 16A
  - PWM-PFM自動切り換え動作と強制PWM動作
  - 自動位相追加/シェディングおよび強制マルチフェーズ動作
  - リモート差動帰還電圧センシング
  - 出力電圧スルー・レートを30mV/µs～0.5mV/µsの範囲でプログラム可能
  - V_{OUT}範囲 = 0.5V～3.36V（DVS付き）
- プログラム可能なスタートアップおよびシャットダウン遅延（イネーブル信号に同期）
- I^2C互換インターフェイスのStandard (100kHz)、Fast (400kHz)、Fast+ (1MHz)、High-Speed (3.4MHz)モードをサポート
- マスクをプログラム可能な割り込み機能
- 負荷電流測定
- 出力短絡および過負荷保護
- 拡散スペクトラム・モードおよび位相インターリーブによるEMIの低減
- 過熱警告および保護
- 低電圧誤動作防止(UVLO)

2 アプリケーション

- スマートフォン、電子書籍、タブレット
- ゲーム機

3 概要

LP8758は、携帯電話や同様の携帯機器アプリケーションに使用される、最先端のアプリケーション・プロセッサの電力管理要件を満たすように設計されています。このデバイスには、4つの降圧型DC/DCコンバータ・コアが内蔵され、互いに結合されて1つの4相降圧型コンバータを形成しています。このデバイスは、I^2C互換のシリアル・インターフェイスにより制御されます。

PWM-PFMの自動切り換え（AUTOモード）動作と、自動位相追加/シェディングにより、広い出力電流範囲にわたって効率を最大化します。LP8758はリモート差動電圧検出をサポートしており、レギュレータ出力と負荷ポイントとの間のIR降下を補償することで、出力電圧の精度を向上させています。

LP8758は、イネーブル信号に同期したプログラム可能なスタートアップおよびシャットダウン遅延をサポートしています。

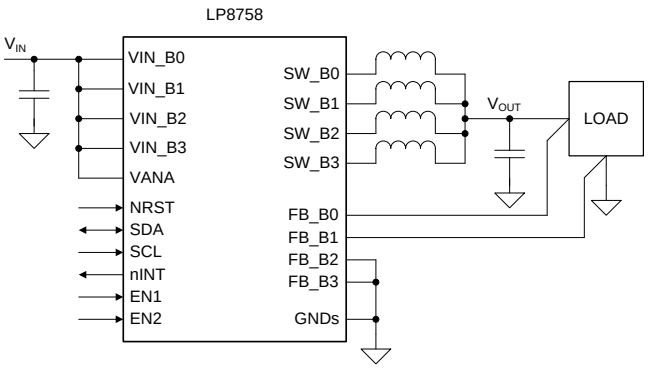
保護機能として、短絡保護、電流制限、入力電源UVLO、温度警告およびシャットダウン機能が搭載されています。デバイスのステータス情報のための複数のエラー・フラグが用意されています。さらに、LP8758デバイスは、外付けの電流検出抵抗を追加しないで負荷電流を測定できます。スタートアップ中と電圧変化時に、本デバイスは出力スルー・レートを制御し、出力電圧のオーバーシュートと突入電流を最小化します。

製品情報(1)

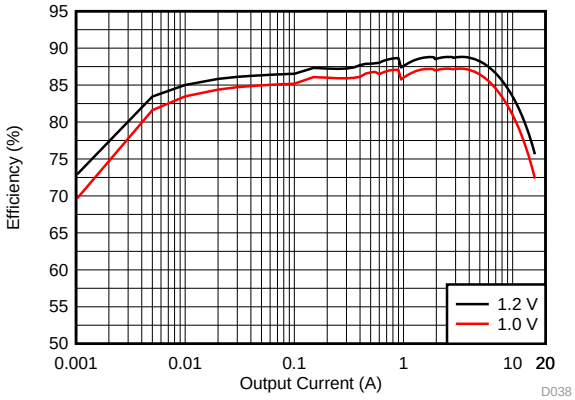
型番	パッケージ	本体サイズ（公称）
LP8758-B0	DSBGA (35)	2.88mm×2.13mm

(1) 利用可能なすべてのパッケージについては、このデータシートの末尾にある注文情報を参照してください。

4 概略回路図



効率と出力電流との関係(V_{IN} = 3.7V)



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5 改訂履歴

Revision B (May 2016) から **Revision C** に変更 **Page**

- Changed logic low level to 0 V and high level to VANA up to 3.6 V ... 5
- Added support for I2C signals up to 3.3V ... 5
- Changed "700 µs" to "1.2 ms" ... 17

Revision A (May 2015) から **Revision B** に変更 **Page**

2015年3月発行のものから更新 **Page**

- 最初のWebリリース ... 1
- 追加「コミュニティ・リソース」セクション ... 50

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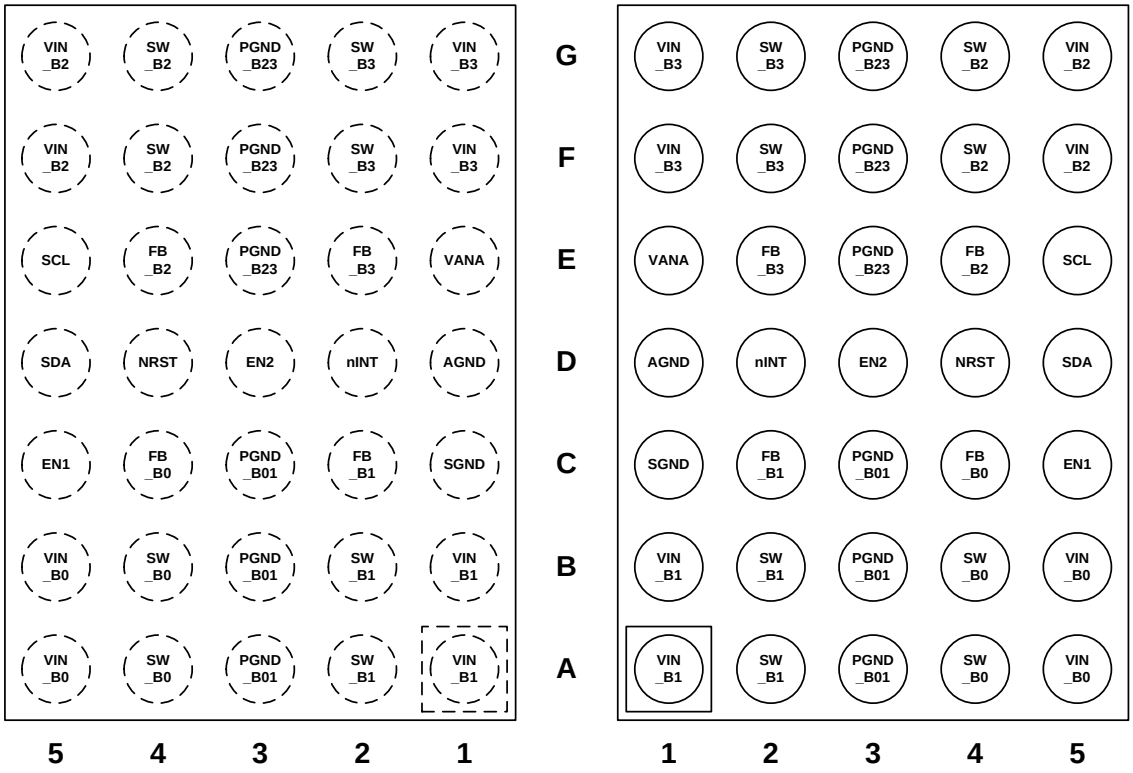
6 Pin Configuration and Functions

YFF Package
35-Pin DSBGA



Top View (Bump-side down)

Bottom View

Pin Functions

PIN		TYPE	DESCRIPTION
NUMBER	NAME		
A1, B1	VIN_B1	P	Input for Buck 1. The separate power pins VIN_Bx are not connected together internally - VIN_Bx pins must be connected together in the application and be locally bypassed.
A2, B2	SW_B1	A	Buck 1 switch node.
A3, B3, C3	PGND_B01	G	Power Ground for Buck 0 and Buck 1.
A4, B4	SW_B0	A	Buck 0 switch node.
A5, B5	VIN_B0	P	Input for Buck 0. The separate power pins VIN_Bx are not connected together internally - VIN_Bx pins must be connected together in the application and be locally bypassed.
C1	SGND	G	Substrate Ground.
C2	FB_B1	A	Output ground feedback (negative) for Buck 0.
C4	FB_B0	A	Output voltage feedback (positive) for Buck 0.
C5	EN1	D/I	Programmable Enable signal for Buck regulator. Can be also configured to switch between two output voltage levels.
D1	AGND	G	Ground.
D2	nINT	D/O	Open-drain interrupt output. Active LOW.
D3	EN2	D/I	Programmable Enable signal for Buck regulator. Can be also configured to switch between two output voltage levels.
D4	NRST	D/I	Reset signal for the device.
D5	SDA	D/I/O	Serial interface data input and output for system access. Connect a pull-up resistor.
E1	VANA	P	Supply voltage for Analog and Digital blocks. VANA pin must be connected to same voltage as VIN_Bx pins.
E2	FB_B3	A	Output voltage feedback (positive) for Buck 3 - Connect to ground in 4-phase configuration.
E4	FB_B2	A	Output voltage feedback (positive) for Buck 2. - Connect to ground in 4-phase configuration.
E5	SCL	D/I	Serial interface clock input for system access. Connect a pull-up resistor.
F1, G1	VIN_B3	P	Input for Buck 3. The separate power pins VIN_Bx are not connected together internally - VIN_Bx pins must be connected together in the application and be locally bypassed.
F2, G2	SW_B3	A	Buck 3 switch node.
E3, F3, G3	PGND_B23	G	Power Ground for Buck 2 and Buck 3.
F4, G4	SW_B2	A	Buck 2 switch node.
F5, G5	VIN_B2	P	Input for Buck 2. The separate power pins VIN_Bx are not connected together internally - VIN_Bx pins must be connected together in the application and be locally bypassed.
A: Analog Pin, D: Digital Pin, G: Ground Pin, P: Power Pin, I: Input Pin, O: Output Pin			

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7 Specifications

7.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)(1)(2)

		MIN	MAX	UNIT
INPUT VOLTAGE				
VIN_Bx, VANA	Voltage on power connections	–0.3	6	V
SW_Bx	Voltage on buck switch nodes	–0.3	(VIN_Bx + 0.3) with 6 V max	V
FB_Bx	Voltage on buck voltage sense nodes	–0.3	(VANA + 0.3) with 6 V max	V
NRST	Voltage on NRST input	–0.3	3.6	V
ENx, SDA, SCL, nINT	Voltage on logic pins (input or output pins)	–0.3	3.6	
CURRENT				
VIN_Bx, SW_Bx, PGND_Bx	Current on power pins (average current over 100k hour lifetime, T_J = 125°C)		0.62	A/pin
TEMPERATURE				
Junction temperature, T_{J-MAX}		−40	150	°C
Storage temperature, T_{stg}		–65	150	°C
Maximum lead temperature (soldering, 10 sec.)(3)			260	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltage values are with respect to network ground.
(3) For detailed soldering specifications and information, refer to *AN-1112 DSBGA Wafer Level Chip Scale Package* (SNVA009).

7.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001(1)	±2000	V
		Charged-device model (CDM), per JEDEC specification JESD22-C101(2)	±500	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
INPUT VOLTAGE				
VIN_Bx, VANA	Voltage on power connections	2.5	5.5	V
NRST	Voltage on NRST	0	VANA up to 3.6	V
ENx, nINT	Voltage on logic pins (input or output pins)	0	VANA up to 3.6	V
SCL, SDA	Voltage on I^2C interface, standard (100 kHz), fast (400 khz), fast+ (1 MHz), and high-speed (3.4 MHz) modes	0	1.95	V
	Voltage on I^2C interface, standard (100 kHz), fast (400 kHz), and fast+ (1 MHz) modes	0	VANA up to 3.6	V
TEMPERATURE				
Junction temperature, T_J		−40	125	°C
Ambient temperature, T_A		−40	85	°C

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7.4 Thermal Information

THERMAL METRIC[1]		LP8758 YFF (DSBGA) 35 PINS	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	56.1	°C/W
$R_{\theta JCtop}$	Junction-to-case (top) thermal resistance	0.2	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	8.5	°C/W
ψ_{JT}	Junction-to-top characterization parameter	0.9	°C/W
ψ_{JB}	Junction-to-board characterization parameter	8.4	°C/W
$R_{\theta JCbot}$	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, SPRA953.

7.5 Electrical Characteristics

Limits apply over the junction temperature range –40°C ≤ T_J ≤ +125°C, specified $V_{(VANA)}$, V_{IN} , $V_{(NRST)}$, V_{OUT} and I_{OUT} range, unless otherwise noted. Typical values are at T_J = 25°C, f_{SW} = 3 MHz, $V_{(VANA)}$ = V_{IN} = 3.7 V and V_{OUT} = 1 V unless otherwise noted.[1][2]

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
EXTERNAL COMPONENTS						
C_{IN}	Input filtering capacitance	Connected from VIN_Bx to PGND_Bx	1.9	10		µF
C_{OUT}	Output filtering capacitance, local	Capacitance per phase	10	22		µF
$C_{OUT-TOTAL}$	Output capacitance, total (local and remote)	Total output capacitance, 4-phase configuration	40		200	µF
ESR_C	Input and output capacitor ESR	[1-10] MHz		2	10	mΩ
L	Inductor	Inductance of the inductor		0.33 or 0.47		µH
			–30%		30%	
DCR_L	Inductor DCR	TOKO, DFE252010F-R33M		16		mΩ
BUCK REGULATOR						
V_{IN}	Input voltage range	Voltage between VIN_Bx and ground pins. VANA must be connected to the same supply as VIN_Bx.	2.5	3.7	5.5	V
V_{OUT}	Output voltage	Programmable voltage range	0.5	1	3.36	V
		Step size, 0.5 V ≤ V_{OUT} < 0.73 V		10		mV
		Step size, 0.73 V ≤ V_{OUT} < 1.4 V		5		
		Step size, 1.4 V ≤ V_{OUT} ≤ 3.36 V		20		
I_{OUT}	Output current	Output current, 4-phase configuration			12[3]	A
		Output current, 4-phase configuration, V_{IN} > 3 V, V_{OUT} < 2 V			16[3]	
	Dropout voltage	V_{IN} – V_{OUT}	0.7			V
	DC output voltage accuracy, includes voltage reference, DC load and line regulations, process and temperature	Forced PWM mode, 0.8 V ≤ V_{OUT} ≤ 1.2 V, 2.5 V ≤ V_{IN} ≤ 4.5 V, T_J = 25°C, 0 ≤ I_{OUT} ≤ $I_{OUT(max)}$	-1%		1.5%	
		PFM mode, the average output voltage level is increased by max. 20 mV	min (–2%, –15 mV)		max (2%, 15 mV) 20 mV	

(1) All voltage values are with respect to network ground.
(2) Minimum (MIN) and maximum (MAX) limits are specified by design, test, or statistical analysis. Typical (TYP) numbers are not verified, but do represent the most likely norm.
(3) The maximum output current is also limited by the junction temperature and maximum average current over lifetime. The power dissipation inside the die increases the junction temperature and limits the maximum current depending of the length of the current pulse, efficiency, board and ambient temperature. The maximum average current/pin over lifetime is described in *Absolute Maximum Ratings*.

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Electrical Characteristics (continued)

Limits apply over the junction temperature range –40°C ≤ T_J ≤ +125°C, specified $V_{(VANA)}$, V_{IN} , $V_{(NRST)}$, V_{OUT} and I_{OUT} range, unless otherwise noted. Typical values are at T_J = 25°C, f_{SW} = 3 MHz, $V_{(VANA)}$ = V_{IN} = 3.7 V and V_{OUT} = 1 V unless otherwise noted.(1)(2)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Ripple, 4-phase configuration	PWM mode, L = 0.33 µH		10		mV_{p-p}
		PFM mode, L = 0.33 µH		10		
DC_{LNR}	DC line regulation	I_{OUT} = $I_{OUT(max)}$		±0.05		%/V
DC_{LDR}	DC load regulation in PWM mode	I_{OUT} from 0 to $I_{OUT(max)}$		0.3%		
T_{LDSR}	Undershoot for transient load step response, 4-phase configuration	I_{OUT} = 1 A to 8 A, T_R = 400 ns, PWM mode, C_{OUT} = 100 µF, L = 0.33 µH		-45		mV
		2.5 V ≤ V_{IN} ≤ 4.5 V, 0.8 V ≤ V_{OUT} ≤ 1.2 V, I_{OUT} = 0.1 A to 4.1 A, T_R = 100 ns, AUTO mode, C_{OUT} = 100 µF, L = 0.33 µH		–35		mV
		3 V ≤ V_{IN} ≤ 4.5 V, 0.8 V ≤ V_{OUT} ≤ 1.2 V, I_{OUT} from 1 A to 12 A, T_R = 1000 ns, C_{OUT} = 100 µF, L = 0.33 µH		–45		mV
	Overshoot for transient load step response, 4-phase configuration	I_{OUT} = 8 A to 1 A, T_F = 400 ns, PWM mode, C_{OUT} = 100 µF, L = 0.33 µH		45		mV
		2.5 V ≤ V_{IN} ≤ 4.5 V, 0.8 V ≤ V_{OUT} ≤ 1.2 V, I_{OUT} = 4.1 A to 0.1 A, T_F = 100 ns, AUTO mode, C_{OUT} = 100 µF, L = 0.33 µH		25		mV
		3 V ≤ V_{IN} ≤ 4.5 V, 0.8 V ≤ V_{OUT} ≤ 1.2 V, I_{OUT} from 12 A to 1 A, T_F = 1000 ns, C_{OUT} = 100 µF, L = 0.33 µH		50		mV
T_{LNSR}	Transient line response	V_{IN} stepping 2.5 V ↔ 3 V, T_R = T_F = 10 µs, I_{OUT} = $I_{OUT(max)}$		±20		mV
$I_{LIM\ FWD}$	Forward current limit (peak for every switching cycle)	Programmable range	1.5		5	A
		Step size		0.5		
		Accuracy, 3 V ≤ V_{IN} ≤ 5.5 V, I_{LIM} = 5 A	–5%	7.5%	20%	
		Accuracy, 2.5 V ≤ V_{IN} < 3 V, I_{LIM} = 5 A	–20%	7.5%	20%	
$I_{LIM\ NEG}$	Negative current limit		1.6	2	2.4	A
$R_{DS(ON)\ HS\ FET}$	On-resistance, high-side FET	Each phase, between VIN_Bx and SW_Bx pins (I = 1 A)		40	90	mΩ
$R_{DS(ON)\ LS\ FET}$	On-resistance, low-side FET	Each phase, between SW_Bx and PGND_Bx pins (I = 1 A)		33	50	mΩ
	Current balancing	Current mismatch between phases, I_{OUT} > 1000 mA / phase, 0.8 V ≤ V_{OUT} ≤ 1.2 V			10%	
	Overshoot during start-up	V_{OUT} = 1 V, Slew rate = 10 mV/µs			50	mV
$I_{PFM-PWM}$	PFM-to-PWM transition - current threshold(4)			600		mA
$I_{PWM-PFM}$	PWM-to-PFM transition - current threshold(4)			240		mA
I_{ADD}	Phase-adding level	From 1-phase to 2-phase		1000		mA
		From 2-phase to 3-phase		2000		
		From 3-phase to 4-phase		3000		
I_{SHED}	Phase-shedding level	From 2-phase to 1-phase		750		mA
		From 3-phase to 2-phase		1500		
		From 4-phase to 3-phase		2300		
	Output pulldown resistance	Regulator disabled	150	250	350	Ω

(4) The final PFM-to-PWM and PWM-to-PFM transition current varies slightly and is dependant on the output voltage, input voltage, and the magnitude of inductor's ripple current.

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Electrical Characteristics (continued)

Limits apply over the junction temperature range –40°C ≤ T_J ≤ +125°C, specified $V_{(VANA)}$, V_{IN} , $V_{(NRST)}$, V_{OUT} and I_{OUT} range, unless otherwise noted. Typical values are at T_J = 25°C, f_{SW} = 3 MHz, $V_{(VANA)}$ = V_{IN} = 3.7 V and V_{OUT} = 1 V unless otherwise noted.(1)(2)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
	Powergood threshold for interrupt BUCKx_INT(BUCKx_SC_INT), difference from final voltage	Rising ramp voltage, enable or voltage change	–23	–17	–10	mV
		Falling ramp, voltage change	10	17	23	
	Powergood threshold for status signal BUCKx_STAT(BUCKx_PG_STAT)	During operation, status signal is forced to '0' during voltage change	–23	–17	–10	mV
PROTECTION FEATURES						
	Thermal warning	Temperature rising, CONFIG(TDIE_WARN_LEVEL) = 0		125		°C
		Temperature rising, CONFIG(TDIE_WARN_LEVEL) = 1		105		
		Hysteresis		15		
	Thermal shutdown	Temperature rising		150		°C
		Hysteresis		15		
$VANA_{UVLO}$	VANA undervoltage lockout	Voltage falling	2.3	2.4	2.5	V
		Hysteresis		50		mV
LOAD CURRENT MEASUREMENT						
	Current measurement range	Maximum code		20.46		A
	Resolution	LSB		20		mA
	Measurement accuracy	I_{OUT} ≥ 2 A		<10%		
CURRENT CONSUMPTION						
	Shutdown current consumption	$V_{(NRST)}$ = 0 V		1		µA
	Standby current consumption, regulator disabled	$V_{(NRST)}$ = 1.8 V		6		µA
	Active current consumption during PFM operation	$V_{(NRST)}$ = 1.8 V, I_{OUT} = 0 mA, not switching		71		µA
	Active current consumption during PWM operation	$V_{(NRST)}$ = 1.8 V, I_{OUT} = 0 mA		18		mA

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Electrical Characteristics (continued)

Limits apply over the junction temperature range –40°C ≤ T_J ≤ +125°C, specified $V_{(VANA)}$, V_{IN} , $V_{(NRST)}$, V_{OUT} and I_{OUT} range, unless otherwise noted. Typical values are at T_J = 25°C, f_{SW} = 3 MHz, $V_{(VANA)}$ = V_{IN} = 3.7 V and V_{OUT} = 1 V unless otherwise noted.(1)(2)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
DIGITAL INPUT SIGNALS NRST, ENx, SCL, SDA						
V_{IL}	Input low level				0.4	V
V_{IH}	Input high level		1.2			V
V_{HYS}	Hysteresis of Schmitt trigger inputs (SCL, SDA)		10	80	160	mV
	ENx pulldown resistance	ENx_PD = 1		500		kΩ
	NRST pulldown resistance	Always present	800	1200	1700	kΩ
DIGITAL OUTPUT SIGNALS nINT, SDA						
V_{OL}	Output low level	I_{SOURCE} = 2 mA			0.4	V
R_P	External pullup resistor for nINT	To VIO Supply		10		kΩ
ALL DIGITAL INPUTS						
I_{LEAK}	Input current	All logic inputs over pin voltage range	−1		1	µA

7.6 I²C Serial Bus Timing Parameter

See(1) and Figure 1.

			MIN	MAX	UNIT
f_{SCL}	Serial clock frequency	Standard mode		100	kHz
		Fast mode		400	kHz
		Fast mode +		1	MHz
		High-speed mode, C_b = 100 pF		3.4	MHz
		High-speed mode, C_b = 400 pF		1.7	MHz
t_{LOW}	SCL low time	Standard mode	4.7		µs
		Fast mode	1.3		
		Fast mode +	0.5		
		High-speed mode, C_b = 100 pF	160		ns
		High-speed mode, C_b = 400 pF	320		
t_{HIGH}	SCL high time	Standard mode	4		µs
		Fast mode	0.6		
		Fast mode +	0.26		
		High-speed mode, C_b = 100 pF	60		ns
		High-speed mode, C_b = 400 pF	120		
$t_{SU;DAT}$	Data setup time	Standard mode	250		ns
		Fast mode	100		
		Fast mode +	50		
		High-speed mode	10		
$t_{HD;DAT}$	Data hold time	Standard mode	0	3.45	µs
		Fast mode	0	0.9	
		Fast mode +	0		
		High-speed mode, C_b = 100 pF	0	70	ns
		High-speed mode, C_b = 400 pF	0	150	

(1) C_b refers to the capacitance of one bus line. C_b is expressed in pF units.

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I²C Serial Bus Timing Parameter (continued)

See(1) and Figure 1.

			MIN	MAX	UNIT
$t_{SU;STA}$	Setup time for a start or a repeated start condition	Standard mode	4.7		µs
		Fast mode	0.6		
		Fast mode +	0.26		
		High-speed mode	160		ns
$t_{HD;STA}$	Hold time for a start or a repeated start condition	Standard mode	4		µs
		Fast mode	0.6		
		Fast mode +	0.26		
		High-speed mode	160		ns
t_{BUF}	Bus free time between a stop and start condition	Standard mode	4.7		µs
		Fast mode	1.3		
		Fast mode +	0.5		
$t_{SU;STO}$	Setup time for a stop condition	Standard mode	4		µs
		Fast mode	0.6		
		Fast mode +	0.26		
		High-speed mode	160		ns
t_{rDA}	Rise time of SDA signal	Standard mode		1000	ns
		Fast mode		300	
		Fast mode +		120	
		High-speed mode, C_b = 100 pF		80	
		High-speed mode, C_b = 400 pF		160	
t_{fDA}	Fall time of SDA signal	Standard mode		250	ns
		Fast mode		250	
		Fast mode +		120	
		High-speed mode, C_b = 100 pF		80	
		High-speed mode, C_b = 400 pF		160	
t_{rCL}	Rise time of SCL signal	Standard mode		1000	ns
		Fast mode		300	
		Fast mode +		120	
		High-speed Mode, C_b = 100 pF		40	
		High-speed Mode, C_b = 400 pF		80	
t_{rCL1}	Rise time of SCL signal after a repeated start condition and after an acknowledge bit	Standard mode		1000	ns
		Fast mode		300	
		Fast mode +		120	
		High-speed mode, C_b = 100 pF		80	
		High-speed mode, C_b = 400 pF		160	
t_{fCL}	Fall time of a SCL signal	Standard mode		300	ns
		Fast mode		300	
		Fast mode +		120	
		High-speed mode, C_b = 100 pF		40	
		High-speed mode, C_b = 400 pF		80	
C_b	Capacitive load for each bus line (SCL and SDA)			400	pF
t_{SP}	Pulse width of spike suppressed in SCL and SDA lines (spikes that are less than the indicated width are suppressed)	Fast mode, fast mode +		50	ns
		High-speed mode		10	

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7.7 Switching Characteristics

Limits apply over the junction temperature range –40°C ≤ T_J ≤ 125°C, specified $V_{(VANA)}$, V_{IN} , $V_{(NRST)}$, V_{OUT} and I_{OUT} range, unless otherwise noted. Typical values are at T_J = 25°C, f_{SW} = 3 MHz, $V_{(VANA)}$ = V_{IN} = 3.7 V and V_{OUT} = 1 V unless otherwise noted.(1)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
f_{SW}	Switching frequency, PWM mode		2.7	3	3.3	MHz
f_{SW-MAX}	Maximum switching frequency, PWM mode Automatically limited to smaller of f_{SW} and f_{SW-MAX}	V_{OUT} ≥ 0.6 V	2.7	3	3.3	MHz
		V_{OUT} < 0.6 V	1.8	2	2.2	
	Regulator start-up time (soft start)	From ENx to V_{OUT} = 0.225 V (slew-rate control begins), C_{OUT_TOTAL} = 88 µF, no load		90		µs
	Output voltage slew-rate(2)	SLEW_RATEx[2:0] = 000, V_{OUT} ≥ 0.5 V	–15%	30	15%	mV/µs
		SLEW_RATEx[2:0] = 001, V_{OUT} ≥ 0.5 V	–15%	15	15%	
		SLEW_RATEx[2:0] = 010, V_{OUT} ≥ 0.5 V	–15%	10	15%	
		SLEW_RATEx[2:0] = 011, V_{OUT} ≥ 0.5 V	–15%	7.5	15%	
		SLEW_RATEx[2:0] = 100, V_{OUT} ≥ 0.5 V	–15%	3.8	15%	
		SLEW_RATEx[2:0] = 101, V_{OUT} ≥ 0.5 V	–15%	1.9	15%	
		SLEW_RATEx[2:0] = 110, V_{OUT} ≥ 0.5 V	–15%	0.94	15%	
		SLEW_RATEx[2:0] = 111, V_{OUT} ≥ 0.5 V	–15%	0.4	15%	
	Load current measurement time	PFM mode (automatically changing to PWM mode for the measurement)		50		µs
		PWM mode		4		

(1) Minimum (MIN) and maximum (MAX) limits are specified by design, test, or statistical analysis. Typical (TYP) numbers are not verified, but do represent the most likely norm.
(2) The slew-rate can be limited by the current limit (forward or negative current limit), output capacitance and load current.

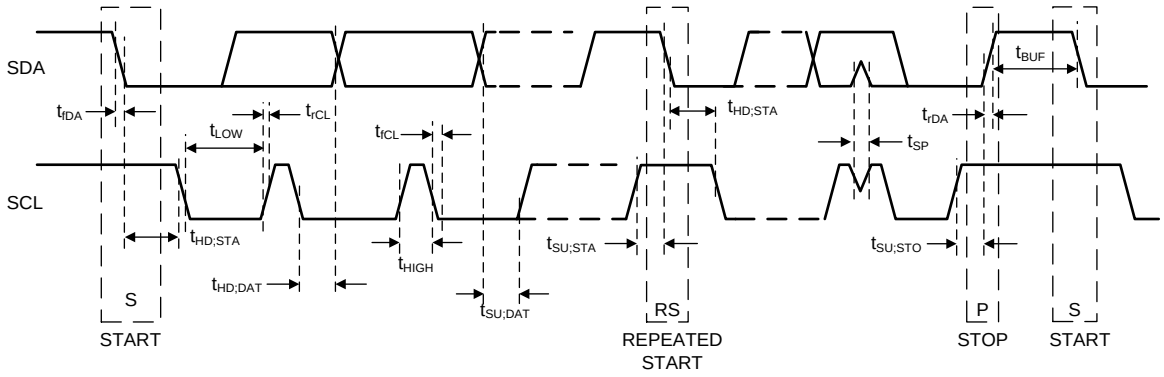
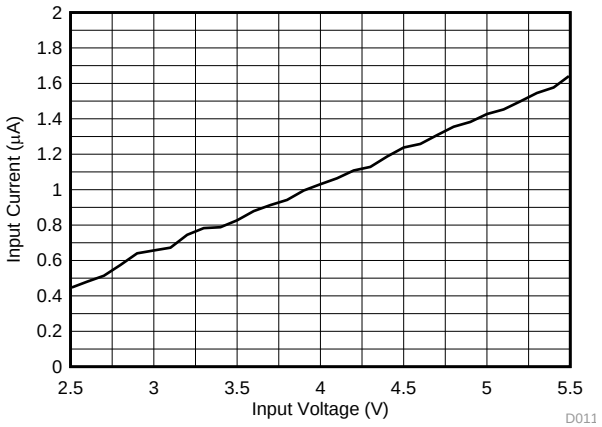


Figure 1. I²C Timing

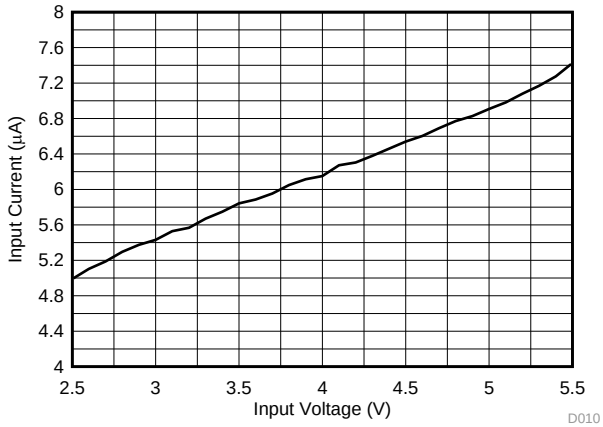
7.8 Typical Characteristics

Unless otherwise specified: T_A = 25°C, V_{IN} = 3.7 V, f_{SW} = 3 MHz.



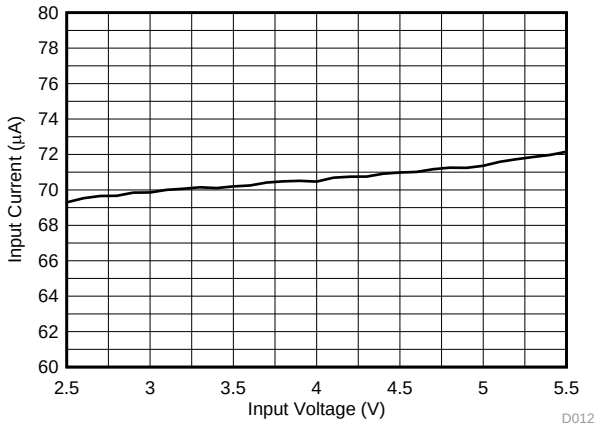
$V_{(NRST)}$ = 0 V

Figure 2. Shutdown Current Consumption vs Input Voltage



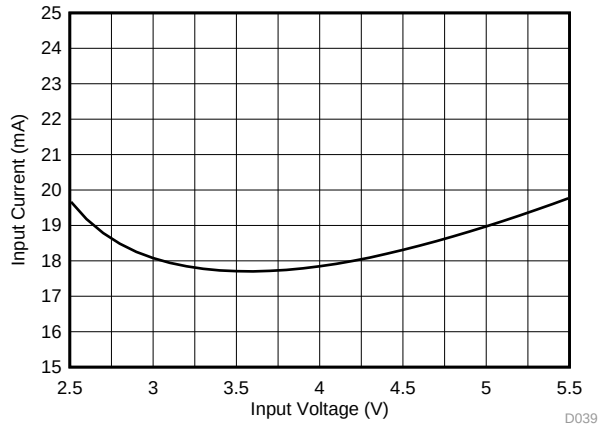
$V_{(NRST)}$ = 1.8 V Regulator disabled

Figure 3. Standby Current Consumption vs Input Voltage



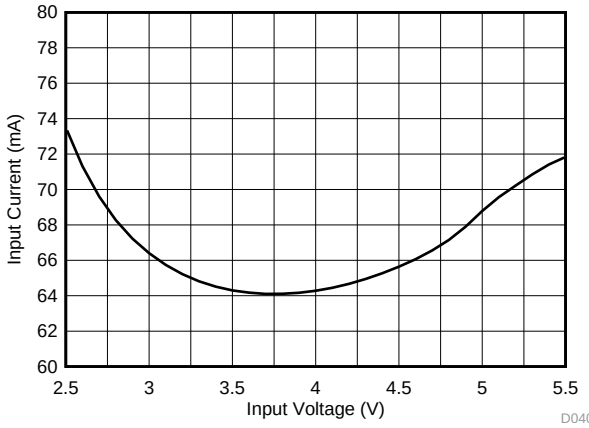
$V_{(NRST)}$ = 1.8 V $V_{(EN1)}$ = 1.8 V V_{OUT} = 1 V
Load = 0 mA L = 330 nH

Figure 4. PFM Mode Current Consumption vs Input Voltage



$V_{(NRST)}$ = 1.8 V $V_{(EN1)}$ = 1.8 V V_{OUT} = 1 V
Load = 0 mA 1 phase active L = 330 nH

Figure 5. PWM Mode Current Consumption vs Input Voltage



$V_{(NRST)}$ = 1.8 V $V_{(EN1)}$ = 1.8 V V_{OUT} = 1 V
Load = 0 mA 4 phases active L = 330 nH

Figure 6. Forced Multi-Phase Current Consumption vs Input Voltage

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8 Detailed Description

8.1 Overview

The LP8758 is a high-efficiency, high-performance power supply device with four step-down DC-DC converter cores. The cores are configured for a single 4-phase configuration. The device delivers 0.5-V to 3.36-V regulated voltage rail from 2.5-V to 5.5-V battery or supply voltage to portable devices such as cell phones, tablets, and PDAs.

There are two modes of operation for the converter, depending on the output current required: pulse-width modulation (PWM) and pulse-frequency modulation (PFM). The converter operates in PWM mode at high load currents of approximately 400 mA or higher. When operating in PWM mode the phases are automatically added/shedded based on the load current level. Lighter output current loads will cause the converter to automatically switch into PFM mode for reduced current consumption and a longer battery life when forced PWM mode is disabled. The forced multi-phase mode can be enabled for highest transient performance.

Additional features include soft-start, undervoltage lockout, overload protection, thermal warning, and thermal shutdown.

8.1.1 Buck Information

The LP8758 has four integrated high-efficiency buck converter cores. The cores are designed for flexibility; most of the functions are programmable, thus giving a possibility to optimize the regulator operation for each application.

8.1.1.1 Operating Modes

- OFF: Output is isolated from the input voltage rail in this mode. Output has an optional pulldown resistor.
- PWM: Converter operates in buck configuration with fixed switching frequency.
- PFM: Converter switches only when output voltage decreases below programmed threshold. Inductor current is discontinuous.

8.1.1.2 Features

- Output voltage
- Forced PWM operation
- Forced multi-phase operation (forces also the PWM operation)
- Switch current limit
- Output voltage slew rate
- Enable and disable delays

8.1.1.3 Programmability

The following parameters can be programmed via registers:

- DVS support with programmable slew-rate
- Automatic mode control based on the loading
- Synchronous rectification
- Current mode loop with PI compensator
- Optional spread spectrum technique to reduce EMI
- Soft start
- Power good flag with maskable interrupt
- Phase control for optimized EMI
- Average output current sensing (for PFM entry, phase shedding/adding, and load current measurement)
- Current balancing between the phases of the converter
- Differential voltage sensing from point of the load
- Dynamic phase shedding/adding, each output being phase shifted

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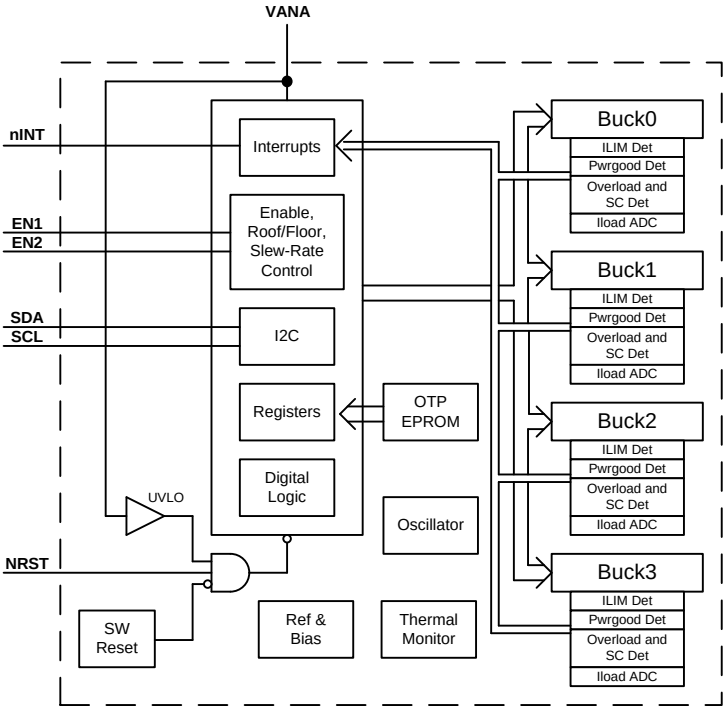
8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Multi-Phase DC-DC Converters

8.3.1.1 Overview

A multi-phase synchronous buck converter offers several advantages over a single power stage converter. For application processor power delivery, lower ripple on the input and output currents and faster transient response to load steps are the most significant advantages. Also, since the load current is evenly shared among multiple channels, the heat generated is greatly reduced for each channel due to the fact that power loss is proportional to square of current. Physical size of the output inductor shrinks significantly due to this heat reduction. A block diagram of a single core is shown in Figure 7.

Interleaving switching action of the converters and channels in a four-phase configuration is illustrated in Figure 8.

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Feature Description (continued)

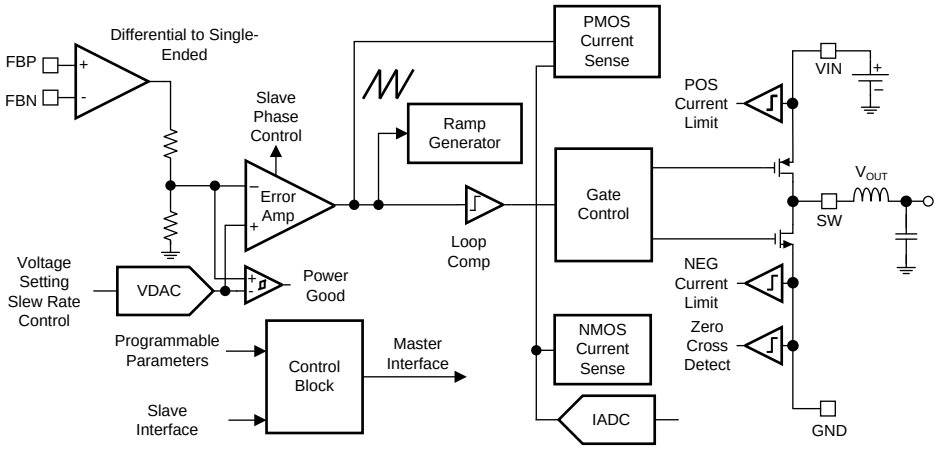

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Figure 7. Detailed Block Diagram Showing One Core

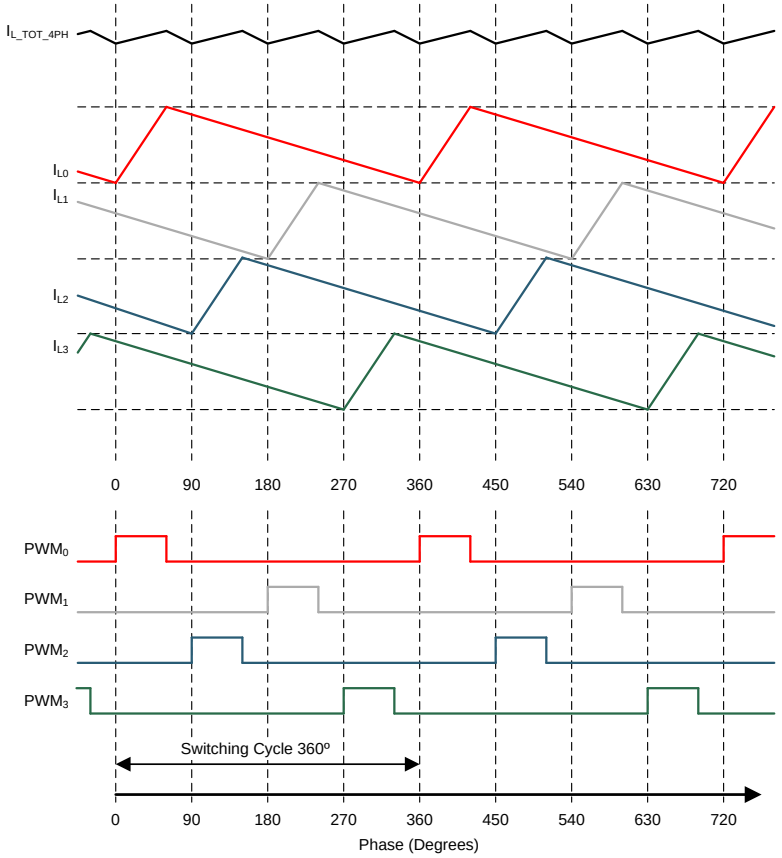

Figure 8. PWM Timings and Inductor Current Waveforms in 4-phase Configuration (1)

(1) Graph is not in scale and is for illustrative purposes only.

Feature Description (continued)

8.3.1.2 Multi-Phase Operation and Phase Adding/Shedding

Under heavy load conditions, the 4-phase converter switches each channel 90° apart. As a result, the 4-phase converter has an effective ripple frequency four times greater than the switching frequency of any one phase. However, the parallel operation decreases the efficiency at light load conditions. In order to overcome this operational inefficiency, the LP8758 can change the number of active phases to optimize efficiency for the variations of the load. This is called phase adding/shedding. The concept is illustrated below in Figure 9.

The converter can be forced to multi-phase operation by the BUCK0_CTRL1.BUCK0_FPWM_MP bit. If the regulator operates in forced multi-phase mode the forced PWM operation is automatically used. If the multi-phase operation is not forced, the number of phases are added and shedded automatically to follow the required output current.

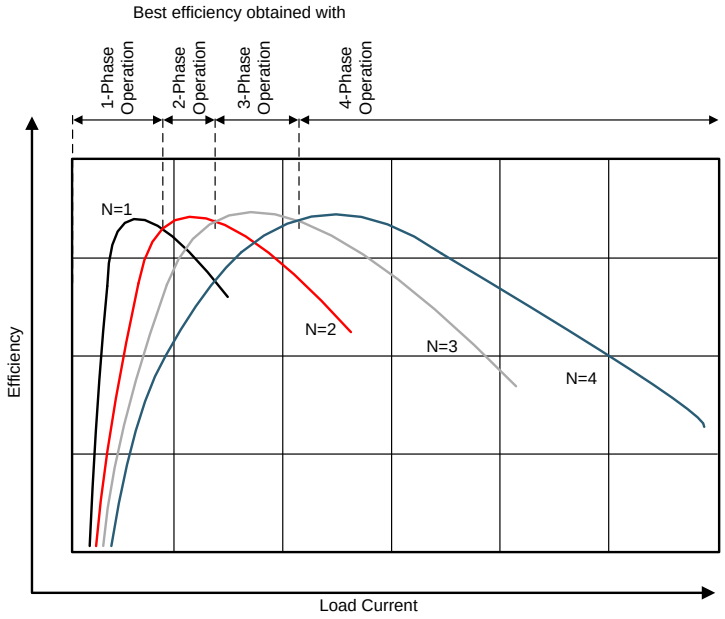

Figure 9. Multi-Phase Buck Converter Efficiency vs Number of Phases - All Converters in PWM Mode (2)

8.3.1.3 Transition Between PWM and PFM Modes

Normal PWM mode operation with phase-adding or phase-shedding optimizes efficiency at mid-to-full load at the expense of light-load efficiency. The LP8758 converter operates in PWM mode at load current of about 400 mA or higher. At lighter load current levels the device automatically switches into PFM mode for reduced current consumption when Forced PWM mode is disabled (AUTO mode operation). By combining the PFM and the PWM modes a high efficiency is achieved over a wide output-load current range.

8.3.1.4 Multi-Phase Switcher Configurations

In the multi-phase configuration the control of the multi-phase regulator settings is done using the control registers of the master buck. The following slave registers are ignored:

- BUCKx_CTRL1
- BUCKx_CTRL2, except ILIMx[2:0] bits
- interrupt bits related to the slave buck, except BUCKx_ILIM_INT

(2) Graph is not in scale and is for illustrative purposes only.

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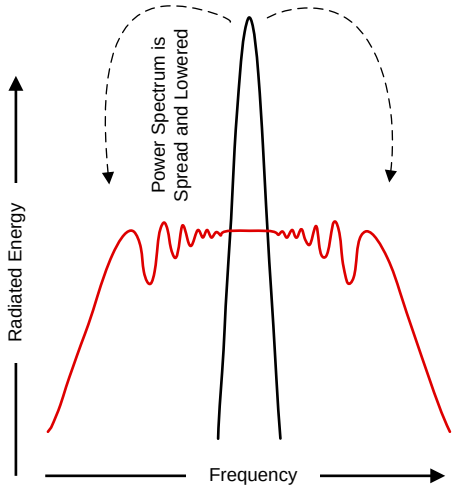
Feature Description (continued)

8.3.1.5 Buck Converter Load Current Measurement

Buck load current can be monitored via I²C registers. The monitored buck converter is selected with the SEL_I_LOAD.LOAD_CURRENT_BUCK_SELECT[1:0] register bits. A write to this selection register starts a current measurement sequence. The measurement sequence is typically 50 µs long. The LP8758 device can be configured to give out an interrupt INT_TOP.I_LOAD_READY after the load current measurement sequence is finished. Load current measurement interrupt can be masked with TOP_MASK.I_LOAD_READY_MASK bit. The measurement result can be read from registers I_LOAD_1 and I_LOAD_2. Register I_LOAD_1 bits BUCK_LOAD_CURRENT[7:0] give out the LSB bits and register I_LOAD_2 bits BUCK_LOAD_CURRENT[9:8] the MSB bits. The measurement result BUCK_LOAD_CURRENT[9:0] LSB is 20 mA, and maximum value of the measurement is 20.46 A. The measured current is the total value of the master and slave phases.

8.3.1.6 Spread-Spectrum Mode

Systems with periodic switching signals may generate a large amount of switching noise in a set of narrowband frequencies (the switching frequency and its harmonics). The usual solution to reduce noise coupling is to add EMI-filters and shields to the boards. The LP8758's register selectable spread-spectrum mode minimizes the need for output filters, ferrite beads, or chokes. In spread spectrum mode, the switching frequency varies randomly by ±5% (depending on selected switching frequency) about the center frequency, reducing the EMI emissions radiated by the converter and associated passive components and PCB traces (see Figure 10). This feature is enabled with the CONFIG.EN_SPREAD_SPEC bit, and it affects all the buck cores.



Where a fixed frequency converter exhibits large amounts of spectral energy at the switching frequency, the spread spectrum architecture of the LP8758 spreads that energy over a large bandwidth.

Figure 10. Spread-Spectrum Modulation

8.3.2 Power-Up

The power-up sequence for the LP8758 is as follows:

- VANA (and VIN_Bx) reach min recommended levels ($V_{(VANA)}$ > $VANA_{UVLO}$).
- NRST is set to high level. This initiates power-on-reset (POR), OTP reading and enables the system I/O interface. The I²C host allows at least 1.2 ms before writing or reading data to the LP8758.
- Device enters STANDBY mode.
- The host can change the default register setting by I²C if needed.
- The regulator can be enabled/disabled by ENx pin(s) and by I²C interface.

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Feature Description (continued)

8.3.3 Regulator Control

8.3.3.1 Enabling and Disabling Regulator

The regulator can be enabled when the device is in STANDBY state. There are two ways for enable and disable the regulator:

- Using BUCK0_CTRL1.EN_BUCK0 register bit (BUCK0_CTRL1.EN_PIN_CTRL0 register bit is '0').
- Using EN1/2 control pins (BUCK0_CTRL1.EN_BUCK0 register bit is '1' *AND* BUCK0_CTRL1.EN_PIN_CTRL0 register bit is '1').

If the EN1/2 control pins are used for enable and disable then the delay from the control signal rising edge to startup is set by BUCK0_DELAY.BUCK0_STARTUP_DELAY[3:0] bits and the delay from control signal falling edge to shutdown is set by BUCK0_DELAY.BUCK0_SHUTDOWN_DELAY[3:0] bits. The delays are valid only for EN1/2 signal and not for control with BUCK0_CTRL1.EN_BUCK0 bit. The delay time implemented by EN1/2 has overall +/-10% timing accuracy.

The control of the regulator (with 0 ms delays) is shown in Table 1. The multi-phase regulator is controlled with registers of the master phase.

Table 1. Regulator Control

CONTROL METHOD	ROW	EN_BUCKx0	BUCK0_CTRL1 EN_PIN_CTRL0	BUCK0_CTRL1 EN_PIN_SELECT0	BUCK0_CTRL1 EN_ROOF_FLOOR0	EN1 PIN	EN2 PIN	BUCK0 OUTPUT VOLTAGE
Enable/disable control with EN_BUCK0 bit	1	0	Don't Care	Don't Care	Don't Care	Don't Care	Don't Care	Disabled
	2	1	0	Don't Care	Don't Care	Don't Care	Don't Care	BUCK0_VOUT.BUCK0_VSET[7:0]
Enable/disable control with EN1 pin	3	1	1	0	0	Low	Don't Care	Disabled
	4	1	1	0	0	High	Don't Care	BUCK0_VOUT.BUCK0_VSET[7:0]
Enable/disable control with EN2 pin	5	1	1	1	0	Don't Care	Low	Disabled
	6	1	1	1	0	Don't Care	High	BUCK0_VOUT.BUCK0_VSET[7:0]
Roof/floor control with EN1 pin	7	1	1	0	1	Low	Don't Care	BUCK0_FLOOR_VOUT.BUCK0_FLOOR_VSET[7:0]
	8	1	1	0	1	High	Don't Care	BUCK0_VOUT.BUCK0_VSET[7:0]
Roof/floor control with EN2 pin	9	1	1	1	1	Don't Care	Low	BUCK0_FLOOR_VOUT.BUCK0_FLOOR_VSET[7:0]
	10	1	1	1	1	Don't Care	High	BUCK0_VOUT.BUCK0_VSET[7:0]

The following configuration allows the enable/disable control using ENx pin:

- BUCK0_CTRL1.EN_BUCK0 = 1
- BUCK0_CTRL1.EN_PIN_CTRL0 = 1
- BUCK0_CTRL1.EN_ROOF_FLOOR0 = 0
- BUCK0_VOUT.BUCK0_VSET[7:0] = Required voltage when ENx is high
- The enable pin for control is selected with BUCK0_CTRL1.EN_PIN_SELECT0

When the ENx pin is low, Table 1 row 3 (or 5) is valid, and the regulator is disabled. By setting ENx pin high, Table 1 row 4 (or 6) is valid, and the regulator is enabled with required voltage.

If the regulator is enabled all the time, and the ENx pin controls selection between two voltage level, the following configuration is used:

- BUCK0_CTRL1.EN_BUCK0 = 1
- BUCK0_CTRL1.EN_PIN_CTRL0 = 1
- BUCK0_CTRL1.EN_ROOF_FLOOR0 = 1
- BUCK0_VOUT.BUCK0_VSET[7:0] = Required voltage when ENx is high
- The enable pin for control is selected with BUCK0_CTRL1.EN_PIN_SELECT0

When the ENx pin is low, Table 1 row 7(or 9) is valid, and the regulator is enabled with a voltage defined by BUCK0_FLOOR_VOUT.BUCK0_FLOOR_VSET[7:0] bits. Setting the ENx pin high, Table 1 row 8 (or 10) is valid, and the regulator is enabled with a voltage defined by BUCK0_VOUT.BUCK0_VSET[7:0] bits.

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If the regulator is controlled by I²C writings, the BUCK0_CTRL1.EN_PIN_CTRL0 bit is set to 0. The enable/disable is controlled by the BUCK0_CTRL1.EN_BUCK0 bit, and when the regulator is enabled, the output voltage is defined by the BUCK0_VOUT.BUCK0_VSET[7:0] bits. The Table 1 rows 1 and 2 are valid for I²C controlled operation (ENx pins are ignored).

The regulator is enabled by the ENx pin or by I²C writing as shown in Figure 11. The soft-start circuit limits the in-rush current during start-up. Output voltage increase rate is around 30 mV/µsec during soft-start. When the output voltage rises to approximately 0.3 V, the output voltage becomes slew-rate controlled. If there is a short circuit at the output, and the output voltage does not increase above a 0.35-V level in 1 ms, the regulator is disabled, and interrupt is set. When the output voltage reaches the powergood threshold level the INT_BUCK_0_1.BUCK0_PG_INT interrupt flag is set. The powergood interrupt flag can be masked using BUCK_0_1_MASK.BUCK0_PG_MASK bit.

The ENx input pins have integrated pull-down resistors. The pull-down resistors are enabled by default and host can disable those with CONFIG.ENx_PD bits.

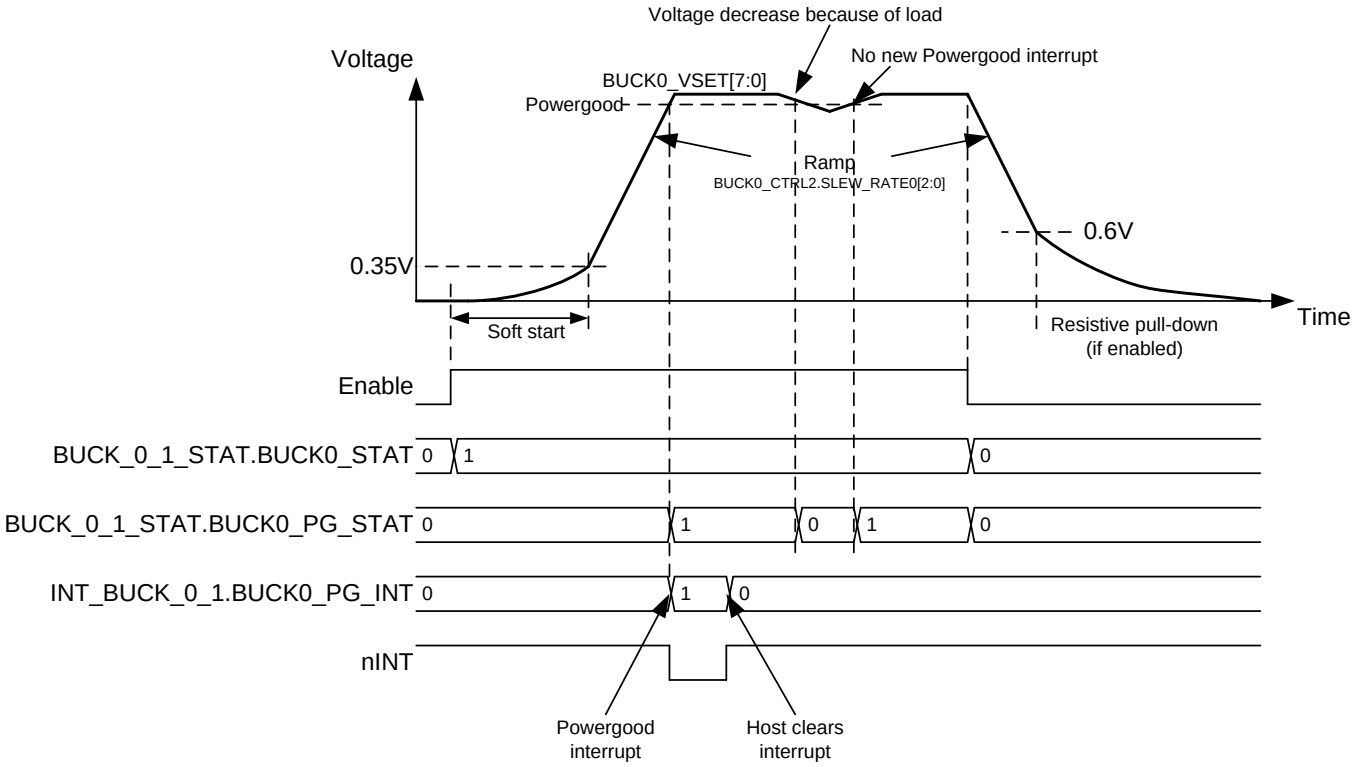

Figure 11. Regulator Enable and Disable

8.3.3.2 Changing Output Voltage

The regulator's output voltage can be changed by the ENx pin (voltage levels defined by the BUCK0_VOUT and BUCK0_FLOOR_VOUT registers) or by writing to the BUCK0_VOUT and BUCK0_FLOOR_VOUT registers. The voltage change is always slew-rate controlled, and the slew-rate is defined by the BUCKx_CTRL2.SLEW_RATE[2:0] bits. During voltage change the Forced PWM mode is used automatically. If the multi-phase operation is forced by the BUCK0_CTRL1. BUCK0_FPWM_MP bit, the regulator operates in multi-phase mode (four phases active). If the multi-phase operation is not forced, the number of phases are added and shedded automatically to follow the required slew rate. When the programmed output voltage is achieved, the mode becomes the one defined by load current, and the BUCK0_CTRL1.BUCK0_FPWM and BUCK0_CTRL1.BUCK0_FPWM_MP bits.

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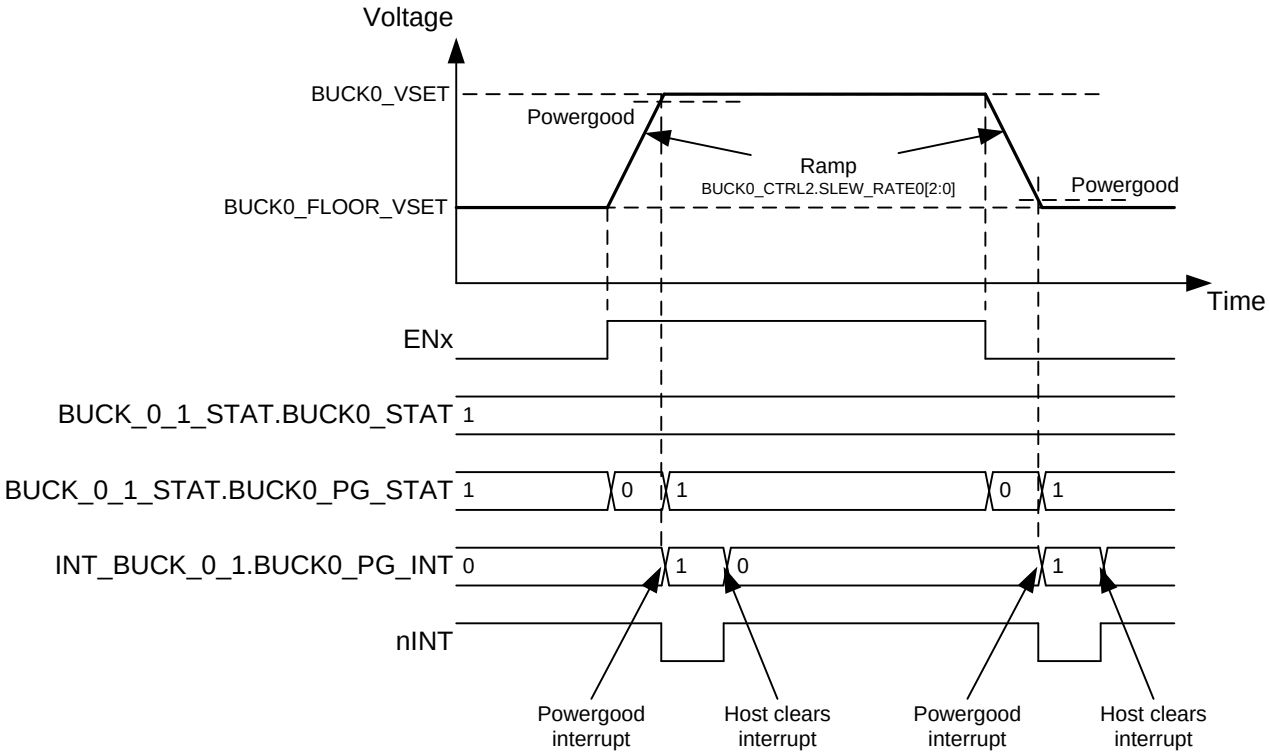

Figure 12. Regulator Output Voltage Change

8.3.4 Device Reset Scenarios

There are three reset methods implemented on the LP8758:

- Software reset with RESET.SW_RESET register bit
- Reset from low logic level of NRST signal
- Undervoltage lockout (UVLO) reset from VANA supply

An SW reset occurs when RESET.SW_RESET bit is written '1'. The bit is automatically cleared after writing. This event disables the regulator immediately, resets all the register bits to the default values and OTP bits are loaded (see Figure 14). I^2C interface is not reset during software reset.

If VANA supply voltage falls below UVLO threshold level or NRST signal is set low, then the regulator is disabled immediately, and all the register bits are reset to the default values. When the VANA supply voltage is above UVLO threshold level and NRST signal rises above threshold level an internal power-on reset (POR) occurs. OTP bits are loaded to the registers, and a start-up is initiated according to the register settings.

8.3.5 Diagnosis and Protection Features

The LP8758 is capable of providing three levels of protection features:

- Warnings for diagnosis which sets interrupt;
- Protection events which are disabling the regulator; and
- Faults which are causing the device to shutdown.

When the device detects warning/protection condition(s), the LP8758 sets the flag bits indicating what protection or warning conditions have occurred, and the nINT pin will be pulled low. nINT will be released again after a clear of flags is complete. The nINT signal stays low until all the pending interrupts are cleared.

When a fault is detected, it is indicated by a INT_TOP.RESET_REG interrupt flag after next start-up.

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Table 2. Summary of Interrupt Signals

EVENT	RESULT	INTERRUPT REGISTER AND BIT	INTERRUPT MASK	STATUS BIT	RECOVERY / INTERRUPT CLEAR
Current limit triggered (20 µs debounce)	No effect	INT_TOP.INT_BUCKx = 1 INT_BUCKx.BUCKx_ILIM_INT = 1	BUCKx_MASK.BUCKx_ILIM_MASK	BUCKx_STAT.BUCKx_ILIM_STAT	Write 1 to INT_BUCKx.BUCKx_ILIM_INT bit Interrupt is not cleared if current limit is active
Short circuit (V_{OUT} < 0.35 V at 1 ms after enable) or Overload (V_{OUT} decreasing below 0.35V during operation, 1 ms debounce)	Regulator disable	INT_TOP.INT_BUCK0 = 1 INT_BUCK_0_1.BUCK0_SC_INT = 1	N/A	N/A	Write 1 to INT_BUCK_0_1.BUCK0_SC_INT bit
Thermal Warning	No effect	INT_TOP.TDIE_WARN = 1	TOP_MASK.TDIE_WARN_MASK	TOP_STAT.TDIE_WARN_STAT	Write 1 to INT_TOP.TDIE_WARN bit Interrupt is not cleared if temperature is above thermal warning level
Thermal Shutdown	Regulator disabled	INT_TOP.TDIE_SD = 1	N/A	TOP_STAT.TDIE_SD_STAT	Write 1 to INT_TOP.TDIE_SD bit Interrupt is not cleared if temperature is above thermal shutdown level
Powergood, output voltage reaches the programmed value	No effect	INT_TOP.INT_BUCK0 = 1 INT_BUCK_0_1.BUCK0_PG_INT = 1	BUCK_0_1_MASK.BUCK0_PG_MASK	BUCK_0_1_STAT.BUCK0_PG_STAT	Write 1 to INT_BUCK_0_1.BUCK0_PG_INT bit
Load current measurement ready	No effect	INT_TOP.I_LOAD_READY = 1	TOP_MASK.I_LOAD_READY_MASK	N/A	Write 1 to INT_TOP.I_LOAD_READY bit
Start-up (NRST rising edge)	Device ready for operation, registers reset to default values	INT_TOP.RESET_REG = 1	TOP_MASK.RESET_REG_MASK	N/A	Write 1 to INT_TOP.RESET_REG bit
Glitch on supply voltage and UVLO triggered (VANA falling and rising)	Immediate shutdown followed by powerup, registers reset to default values	INT_TOP.RESET_REG = 1	TOP_MASK.RESET_REG_MASK	N/A	Write 1 to INT_TOP.RESET_REG bit
Software requested reset	Immediate shutdown followed by powerup, registers reset to default values	INT_TOP.RESET_REG = 1	TOP_MASK.RESET_REG_MASK	N/A	Write 1 to INT_TOP.RESET_REG bit

8.3.5.1 Warnings for Diagnosis (Interrupt)

8.3.5.1.1 Output Current Limit

The buck regulators have programmable output peak current limits. The limits are individually programmed for all buck regulators with BUCKx_CTRL2.ILIMx[2:0] bits. The current limit settings of master and slave regulators used for the same output voltage rail must be identical. If the load current is increased so that the current limit is triggered, the regulator continues to regulate to the limit current level (current peak regulation). The voltage may decrease if the load current is higher than limit current. If the current regulation continues for 20 µs, the LP8758 device sets the INT_BUCKx.BUCKx_ILIM_INT bit and pulls the nINT pin low. The host processor can read BUCKx_STAT.BUCKx_ILIM_STAT bits to see if the regulator is still in peak current regulation mode.

If the load is so high that the output voltage decreases below a 350-mV level, the LP8758 device disables the regulator and sets the INT_BUCK_0_1.BUCK0_SC_INT bit. In addition the BUCK_0_1_STAT.BUCK0_STAT bit is set to 0. The interrupt is cleared when the host processor writes 1 to INT_BUCK_0_1.BUCK0_SC_INT bit. The overload situation is shown in Figure 13.

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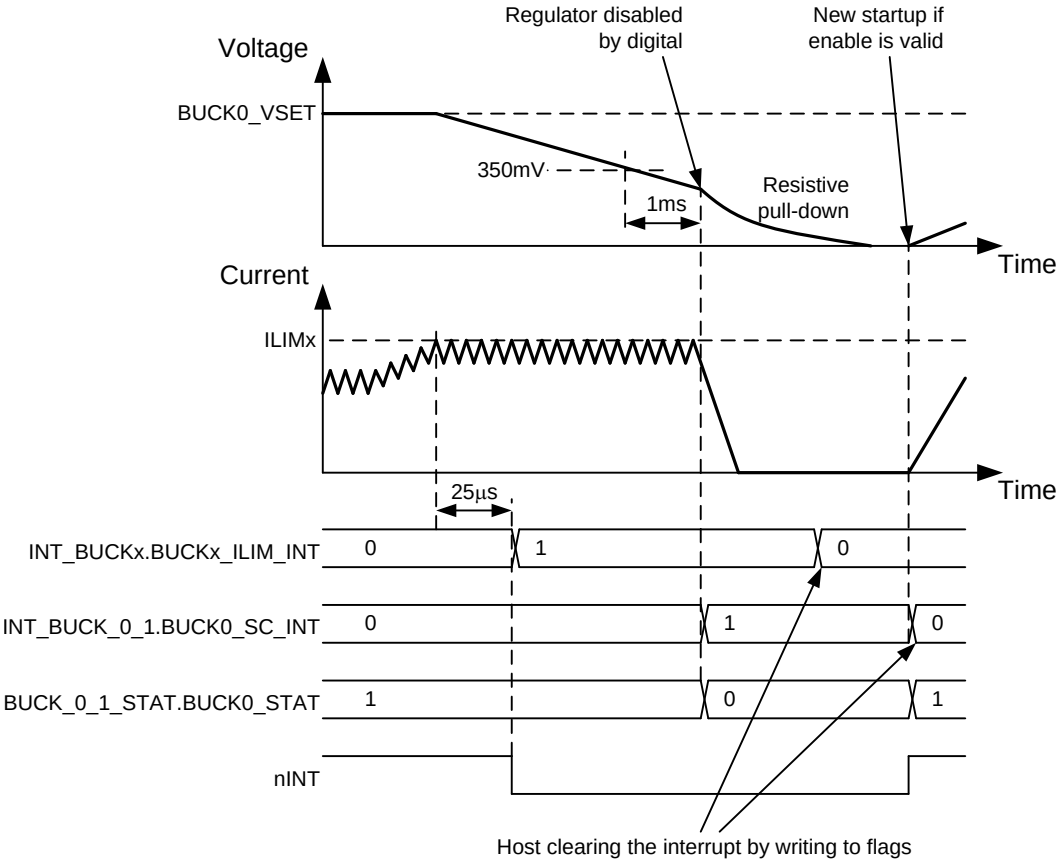

Figure 13. Overload Situation

8.3.5.1.2 Thermal Warning

The LP8758 device includes protection feature against over-temperature by setting an interrupt for host processor. The threshold level of the thermal warning is selected with CONFIG.TDIE_WARN_LEVEL bit.

If the LP8758 device temperature increases above thermal warning level the device sets INT_TOP.TDIE_WARN bit and pulls nINT pin low. The status of the thermal warning can be read from TOP_STAT.TDIE_WARN_STAT bit and the interrupt is cleared by writing 1 to INT_TOP.TDIE_WARN bit.

8.3.5.2 Protection (Regulator Disable)

If the regulator is disabled because of protection or fault (short-circuit protection, overload protection, thermal shutdown, or undervoltage lockout), the output power FETs are set to high-impedance mode, and the output pull-down resistor is enabled (if enabled with BUCKx_CTRL1.EN_RDISx bits). The turn-off time of the output voltage is defined by the output capacitance, load current, and the resistance of the integrated pulldown resistor.

8.3.5.2.1 Short-Circuit and Overload Protection

A short-circuit protection feature allows the LP8758 to protect itself and external components against short circuit at the output or against overload during start-up. The fault threshold is 350 mV, and the protection is triggered, and the regulator disabled, if the output voltage is below the threshold level 1 ms after the regulator is enabled.

In a similar way the overload situation is protected during normal operation. If the regulator's feedback-pin voltage falls below 0.35 V, and remains below the threshold level for 1 ms, the regulator is disabled.

In the short-circuit and overload situations the INT_BUCK_0_1.BUCK0_SC_INT and the INT_TOP.INT_BUCK0 bits are set to 1, the BUCK_0_1_STAT.BUCK0_STAT bit is set to 0 and the nINT signal is pulled low. The host processor clears the interrupt by writing 1 to the INT_BUCK_0_1.BUCK0_SC_INT bit. Upon clearing the interrupt the regulator makes a new start-up attempt if the enable register bits and/or ENx control signal is valid.

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8.3.5.2.2 Thermal Shutdown

The LP8758 has an overtemperature protection function that operates to protect itself from short-term misuse and overload conditions. When the junction temperature exceeds around 150°C, the regulator is disabled, the INT_TOP.TDIE_SD bit is set to 1, the nINT signal is pulled low, and the device enters STANDBY. nINT will be cleared by writing 1 to the INT_TOP.TDIE_SD bit. If the temperature is above thermal shutdown level the interrupt is not cleared. The host can read the status of the thermal shutdown from the TOP_STAT.TDIE_SD_STAT bit. Regulator cannot be enabled as long as the junction temperature is above thermal shutdown level or the thermal shutdown interrupt is pending.

8.3.5.3 Fault (Power Down)

8.3.5.3.1 Undervoltage Lockout

When the input voltage falls below $VANA_{UVLO}$ at the VANA pin, the buck converters are disabled immediately, and the output capacitor is discharged using the pulldown resistor and the LP8758 device enters SHUTDOWN. When VANA voltage is above UVLO threshold level and NRST signal is high, the device powers up to STANDBY state.

If the reset interrupt is unmasked by default (TOP_MASK.RESET_REG_MASK = 0) the INT_TOP.RESET_REG interrupt indicates that the device has been in SHUTDOWN. The host processor must clear the interrupt by writing 1 to the INT_TOP.RESET_REG bit. If the host processor reads the INT_TOP.RESET_REG flag after detecting an nINT low signal, it knows that the input supply voltage has been below UVLO level (or the host has requested reset), and the registers are reset to default values.

8.3.6 Digital Signal Filtering

The digital signals have a debounce filtering. The signal/supply is sampled with a clock signal and a counter. This results as an accuracy of one clock period for the debounce window.

Table 3. Digital Signal Filtering

EVENT	SIGNAL / SUPPLY	RISING EDGE LENGTH	FALLING EDGE LENGTH
Enable/Disable/Voltage Select for BUCK0	EN1	3µs (1)	3µs (1)
Enable/Disable/Voltage Select for BUCK0	EN2	3µs (1)	3µs (1)
VANA undervoltage lockout	VANA	Immediate	Immediate
Thermal warning	TDIE_WARN	20 µs	20 µs
Thermal shutdown	TDIE_SD	20 µs	20 µs
Current limit	VOUTx_ILIM	20 µs	20 µs
Overload	FB_B0 - FB_B1, FB_B2 - FB_F3	1 ms	1 ms
Powergood	FB_B0 - FB_B1, FB_B2 - FB_F3	20 µs	20 µs

(1) No glitch filtering, only synchronization.

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8.4 Device Functional Modes

8.4.1 Modes of Operation

SHUTDOWN: The $V_{(NRST)}$ voltage is below threshold level. All switch, reference, control and bias circuitry of the LP8758 device are turned off.

WAIT-ON: The $V_{(NRST)}$ voltage is above threshold level. The reference and bias circuitry are enabled. The regulator of the LP8758 device is turned off.

READ OTP: The main supply voltage $V_{(VANA)}$ is above $VANA_{UVLO}$ level and $V_{(NRST)}$ voltage is above threshold level. The regulator is disabled and the reference and bias circuitry of the LP8758 are enabled. The OTP bits are loaded to registers.

STANDBY: The main supply voltage $V_{(VANA)}$ is above $VANA_{UVLO}$ level and $V_{(NRST)}$ voltage is above threshold level. The regulator is disabled and the reference, control and bias circuitry of the LP8758 are enabled. All registers can be read or written by the host processor via the system serial interface. The regulator can be enabled if needed.

ACTIVE: The main supply voltage $V_{(VANA)}$ is above $VANA_{UVLO}$ level and $V_{(NRST)}$ voltage is above threshold level. At least one regulated DC-DC converter is enabled. All registers can be read or written by the host processor via the system serial interface.

The operating modes and transitions between the modes are shown in Figure 14.

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Device Functional Modes (continued)

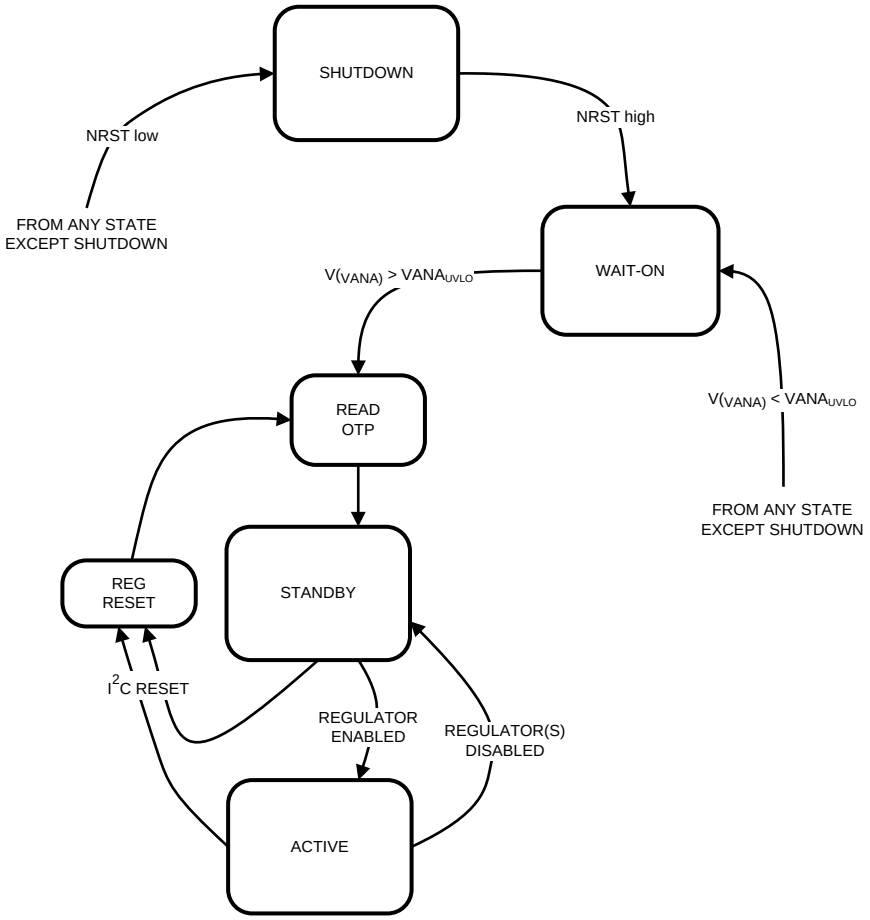

Figure 14. Device Operation Modes

8.5 Programming

8.5.1 I²C-Compatible Interface

The I²C-compatible synchronous serial interface provides access to the programmable functions and registers on the device. This protocol uses a two-wire interface for bidirectional communications between the devices connected to the bus. The two interface lines are the Serial Data Line (SDA), and the Serial Clock Line (SCL). Every device on the bus is assigned a unique address and acts as either a Master or a Slave depending on whether it generates or receives the serial clock SCL. The SCL and SDA lines should each have a pullup resistor placed somewhere on the line and remain HIGH even when the bus is idle. The LP8758 supports standard mode (100 kHz), fast mode (400 kHz), fast mode plus (1 MHz), and high-speed mode (3.4 MHz).

8.5.1.1 Data Validity

The data on the SDA line must be stable during the HIGH period of the clock signal (SCL). In other words, the state of the data line can only be changed when clock signal is LOW.

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Programming (continued)

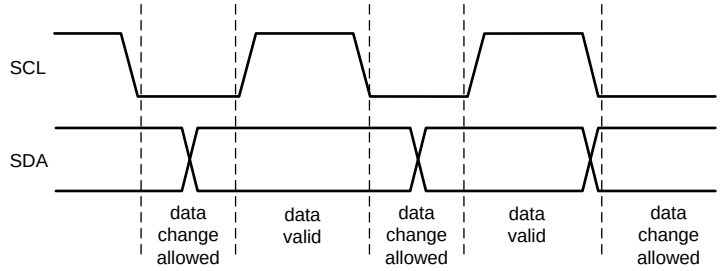

Figure 15. Data Validity Diagram

8.5.1.2 Start and Stop Conditions

The LP8758 is controlled via an I²C-compatible interface. START and STOP conditions classify the beginning and end of the I²C session. A START condition is defined as SDA transitions from HIGH to LOW while SCL is HIGH. A STOP condition is defined as SDA transition from LOW to HIGH while SCL is HIGH. The I²C master always generates the START and STOP conditions.

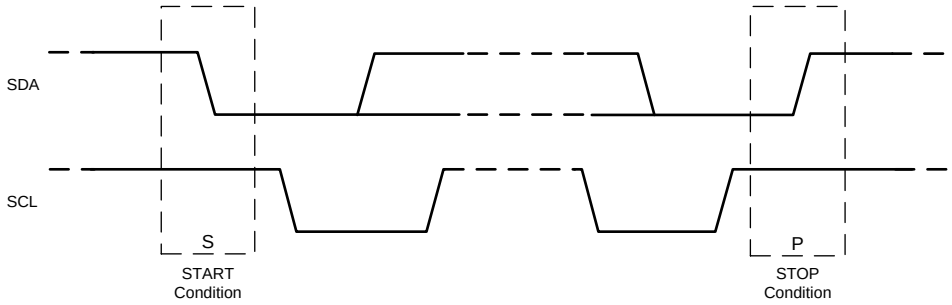

Figure 16. Start and Stop Sequences

The I²C bus is considered busy after a START condition and free after a STOP condition. During data transmission the I²C master can generate repeated START conditions. A START and a repeated START condition are equivalent function-wise. The data on SDA must be stable during the HIGH period of the clock signal (SCL). In other words, the state of SDA can only be changed when SCL is LOW. Figure 17 shows the SDA and SCL signal timing for the I²C-Compatible Bus. See the *I²C Serial Bus Timing Parameter* for timing values.

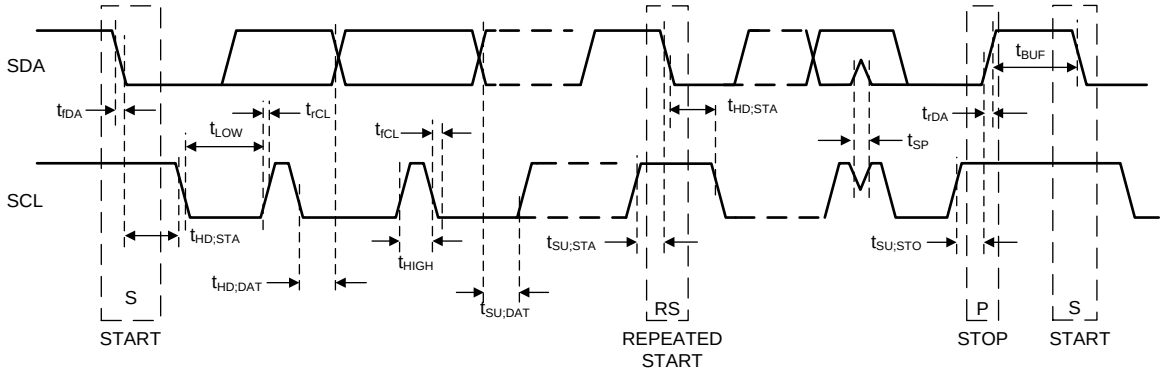

Figure 17. I²C-Compatible Timing

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Programming (continued)

8.5.1.3 Transferring Data

Every byte put on the SDA line must be eight bits long, with the most significant bit (MSB) being transferred first. Each byte of data has to be followed by an acknowledge bit. The acknowledge related clock pulse is generated by the master. The master releases the SDA line (HIGH) during the acknowledge clock pulse. The LP8758 pulls down the SDA line during the 9th clock pulse, signifying an acknowledge. The LP8758 generates an acknowledge after each byte has been received.

There is one exception to the “acknowledge after every byte” rule. When the master is the receiver, it must indicate to the transmitter an end of data by not-acknowledging (“negative acknowledge”) the last byte clocked out of the slave. This “negative acknowledge” still includes the acknowledge clock pulse (generated by the master), but the SDA line is not pulled down.

NOTE

If the NRST signal is low during I^2C communication the LP8758 device does not drive SDA line. The ACK signal and data transfer to the master is disabled at that time.

After the START condition, the bus master sends a chip address. This address is seven bits long followed by an eighth bit which is a data direction bit (READ or WRITE). For the eighth bit, a “0” indicates a WRITE and a “1” indicates a READ. The second byte selects the register to which the data will be written. The third byte contains data to write to the selected register.

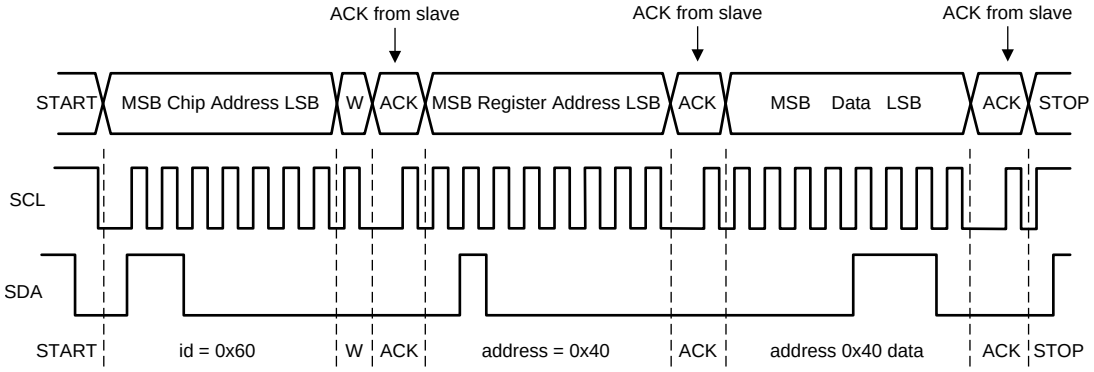
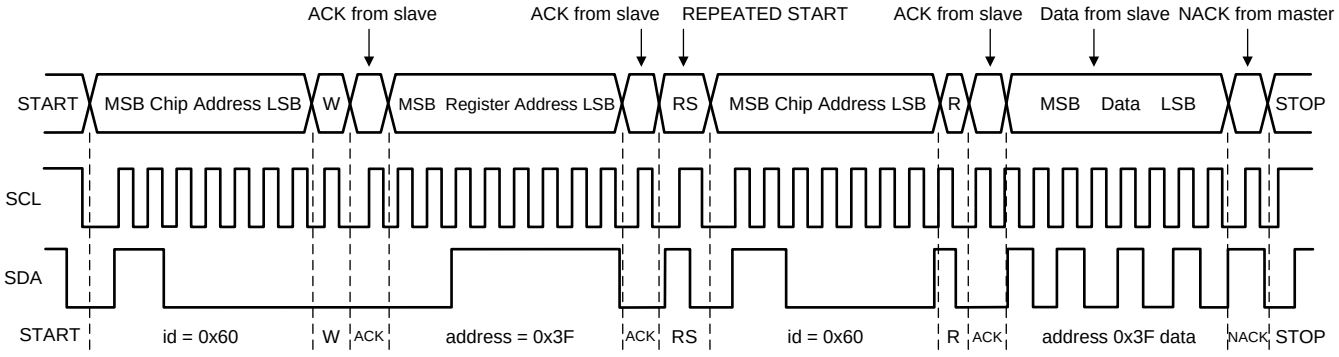


Figure 18. Write Cycle (w = write; SDA = '0'), id = Device Address = 60Hex for LP8758



When READ function is to be accomplished, a WRITE function must precede the READ function as shown above.

Figure 19. Read Cycle (r = read; SDA = '1'), id = Device Address = 60Hex for LP8758

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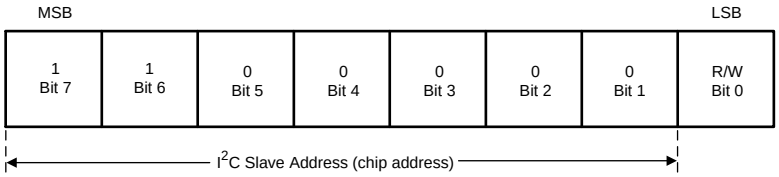
Programming (continued)

8.5.1.4 I²C-Compatible Chip Address

The device address for the LP8758 is 0x60. After the START condition, the I²C master sends the 7-bit address followed by an eighth bit, read or write (R/W). R/W = 0 indicates a WRITE and R/W = 1 indicates a READ. The second byte following the device address selects the register address to which the data will be written. The third byte contains the data for the selected register.



Here device address is 110 0000Bin = 60Hex.

Figure 20. Device Address

8.5.1.5 Auto Increment Feature

The auto-increment feature allows writing several consecutive registers within one transmission. Every time an 8-bit word is sent to the LP8758, the internal address index counter will be incremented by one and the next register will be written. Table 4 below shows writing sequence to two consecutive registers. Note that the auto-increment feature does not work for read.

Table 4. Auto-Increment Example

Master Action	Start	Device Address = 60H	Write		Register Address		Data		Data		Stop
LP8758 Action				ACK		ACK		ACK		ACK	

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8.6 Register Maps

8.6.1 Register Descriptions

The LP8758 is controlled by a set of registers through the system serial interface port. The device registers, their addresses and their abbreviations are listed in Table 5. A more detailed description is given in sections *DEV_REV* to *I_LOAD_1*.

Table 5. Summary of LP8758 Control Registers

Addr	Register	Read / Write	D7	D6	D5	D4	D3	D2	D1	D0
0x01	OTP_REV	R	OTP_ID[7:0]							
0x02	BUCK0_CTRL1	R/W	EN_BUCK0	EN_PIN_CTRL0	EN_PIN_SELECT0	EN_ROOF_FLOOR0	EN_RDIS0	Reserved	BUCK0_FPWM	BUCK0_FPWM_MP
0x03	BUCK0_CTRL2	R/W	Reserved		ILIM0[2:0]			SLEW_RATE0[2:0]		
0x05	BUCK1_CTRL2	R/W	Reserved		ILIM1[2:0]			Reserved		
0x07	BUCK2_CTRL2	R/W	Reserved		ILIM2[2:0]			Reserved		
0x09	BUCK3_CTRL2	R/W	Reserved		ILIM3[2:0]			Reserved		
0x0A	BUCK0_VOUT	R/W	BUCK0_VSET[7:0]							
0x0B	BUCK0_FLOOR_VOUT	R/W	BUCK0_FLOOR_VSET[7:0]							
0x12	BUCK0_DELAY	R/W	BUCK0_SHUTDOWN_DELAY[3:0]				BUCK0_STARTUP_DELAY[3:0]			
0x16	RESET	R/W	Reserved							SW_RESET
0x17	CONFIG	R/W	Reserved				TDIE_WARN_LEVEL	EN2_PD	EN1_PD	EN_SPREAD_SPEC
0x18	INT_TOP	R/W	INT_BUCK3	INT_BUCK2	INT_BUCK1	INT_BUCK0	TDIE_SD	TDIE_WARN	RESET_REG	I_LOAD_READY
0x19	INT_BUCK_0_1	R/W	Reserved			BUCK1_ILIM_INT	Reserved	BUCK0_PG_INT	BUCK0_SC_INT	BUCK0_ILIM_INT
0x1A	INT_BUCK_2_3	R/W	Reserved			BUCK3_ILIM_INT	Reserved			BUCK2_ILIM_INT
0x1B	TOP_STAT	R	Reserved				TDIE_SD_STAT	TDIE_WARN_STAT	Reserved	
0x1C	BUCK_0_1_STAT	R	Reserved			BUCK1_ILIM_STAT	BUCK0_STAT	BUCK0_PG_STAT	Reserved	BUCK0_ILIM_STAT
0x1D	BUCK_2_3_STAT	R	Reserved			BUCK3_ILIM_STAT	Reserved			BUCK2_ILIM_STAT
0x1E	TOP_MASK	R/W	Reserved					TDIE_WARN_MASK	RESET_REG_MASK	I_LOAD_READY_MASK
0x1F	BUCK_0_1_MASK	R/W	Reserved			BUCK1_ILIM_MASK	Reserved	BUCK0_PG_MASK	Reserved	BUCK0_ILIM_MASK
0x20	BUCK_2_3_MASK	R/W	Reserved			BUCK3_ILIM_MASK	Reserved			BUCK2_ILIM_MASK
0x21	SEL_I_LOAD	R/W	Reserved						LOAD_CURRENT_BUCK_SELECT[1:0]	
0x22	I_LOAD_2	R/W	Reserved						BUCK_LOAD_CURRENT[9:8]	
0x23	I_LOAD_1	R/W	BUCK_LOAD_CURRENT[7:0]							

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8.6.1.1 DEV_REV

Address: 0x00

D7	D6	D5	D4	D3	D2	D1	D0
DEVICE_ID[1:0]		ALL_LAYER[1:0]		METAL_LAYER[3:0]			

Bits	Field	Type	Default	Description
7:6	DEVICE_ID[1:0]	R	00	Device specific ID code.
5:4	ALL_LAYER[1:0]	R	00	Shows the all layer version of the device: 00 - First all layer version 01 - Second all layer version 10 - Third all layer version 11 - Fourth all layer version
3:0	METAL_LAYER [3:0]	R	0001	Shows the metal layer version of the device: 0000 - All layer version 0001 - First metal layer spin ... 1111 - 15th metal layer spin

8.6.1.2 OTP_REV

Address: 0x01

D7	D6	D5	D4	D3	D2	D1	D0
OTP_ID[7:0]							

Bits	Field	Type	Default	Description
7:0	OTP_ID[7:0]	R	1011 0000	Identification Code of the OTP EPROM Version.

8.6.1.3 BUCK0_CTRL1

Address: 0x02

D7	D6	D5	D4	D3	D2	D1	D0
EN_BUCK0	EN_PIN_ CTRL0	EN_PIN_ SELECT0	EN_ROOF_ FLOOR0	EN_RDIS0	Reserved	BUCK0_FPWM	BUCK0_FPWM _MP

Bits	Field	Type	Default	Description
7	EN_BUCK0	R/W	1	Enable BUCK0 regulator: 0 - BUCK0 regulator is disabled 1 - BUCK0 regulator is enabled.
6	EN_PIN_CTRL0	R/W	1	Enable EN1/2 pin control for BUCK0: 0 - only EN_BUCK0 bit controls BUCK0 1 - EN_BUCK0 bit AND EN1/2 pin control BUCK0.
5	EN_PIN_SELECT0	R/W	0	Select which ENx pin controls BUCK0 if EN_PIN_CTRL0 = 1: 0 - EN1 pin 1 - EN2 pin.
4	EN_ROOF_ FLOOR0	R/W	0	Enable Roof/Floor control of EN1/2 pin if EN_PIN_CTRL0 = 1: 0 - Enable/Disable (1/0) control 1 - Roof/Floor (1/0) control.
3	EN_RDIS0	R/W	1	Enable output discharge resistor when BUCK0 is disabled: 0 - Discharge resistor disabled 1 - Discharge resistor enabled.
2	Reserved	R/W	0	
1	BUCK0_FPWM	R/W	0	Forces the BUCK0 regulator to operate in PWM mode: 0 - Automatic transitions between PFM and PWM modes (AUTO mode). 1 - Forced to PWM operation.
0	BUCK0_FPWM _MP	R/W	0	Forces the BUCK0 regulator to operate always in multi-phase and forced PWM operation mode: 0 - Automatic phase adding and shedding. 1 - Forced to multi-phase operation, 2 phases in the 2-phase configuration, 3 phases in the 3-phase configuration and 4 phases in the 4-phase configuration.

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8.6.1.4 BUCK0_CTRL2

Address: 0x03

D7	D6	D5	D4	D3	D2	D1	D0
Reserved		ILIM0[2:0]			SLEW_RATE0[2:0]		

Bits	Field	Type	Default	Description
7:6	Reserved	R/W	00	
5:3	ILIM0[2:0]	R/W	111	Sets the switch current limit of BUCK0. Can be programmed at any time during operation: 000 - 1.5 A 001 - 2.0 A 010 - 2.5 A 011 - 3.0 A 100 - 3.5 A 101 - 4.0 A 110 - 4.5 A 111 - 5.0 A (Default)
2:0	SLEW_RATE0[2:0]	R/W	010	Sets the output voltage slew rate for BUCK0 regulator (rising and falling edges): 000 - 30 mV/µs 001 - 15 mV/µs 010 - 10 mV/µs (Default) 011 - 7.5 mV/µs 100 - 3.8 mV/µs 101 - 1.9 mV/µs 110 - 0.94 mV/µs 111 - 0.4 mV/µs

8.6.1.5 BUCK1_CTRL2

Address: 0x05

D7	D6	D5	D4	D3	D2	D1	D0
Reserved		ILIM1[2:0]			Reserved		

Bits	Field	Type	Default	Description
7:6	Reserved	R/W	00	
5:3	ILIM1[2:0]	R/W	111	Sets the switch current limit of BUCK1. Can be programmed at any time during operation: 000 - 1.5 A 001 - 2.0 A 010 - 2.5 A 011 - 3.0 A 100 - 3.5 A 101 - 4.0 A 110 - 4.5 A 111 - 5.0 A (Default)
2:0	Reserved	R/W	010	

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8.6.1.6 BUCK2_CTRL2

Address: 0x07

D7	D6	D5	D4	D3	D2	D1	D0
Reserved		ILIM2[2:0]			Reserved		

Bits	Field	Type	Default	Description
7:6	Reserved	R/W	00	
5:3	ILIM2[2:0]	R/W	111	Sets the switch current limit of BUCK2. Can be programmed at any time during operation: 000 - 1.5 A 001 - 2.0 A 010 - 2.5 A 011 - 3.0 A 100 - 3.5 A 101 - 4.0 A 110 - 4.5 A 111 - 5.0 A (Default)
2:0	Reserved	R/W	010	

8.6.1.7 BUCK3_CTRL2

Address: 0x09

D7	D6	D5	D4	D3	D2	D1	D0
Reserved		ILIM3[2:0]			Reserved		

Bits	Field	Type	Default	Description
7:6	Reserved	R/W	00	
5:3	ILIM3[2:0]	R/W	111	Sets the switch current limit of BUCK3. Can be programmed at any time during operation: 000 - 1.5 A 001 - 2.0 A 010 - 2.5 A 011 - 3.0 A 100 - 3.5 A 101 - 4.0 A 110 - 4.5 A 111 - 5.0 A (Default)
2:0	Reserved	R/W	010	

8.6.1.8 BUCK0_VOUT

Address: 0x0A

D7	D6	D5	D4	D3	D2	D1	D0
BUCK0_VSET[7:0]							

Bits	Field	Type	Default	Description
7:0	BUCK0_VSET[7:0]	R/W	0110 0001	Sets the output voltage of BUCK0 regulator **0.5 V - 0.73 V, 10 mV steps** 0000 0000 - 0.5V ... 0001 0111 - 0.73 V **0.73 V - 1.4 V, 5 mV steps** 0001 1000 - 0.735 V ... 1001 1101 - 1.4 V **1.4 V - 3.36 V, 20 mV steps** 1001 1110 - 1.42 V ... 1111 1111 - 3.36 V

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8.6.1.9 BUCK0_FLOOR_VOUT

Address: 0x0B

D7	D6	D5	D4	D3	D2	D1	D0
BUCK0_FLOOR_VSET[7:0]							

Bits	Field	Type	Default	Description
7:0	BUCK0_FLOOR_VSET[7:0]	R/W	0000 0000	Sets the output voltage of BUCK0 regulator when Floor state is used **0.5 V - 0.73 V, 10 mV steps** 0000 0000 - 0.5V ... 0001 0111 - 0.73 V **0.73 V - 1.4 V, 5 mV steps** 0001 1000 - 0.735 V ... 1001 1101 - 1.4 V **1.4 V - 3.36 V, 20 mV steps** 1001 1110 - 1.42 V ... 1111 1111 - 3.36 V

8.6.1.10 BUCK0_DELAY

Address: 0x12

D7	D6	D5	D4	D3	D2	D1	D0
BUCK0_SHUTDOWN_DELAY[3:0]				BUCK0_STARTUP_DELAY[3:0]			

Bits	Field	Type	Default	Description
7:4	BUCK0_SHUTDOWN_DELAY[3:0]	R/W	0000	Shutdown delay of BUCK0 from falling edge of ENx signal: 0000 - 0 ms 0001 - 1 ms ... 1111 - 15 ms
3:0	BUCK0_STARTUP_DELAY[3:0]	R/W	0000	Start-up delay of BUCK0 from rising edge of ENx signal: 0000 - 0 ms 0001 - 1 ms ... 1111 - 15 ms

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8.6.1.11 RESET

Address: 0x16

D7	D6	D5	D4	D3	D2	D1	D0
Reserved							SW_RESET

Bits	Field	Type	Default	Description
7:1	Reserved	R/W	000 0000	
0	SW_RESET	R/W	0	Software commanded reset. When written to 1, the registers will be reset to default values, OTP memory is read, and the I^2C interface is reset. The bit is automatically cleared.

8.6.1.12 CONFIG

Address: 0x17

D7	D6	D5	D4	D3	D2	D1	D0
Reserved				TDIE_WARN_ LEVEL	EN2_PD	EN1_PD	EN_SPREAD _SPEC

Bits	Field	Type	Default	Description
7:4	Reserved	R/W	0000	
3	TDIE_WARN_ LEVEL	R/W	0	Thermal warning threshold level. 0 - 125°C 1 - 105°C.
2	EN2_PD	R/W	1	Selects the pull down resistor on the EN2 input pin. 0 - Pull-down resistor is disabled. 1 - Pull-down resistor is enabled.
1	EN1_PD	R/W	1	Selects the pull down resistor on the EN1 input pin. 0 - Pull-down resistor is disabled. 1 - Pull-down resistor is enabled.
0	EN_SPREAD _SPEC	R/W	0	Enable spread spectrum feature: 0 - Disabled 1 - Enabled

8.6.1.13 INT_TOP

Address: 0x18

D7	D6	D5	D4	D3	D2	D1	D0
INT_BUCK3	INT_BUCK2	INT_BUCK1	INT_BUCK0	TDIE_SD	TDIE_WARN	RESET_REG	I_LOAD_ READY

Bits	Field	Type	Default	Description
7	INT_BUCK3	R	0	Interrupt indicating that output BUCK3 has a pending interrupt. The reason for the interrupt is indicated in INT_BUCK3 register. This bit is cleared automatically when INT_BUCK3 register is cleared to 0x00.
6	INT_BUCK2	R	0	Interrupt indicating that output BUCK2 has a pending interrupt. The reason for the interrupt is indicated in INT_BUCK2 register. This bit is cleared automatically when INT_BUCK2 register is cleared to 0x00.
5	INT_BUCK1	R	0	Interrupt indicating that output BUCK1 has a pending interrupt. The reason for the interrupt is indicated in INT_BUCK1 register. This bit is cleared automatically when INT_BUCK1 register is cleared to 0x00.
4	INT_BUCK0	R	0	Interrupt indicating that output BUCK0 has a pending interrupt. The reason for the interrupt is indicated in INT_BUCK0 register. This bit is cleared automatically when INT_BUCK0 register is cleared to 0x00.
3	TDIE_SD	R/W	0	Latched status bit indicating that the die junction temperature has exceeded the thermal shutdown level. The regulator has been disabled if it was enabled. The regulator cannot be enabled if this bit is active. The actual status of the thermal warning is indicated by TOP_STAT.TDIE_SD_STAT bit. Write 1 to clear interrupt.

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Bits	Field	Type	Default	Description
2	TDIE_WARN	R/W	0	Latched status bit indicating that the die junction temperature has exceeded the thermal warning level. The actual status of the thermal warning is indicated by TOP_STAT.TDIE_WARN_STAT bit. Write 1 to clear interrupt.
1	RESET_REG	R/W	0	Latched status bit indicating that either startup (NRST rising edge) has done, VANA supply voltage has been below undervoltage threshold level or the host has requested a reset (RESET.SW_RESET). The regulator has been disabled, and registers are reset to default values and the normal startup procedure is done. Write 1 to clear interrupt.
0	I_LOAD_READY	R/W	0	Latched status bit indicating that the load current measurement result is available in I_LOAD_1 and I_LOAD_2 registers. Write 1 to clear interrupt.

8.6.1.14 INT_BUCK_0_1

Address: 0x19

D7	D6	D5	D4	D3	D2	D1	D0
Reserved			BUCK1_ILIM_INT	Reserved	BUCK0_PG_INT	BUCK0_SC_INT	BUCK0_ILIM_INT

Bits	Field	Type	Default	Description
7:5	Reserved	R/W	000	
4	BUCK1_ILIM_INT	R/W	0	Latched status bit indicating that output current limit has been active. Write 1 to clear.
3	Reserved	R/W	0	
2	BUCK0_PG_INT	R/W	0	Latched status bit indicating that BUCK0 output voltage has reached powergood threshold level. Write 1 to clear.
1	BUCK0_SC_INT	R/W	0	Latched status bit indicating that the BUCK0 output voltage has fallen below 0.35V level during operation or BUCK0 output didn't reach 0.35 V level in 1 ms from enable. Write 1 to clear.
0	BUCK0_ILIM_INT	R/W	0	Latched status bit indicating that output current limit has been active. Write 1 to clear.

8.6.1.15 INT_BUCK_2_3

Address: 0x1A

D7	D6	D5	D4	D3	D2	D1	D0
Reserved			BUCK3_ILIM_INT	Reserved			BUCK2_ILIM_INT

Bits	Field	Type	Default	Description
7:5	Reserved	R/W	000	
4	BUCK3_ILIM_INT	R/W	0	Latched status bit indicating that output current limit has been active. Write 1 to clear.
3:1	Reserved	R/W	000	
0	BUCK2_ILIM_INT	R/W	0	Latched status bit indicating that output current limit has been active. Write 1 to clear.

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8.6.1.16 TOP_STAT

Address: 0x1B

D7	D6	D5	D4	D3	D2	D1	D0
Reserved				TDIE_SD_STAT	TDIE_WARN_STAT	Reserved	

Bits	Field	Type	Default	Description
7:4	Reserved	R	0000	
3	TDIE_SD_STAT	R	0	Status bit indicating the status of thermal shutdown: 0 - Die temperature below thermal shutdown level 1 - Die temperature above thermal shutdown level.
2	TDIE_WARN_STAT	R	0	Status bit indicating the status of thermal warning: 0 - Die temperature below thermal warning level 1 - Die temperature above thermal warning level.
1:0	Reserved	R	00	

8.6.1.17 BUCK_0_1_STAT

Address: 0x1C

D7	D6	D5	D4	D3	D2	D1	D0
Reserved			BUCK1_ILIM_STAT	BUCK0_STAT	BUCK0_PG_STAT	Reserved	BUCK0_ILIM_STAT

Bits	Field	Type	Default	Description
7:5	Reserved	R	000	
4	BUCK1_ILIM_STAT	R	0	Status bit indicating BUCK1 current limit status (raw status) 0 - BUCK1 output current is below current limit level 1 - BUCK1 output current limit is active.
3	BUCK0_STAT	R	0	Status bit indicating the enable/disable status of BUCK0: 0 - BUCK0 regulator is disabled 1 - BUCK0 regulator is enabled.
2	BUCK0_PG_STAT	R	0	Status bit indicating BUCK0 output voltage validity (raw status) 0 - BUCK0 output is above powergood threshold level 1 - BUCK0 output is below powergood threshold level.
1	Reserved	R	0	
0	BUCK0_ILIM_STAT	R	0	Status bit indicating BUCK0 current limit status (raw status) 0 - BUCK0 output current is below current limit level 1 - BUCK0 output current limit is active.

8.6.1.18 BUCK_2_3_STAT

Address: 0x1D

D7	D6	D5	D4	D3	D2	D1	D0
Reserved			BUCK3_ILIM_STAT	Reserved			BUCK2_ILIM_STAT

Bits	Field	Type	Default	Description
7:5	Reserved	R	000	
4	BUCK3_ILIM_STAT	R	0	Status bit indicating BUCK3 current limit status (raw status) 0 - BUCK3 output current is below current limit level 1 - BUCK3 output current limit is active.
3:1	Reserved	R	000	
0	BUCK2_ILIM_STAT	R	0	Status bit indicating BUCK2 current limit status (raw status) 0 - BUCK2 output current is below current limit level 1 - BUCK2 output current limit is active.

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8.6.1.19 TOP_MASK

Address: 0x1E

D7	D6	D5	D4	D3	D2	D1	D0
Reserved					TDIE_WARN _MASK	RESET_REG _MASK	I_LOAD_ READY_MASK

Bits	Field	Type	Default	Description
7:3	Reserved	R/W	0 0000	
2	TDIE_WARN _MASK	R/W	0	Masking for thermal warning interrupt INT_TOP.TDIE_WARN: 0 - Interrupt generated 1 - Interrupt not generated. This bit does not affect TOP_STAT.TDIE_WARN_STAT status bit.
1	RESET_REG _MASK	R/W	1	Masking for register reset interrupt INT_TOP.RESET_REG: 0 - Interrupt generated 1 - Interrupt not generated.
0	I_LOAD_ READY_MASK	R/W	0	Masking for load current measurement ready interrupt INT_TOP.I_LOAD_READY. 0 - Interrupt generated 1 - Interrupt not generated.

8.6.1.20 BUCK_0_1_MASK

Address: 0x1F

D7	D6	D5	D4	D3	D2	D1	D0
Reserved			BUCK1_ILIM _MASK	Reserved	BUCK0_PG _MASK	Reserved	BUCK0_ILIM _MASK

Bits	Field	Type	Default	Description
7:5	Reserved	R/W	000	
4	BUCK1_ILIM _MASK	R/W	1	Masking for BUCK1 current limit detection interrupt INT_BUCK_0_1.BUCK1_ILIM_INT: 0 - Interrupt generated 1 - Interrupt not generated. This bit does not affect BUCK_0_1_STAT.BUCK1_ILIM_STAT status bit.
3	Reserved	R/W	0	
2	BUCK0_PG_MASK	R/W	0	Masking for BUCK0 power good interrupt INT_BUCK_0_1.BUCK0_PG_INT: 0 - Interrupt generated 1 - Interrupt not generated. This bit does not affect BUCK_0_1_STAT.BUCK1_PG_STAT status bit.
1	Reserved	R	0	
0	BUCK0_ILIM _MASK	R/W	0	Masking for BUCK0 current limit detection interrupt INT_BUCK_0_1.BUCK0_ILIM_INT: 0 - Interrupt generated 1 - Interrupt not generated. This bit does not affect BUCK_0_1_STAT.BUCK1_ILIM_STAT status bit.

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8.6.1.21 BUCK_2_3_MASK

Address: 0x20

D7	D6	D5	D4	D3	D2	D1	D0
Reserved			BUCK3_ILIM_MASK	Reserved			BUCK2_ILIM_MASK

Bits	Field	Type	Default	Description
7:5	Reserved	R/W	000	
4	BUCK3_ILIM_MASK	R/W	1	Masking for BUCK3 current limit detection interrupt INT_BUCK_2_3.BUCK3_ILIM_INT: 0 - Interrupt generated 1 - Interrupt not generated. This bit does not affect BUCK_2_3_STAT.BUCK3_ILIM_STAT status bit.
3:1	Reserved	R/W	000	
0	BUCK2_ILIM_MASK	R/W	1	Masking for BUCK2 current limit detection interrupt INT_BUCK_2_3.BUCK2_ILIM_INT: 0 - Interrupt generated 1 - Interrupt not generated. This bit does not affect BUCK_2_3_STAT.BUCK1_ILIM_STAT status bit.

8.6.1.22 SEL_I_LOAD

Address: 0x21

D7	D6	D5	D4	D3	D2	D1	D0
Reserved						LOAD_CURRENT_BUCK_SELECT[1:0]	

Bits	Field	Type	Default	Description
7:2	Reserved	R/W	00 0000	
1:0	LOAD_CURRENT_BUCK_SELECT [1:0]	R/W	00	Start the current measurement on the selected regulator: 00 - BUCK0 01 - BUCK1 10 - BUCK2 11 - BUCK3 The measurement is started when register is written. If the selected buck is master, the measurement result is a sum current of master and slave bucks. If the selected buck is slave, the measurement result is a current of the selected slave bucks.

8.6.1.23 I_LOAD_2

Address: 0x22

D7	D6	D5	D4	D3	D2	D1	D0
Reserved						BUCK_LOAD_CURRENT[9:8]	

Bits	Field	Type	Default	Description
7:2	Reserved	R	00 0000	
1:0	BUCK_LOAD_CURRENT[9:8]	R	00	This register describes 3 MSB bits of the average load current on selected regulator with a resolution of 20 mA per LSB and max 20 A current.

8.6.1.24 I_LOAD_1

Address: 0x23

D7	D6	D5	D4	D3	D2	D1	D0
BUCK_LOAD_CURRENT[7:0]							

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Bits	Field	Type	Default	Description
7:0	BUCK_LOAD_ CURRENT[7:0]	R	0000 0000	This register describes 8 LSB bits of the average load current on selected regulator with a resolution of 10 mA per LSB and max 20 A current.

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9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The LP8758 is a multi-phase step-down converter with four switcher cores bundled together.

9.2 Typical Application

4-Phase configuration

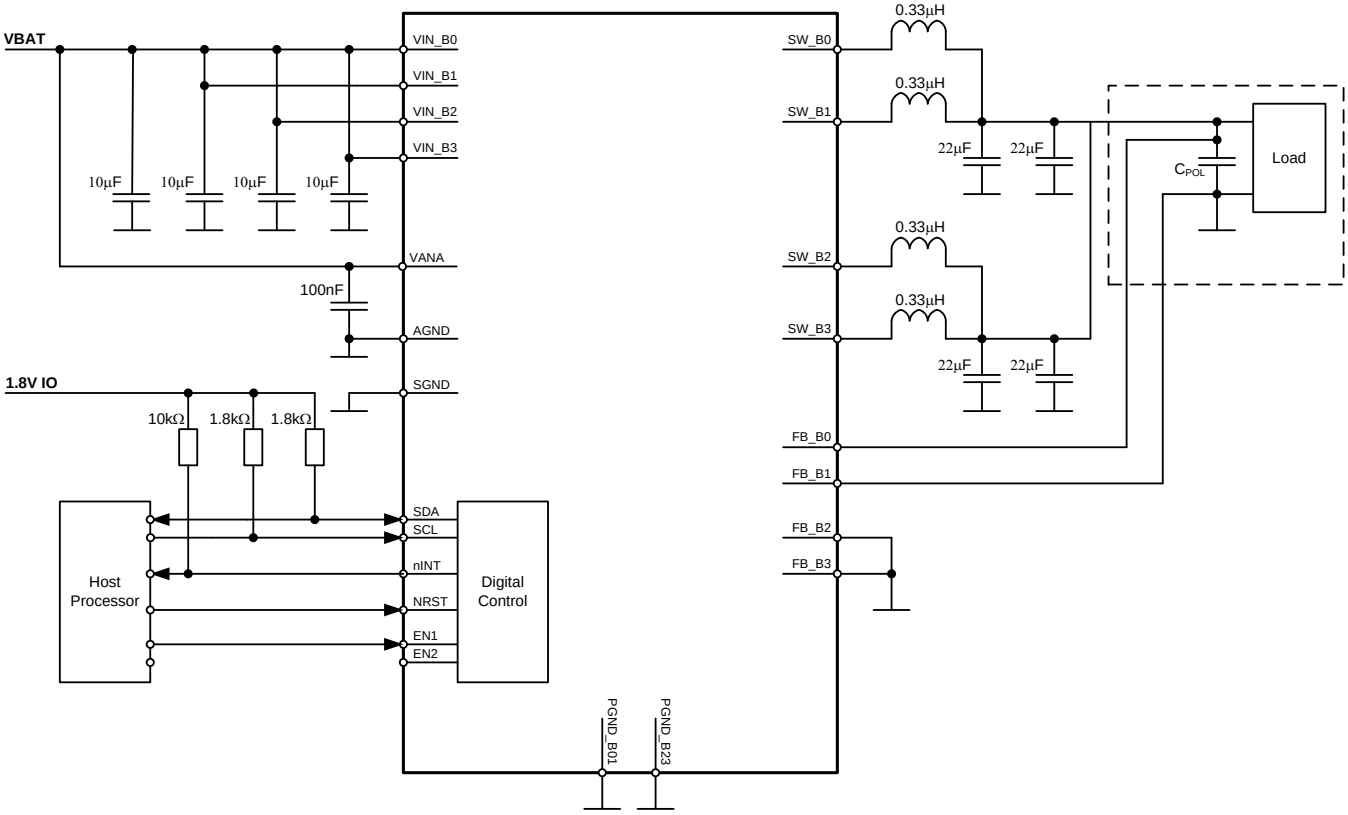

Figure 21. LP8758 Typical Application Circuit

9.2.1 Design Requirements

Table 6. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Input voltage range	2.5 V to 5.5 V
Output voltage	1.1 V
Converter operation mode	Auto mode (PWM-PFM)
Maximum load current	16 A
Inductor current limit	5 A

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9.2.2 Detailed Design Procedure

The performance of the LP8758 device depends greatly on the care taken in designing the printed circuit board (PCB). The use of low-inductance and low serial-resistance ceramic capacitors is strongly recommended, while proper grounding is crucial. Attention must be given to decoupling the power supplies. Decoupling capacitors must be connected close to the device and between the power and ground pins to support high peak currents being drawn from system power rail during turn-on of the switching MOSFETs. Keep input and output traces as short as possible, because trace inductance, resistance, and capacitance can easily become the performance limiting items. The separate power pins VIN_Bx are not connected together internally. The VIN_Bx power connections shall be connected together outside the package using power plane construction.

9.2.2.1 Application Components

9.2.2.1.1 Inductor Selection

DC bias current characteristics of inductors must be considered. Different manufacturers follow different saturation current rating specifications, so attention must be given to details. DC bias curves should be requested from them as part of the inductor selection process. Minimum effective value of inductance to ensure good performance is 0.22 µH at 4-A bias current over the inductor's operating temperature range. The inductor's DC resistance should be less than 0.05 Ω for good efficiency at high current condition. The inductor AC loss (resistance) also affects conversion efficiency. Higher Q factor at switching frequency usually gives better efficiency at light load to middle load. See Table 7. Shielded inductors are preferred as they radiate less noise.

Table 7. Recommended Inductors

MANUFACTURER	PART NUMBER	VALUE	DIMENSIONS L×W×H (mm)	DCR (mΩ)
TOKO	DFE252010F-R33M	0.33 µH	2.5 × 2.0 × 1.0	16 (typ), 21 (max)
TDK	VLS252010HBX-R33M	0.33 µH	2.5 × 2.0 × 1.0	25 (typ), 31 (max)
TDK	VLS252010HBX-R47M	0.47 µH	2.5 × 2.0 × 1.0	29 (typ), 35 (max)
TDK	TFM2016GHM-0R47M	0.47 µH	2.0 × 1.6 × 1.0	46 (max)
TOKO	DFE322512C R47	0.47 µH	3.2 × 2.5 × 1.2	21 (typ), 31 (max)

9.2.2.1.2 Input Capacitor Selection

A ceramic input capacitor of 10 µF, 6.3 V is sufficient for most applications. Place the power input capacitor as close as possible to the VIN_Bx pin and PGND_Bx pin of the device. A larger value or higher voltage rating may be used to improve input voltage filtering. Use X7R or X5R types, do not use Y5V or F. DC bias characteristics of ceramic capacitors must be considered when selecting case sizes like 0402. Minimum effective input capacitance to ensure good performance is 1.9 µF per buck input at maximum input voltage DC bias including tolerances and over ambient temp range, assuming that there are at least 22 µF of additional capacitance common for all the power input pins on the system power rail. See Table 8.

The input filter capacitor supplies current to the high-side FET switch in the first half of each cycle and reduces voltage ripple imposed on the input power source. A ceramic capacitor's low equivalent series resistance (ESR) provides the best noise filtering of the input voltage spikes due to this rapidly changing current. Select an input filter capacitor with sufficient ripple current rating.

The VANA input is used to supply analog and digital circuits in the device. See recommended components from Table 9 for VANA input supply filtering.

Table 8. Recommended Power Input Capacitors (X5R Dielectric)

MANUFACTURER	PART NUMBER	VALUE	CASE SIZE	DIMENSIONS L×W×H (mm)	VOLTAGE RATING
Murata	GRM188R60J106ME47	10 µF (20%)	0603	1.6 × 0.8 × 0.8	6.3 V

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Table 9. Recommended VANA Supply Filtering Components

MANUFACTURER	PART NUMBER	VALUE	CASE SIZE	DIMENSIONS L×W×H (mm)	VOLTAGE RATING
Samsung	CL03A104KP3NNNC	100 nF (10%)	0201	0.6 × 0.3 × 0.3	10 V
Murata	GRM033R61A104KE84	100 nF (10%)	0201	0.6 × 0.3 × 0.3	6.3 V

9.2.2.1.3 Output Capacitor Selection

Use ceramic capacitors, X7R or X5R types; do not use Y5V or F. DC bias voltage characteristics of ceramic capacitors must be considered. DC bias characteristics vary from manufacturer to manufacturer, and DC bias curves should be requested from them as part of the capacitor selection process. The output filter capacitor smooths out current flow from the inductor to the load, helps maintain a steady output voltage during transient load changes and reduces output voltage ripple. These capacitors must be selected with sufficient capacitance and sufficiently low ESR and ESL to perform these functions. Minimum effective output capacitance to ensure good performance is 10 µF per phase at the output voltage DC bias including tolerances and over ambient temp range.

The output voltage ripple is caused by the charging and discharging of the output capacitor and also due to its R_{ESR}. The R_{ESR} is frequency dependent (as well as temperature dependent); make sure the value used for selection process is at the switching frequency of the part. See Table 10.

A higher output capacitance improves the load step behavior and reduces the output voltage ripple as well as decreases the PFM switching frequency. For most 4-phase applications 4 x 22 µF 0603 capacitors for C_{OUT} are suitable. A point-of-load (POL) capacitance C_{POL} can be added with remove feedback as shown in Figure 21. Although a converter's loop compensation can be programmed to adapt to virtually several hundreds of microfarads C_{OUT}, it is preferable for C_{OUT} to be < 200 µF (4-phase configuration). Choosing higher than that is not necessarily of any benefit. Note that the output capacitor may be the limiting factor in the output voltage ramp, especially for very large (> 100 µF) output capacitors. For large output capacitors, the output voltage might be slower than the programmed ramp rate at voltage transitions, because of the higher energy stored on the output capacitance. Also at start-up, the time required to charge the output capacitor to target value might be longer. At shutdown, if the output capacitor is discharged by the internal discharge resistor, more time is required to settle V_{OUT} down as a consequence of the increased time constant.

Table 10. Recommended Output Capacitors (X5R Dielectric)

MANUFACTURER	PART NUMBER	VALUE	CASE SIZE	DIMENSIONS L×W×H (mm)	VOLTAGE RATING
Samsung	CL10A226MP8NUNE	22 µF (20%)	0603	1.6 × 0.8 × 0.8	10 V
Murata	GRM188R60J226MEA0	22 µF (20%)	0603	1.6 × 0.8 × 0.8	6.3 V

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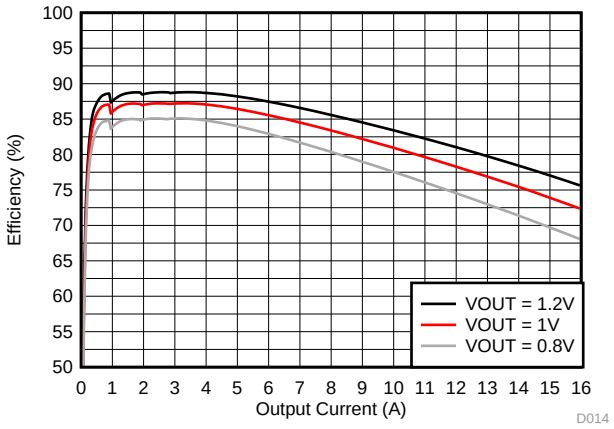
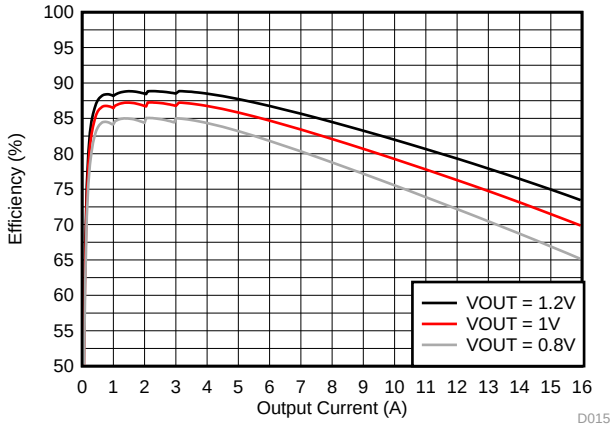
9.2.3 Application Curves

Unless otherwise specified: V_{IN} = 3.7 V, V_{OUT} = 1 V, $V_{(NRST)}$ = 1.8 V, T_A = 25 °C, f_{SW} = 3 MHz, L = 330 nH (TOKO DFE252010F-R33M), C_{POL} = 22 µF. Measurements done with connections in Figure 21.



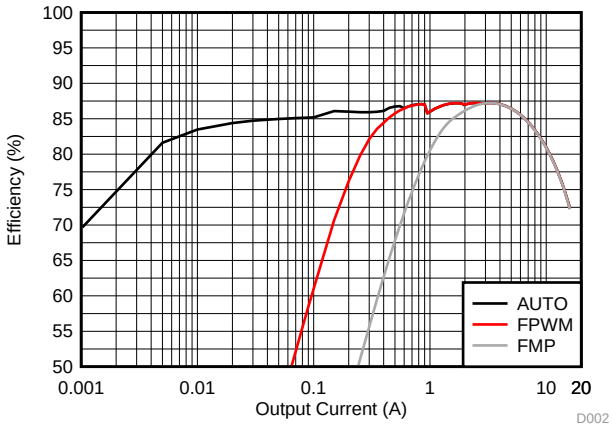
V_{IN} = 3.7 V

Figure 22. Efficiency in Forced PWM Mode



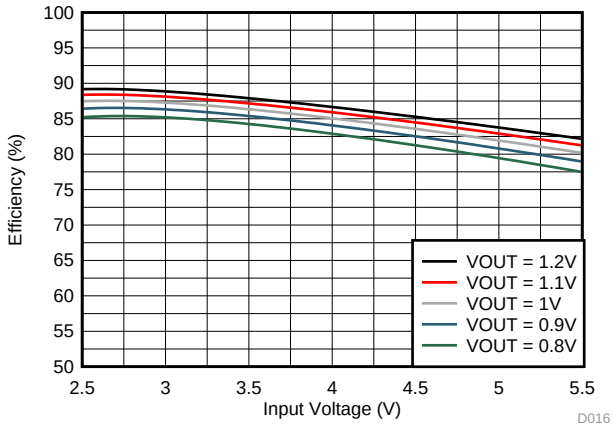
V_{IN} = 3.7 V Inductor = TDK VLS252010HBX-R47M

Figure 23. Efficiency in Forced PWM Mode



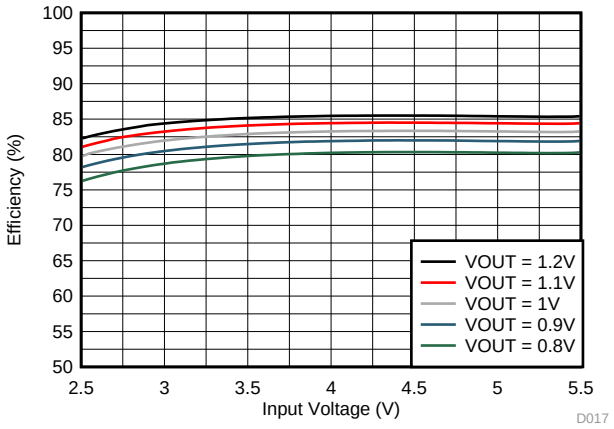
V_{IN} = 3.7 V V_{OUT} = 1 V

Figure 24. Efficiency in PFM, PWM and Forced Multi-Phase Mode



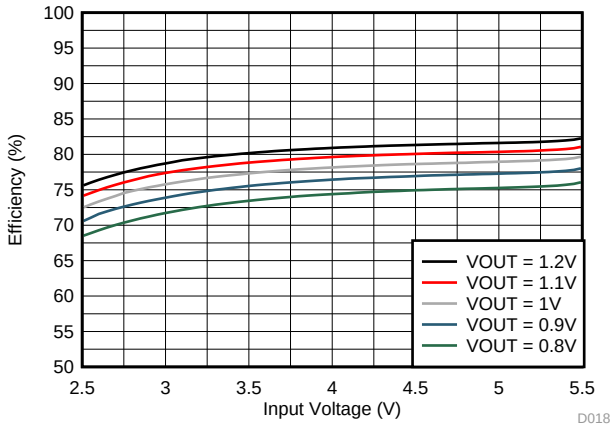
Load = 1 A

Figure 25. Efficiency vs Input Voltage



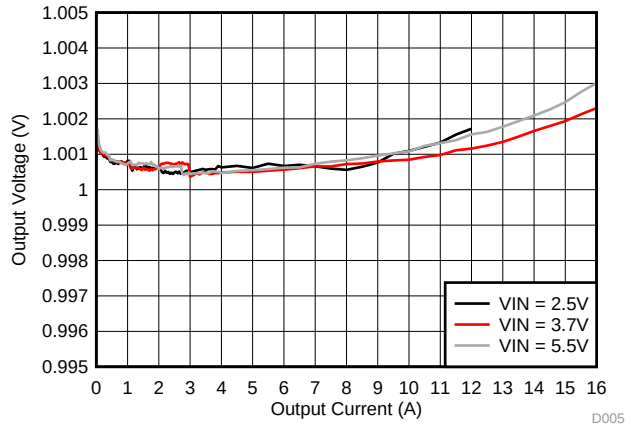
Load = 8 A

Figure 26. Efficiency vs Input Voltage



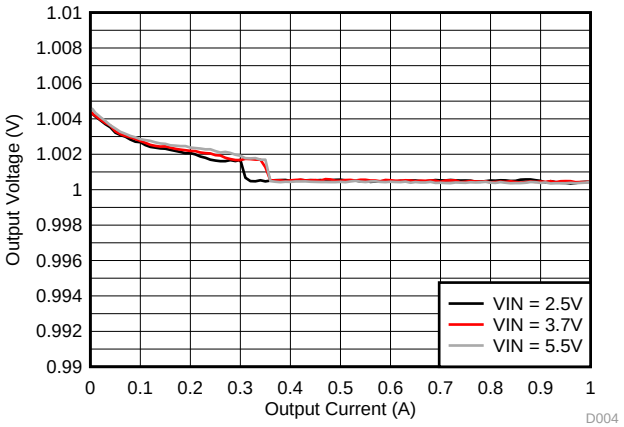
Load = 12 A

Figure 27. Efficiency vs Input Voltage



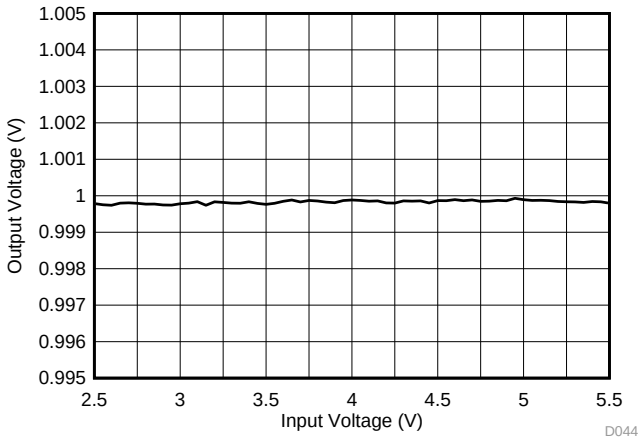
V_{OUT} = 1 V

Figure 28. Output Voltage vs Load Current in Forced PWM Mode



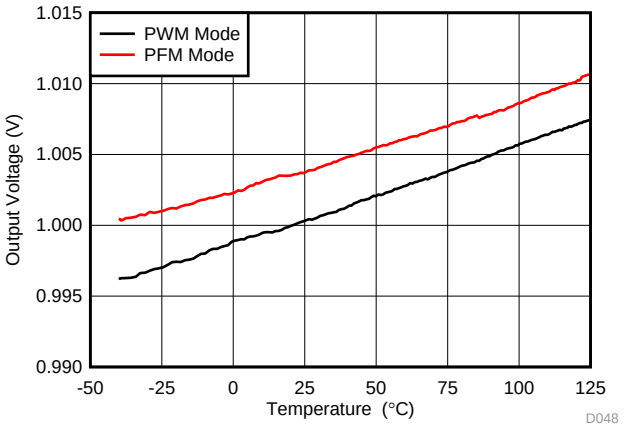
V_{OUT} = 1 V

Figure 29. Output Voltage vs Load Current in PWM-PFM Mode



V_{OUT} = 1 V Load = 1 A

Figure 30. Output Voltage vs Input Voltage in PWM Mode



V_{OUT} = 1 V
Load = 3 A (PWM Mode) and 100 mA (PFM Mode)

Figure 31. Output Voltage vs Temperature

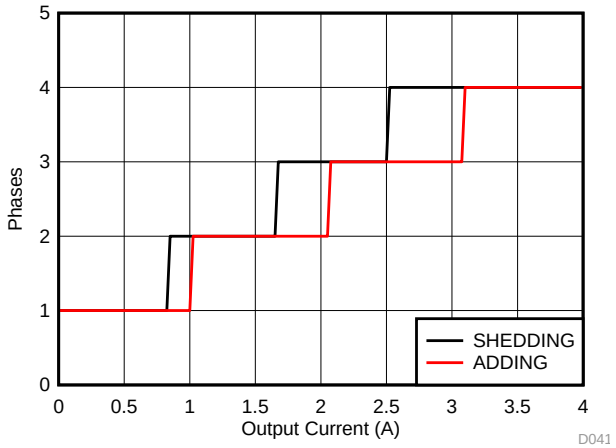
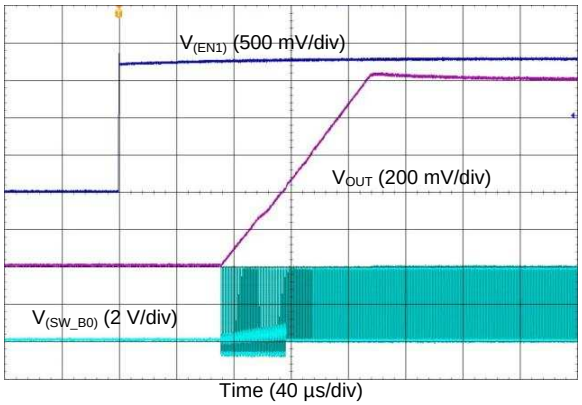


Figure 32. Phase Adding and Shedding vs Load Current



Load = 0 A

Figure 33. Start-up with EN1, Forced PWM

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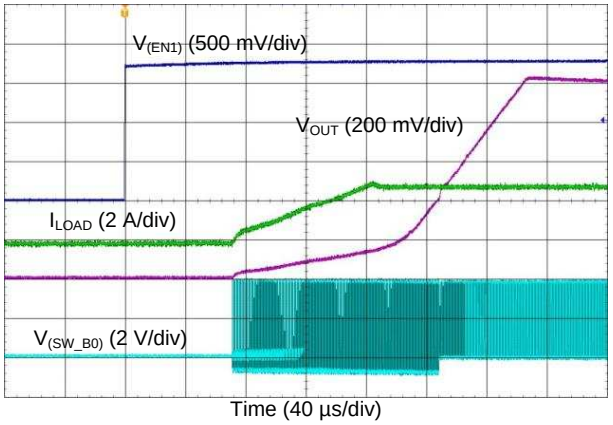
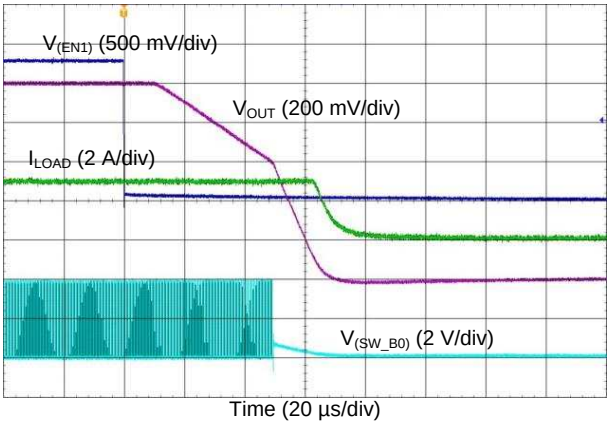


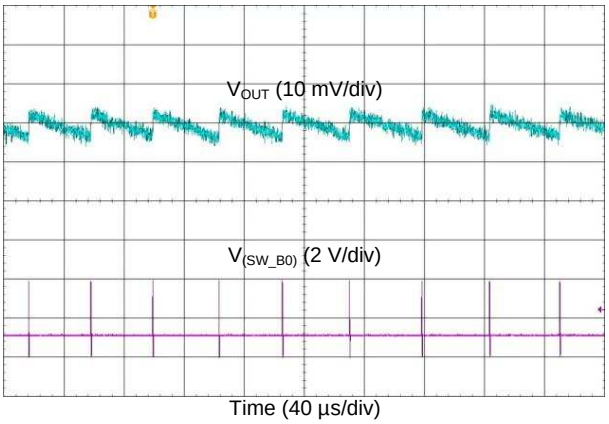
Load = 3 A

Figure 34. Start-up with EN1, Forced PWM



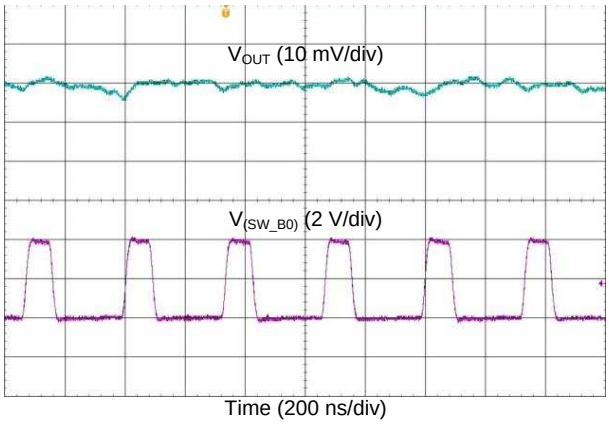
Load = 3 A

Figure 35. Shutdown with EN1, Forced PWM



Load = 10 mA

Figure 36. Output Voltage Ripple, PFM Mode



Load = 200 mA

Figure 37. Output Voltage Ripple, Forced PWM Mode

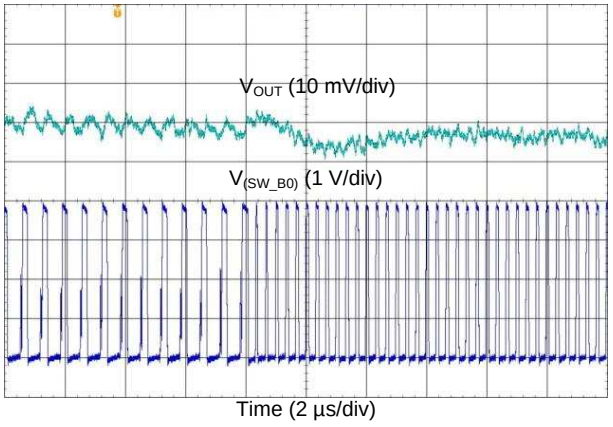


Figure 38. Transient from PFM-to-PWM Mode

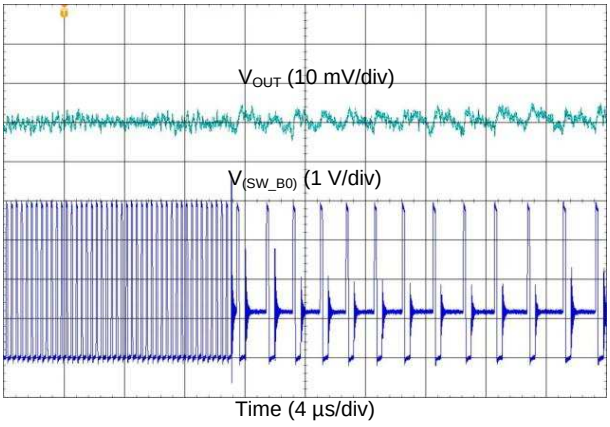


Figure 39. Transient from PWM-to-PFM Mode

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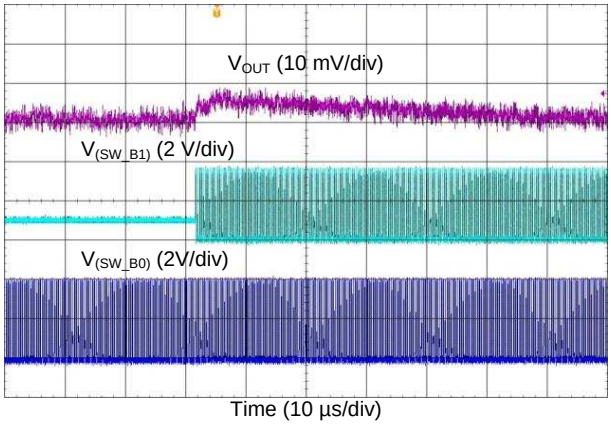

Figure 40. Transient from 1-Phase to 2-Phase Operation

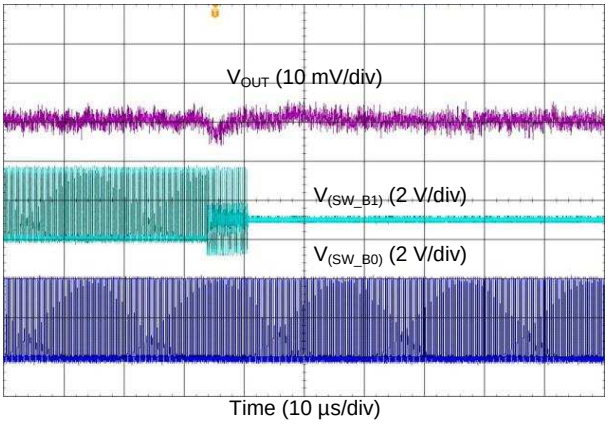

Figure 41. Transient from 2-Phase to 1-Phase Operation

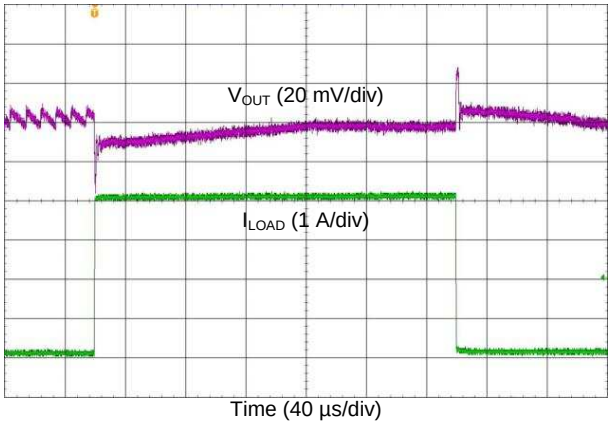

Figure 42. Transient Load Step Response, AUTO Mode

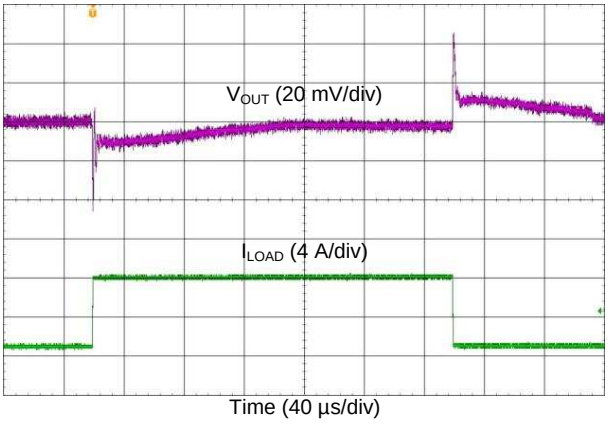

Figure 43. Transient Load Step Response, FPWM Mode

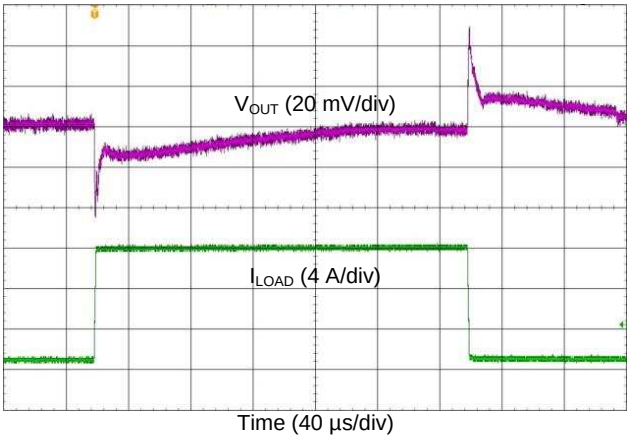

Figure 44. Transient Load Step Response, FPWM Mode

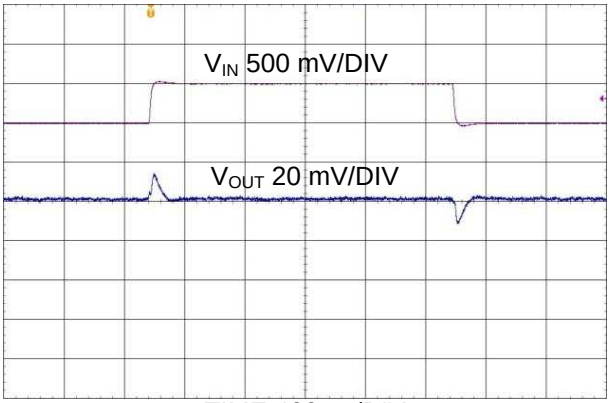

Figure 45. Transient Line Response

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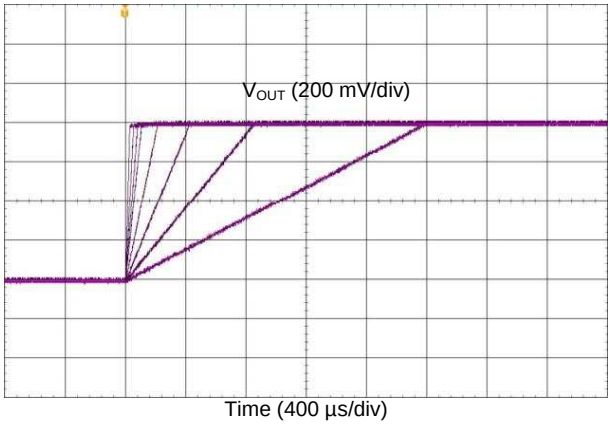

Figure 46. V_{OUT} Transition from 0.6 V to 1.4 V with Different Slew Rate Settings

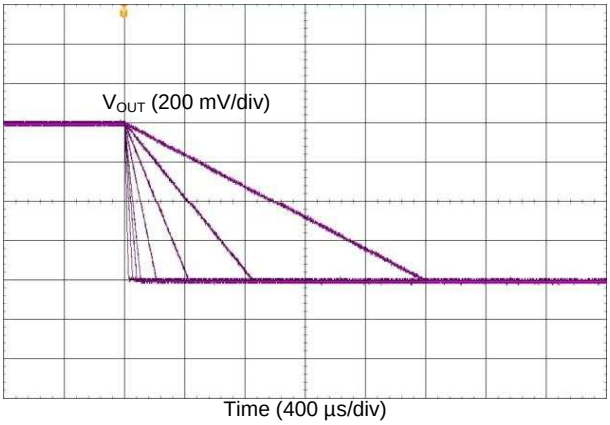

Figure 47. V_{OUT} Transition from 1.4 V to 0.6 V with Different Slew Rate Settings

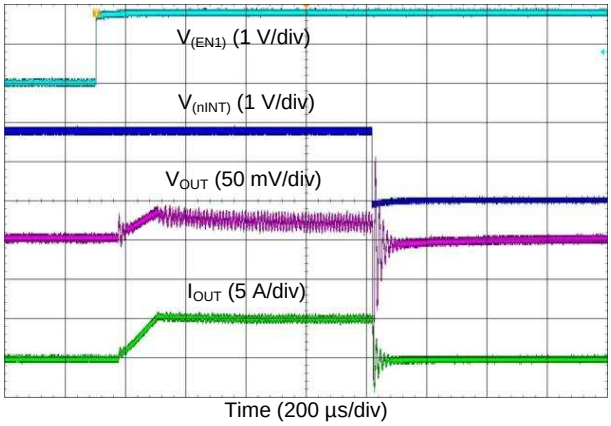

Figure 48. Start-up with Short on Output

10 Power Supply Recommendations

The device is designed to operate from an input voltage supply range between 2.5 V and 5.5 V. This input supply should be well-regulated and able to withstand maximum input current and maintain stable voltage without voltage drop even at load transition condition. The resistance of the input supply rail should be low enough that the input current transient does not cause too high drop in the LP8758 supply voltage that can cause false UVLO fault triggering. If the input supply is located more than a few inches from the LP8758 additional bulk capacitance may be required in addition to the ceramic bypass capacitors.

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11 Layout

11.1 Layout Guidelines

The high frequency and large switching currents of the LP8758 make the choice of layout important. Good power supply results only occur when care is given to proper design and layout. Layout affects noise pickup and generation and can cause a good design to perform with less-than-expected results. With a range of output currents from milliamps to 10 A and over, good power supply layout is much more difficult than most general PCB design. Use the following steps as a reference to ensure the device is stable and maintains proper voltage and current regulation across its intended operating voltage and current range.

1. Place C_{IN} as close to the VIN_Bx pin and the PGND_Bxx pin as possible. Route the V_{IN} trace wide and thick to avoid IR drops. The trace between the input capacitor's positive node and LP8758's VIN_Bx pin(s) as well as the trace between the input capacitor's negative node and power PGND_Bxx pin(s) must be kept as short as possible. The input capacitance provides a low-impedance voltage source for the switching converter. The inductance of the connection is the most important parameter of a local decoupling capacitor — parasitic inductance on these traces must be kept as tiny as possible for proper device operation.
2. The output filter, consisting of L_x and C_{OUTx}, converts the switching signal at SW_Bx to the noiseless output voltage. Place the output filter as close to the device as possible, keeping the switch node small, for best EMI behavior. Route the traces between the LP8758's output capacitors and the load's input capacitors direct and wide to avoid losses due to the IR drop.
3. Input for analog blocks (VANA and AGND) must be isolated from noisy signals. Connect VANA directly to a quiet system voltage node and AGND to a quiet ground point where no IR drop occurs. Place the decoupling capacitor as close to the VANA pin as possible. VANA must be connected to the same power node as VIN_Bx pins.
4. If the processor load supports remote voltage sensing, connect the LP8758's feedback pins FB_Bx to the respective sense pins on the processor. The sense lines are susceptible to noise. They must be kept away from noisy signals such as PGND_Bxx, VIN_Bx, and SW_Bx, as well as high bandwidth signals such as the I^2C. Avoid both capacitive as well as inductive coupling by keeping the sense lines short, direct and close to each other. Run the lines in a quiet layer. Isolate them from noisy signals by a voltage or ground plane if possible. TI recommends running the signal as a differential pair.
5. Route PGND_Bxx, VIN_Bx and SW_Bx on thick layers. They must not surround inner signal layers which are not able to withstand interference from noisy PGND_Bxx, VIN_Bx and SW_Bx.

Due to the small package of this converter and the overall small solution size, the thermal performance of the PCB layout is important. Many system-dependent issues such as thermal coupling, airflow, added heat sinks and convection surfaces, and the presence of other heat-generating components affect the power dissipation limits of a given component. Proper PCB layout, focusing on thermal performance, results in lower die temperatures. Wide power traces come with the ability to sink dissipated heat. This can be improved further on multi-layer PCB designs with vias to different planes. This results in reduced junction-to-ambient ($R_{\theta JA}$) and junction-to-board ($R_{\theta JB}$) thermal resistances, thereby reducing the device junction temperature, T_J. Performing a careful system-level 2D or full 3D dynamic thermal analysis at the beginning product design process is strongly recommended, using a thermal modeling analysis software.

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11.2 Layout Example

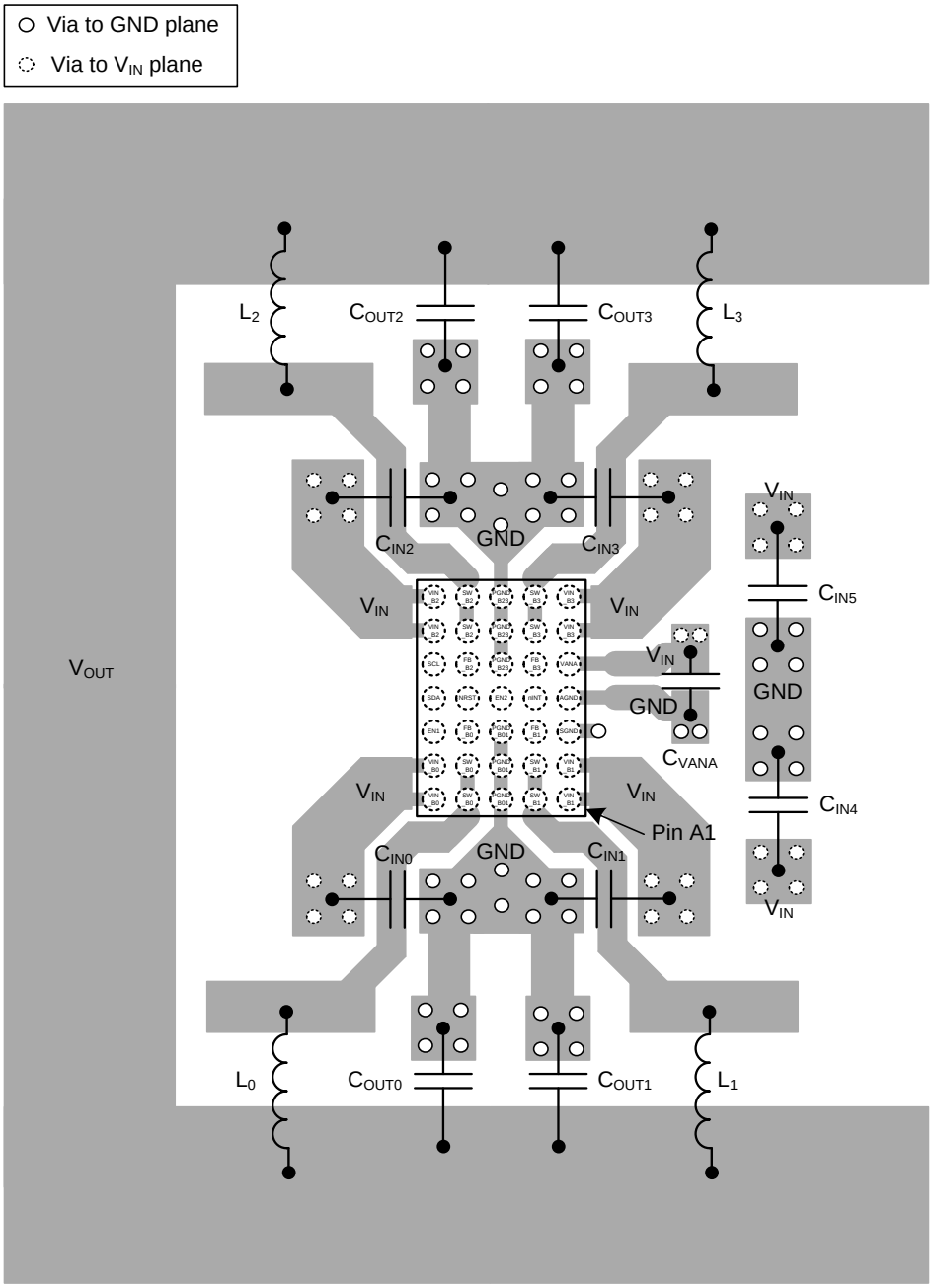

Figure 49. LP8758 Board Layout

12 デバイスおよびドキュメントのサポート

12.1 デバイス・サポート

12.1.1 デベロッパー・ネットワークの製品に関する免責事項

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12.2 ドキュメントのサポート

12.2.1 関連資料

関連資料については、以下を参照してください。

『AN-1112 DSBGAウェハー・レベル・チップ・スケール・パッケージ』

12.3 ドキュメントの更新通知を受け取る方法

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12.4 コミュニティ・リソース

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

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12.5 商標

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12.7 Glossary

SLYZ022 — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 メカニカル、パッケージ、および注文情報

以降のページには、メカニカル、パッケージ、および注文に関する情報が記載されています。この情報は、そのデバイスについて利用可能な最新のデータです。このデータは予告なく変更されることがあり、ドキュメントが改訂される場合もあります。本データシートのブラウザ版を使用されている場合は、画面左側の説明をご覧ください。

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LP8758A1B0YFFR	ACTIVE	DSBGA	YFF	35	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85	8758A1B0	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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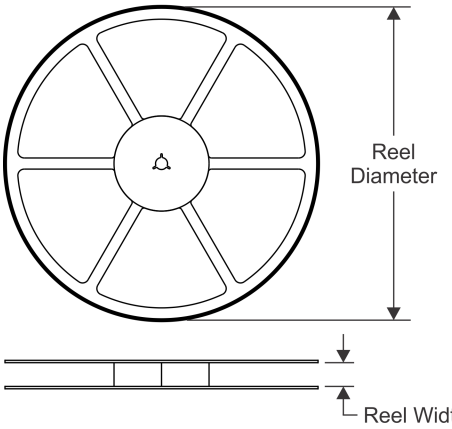
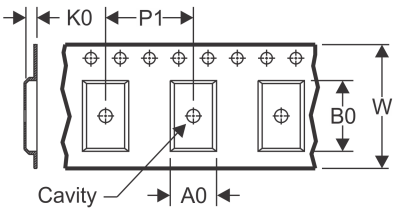
TAPE AND REEL INFORMATION

REEL DIMENSIONS

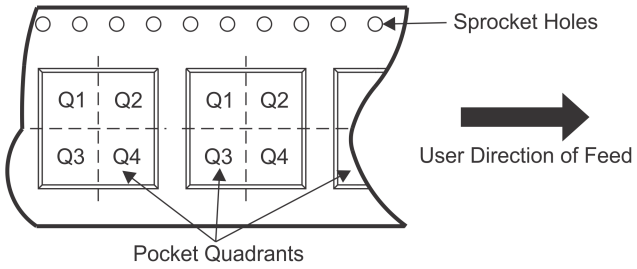


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

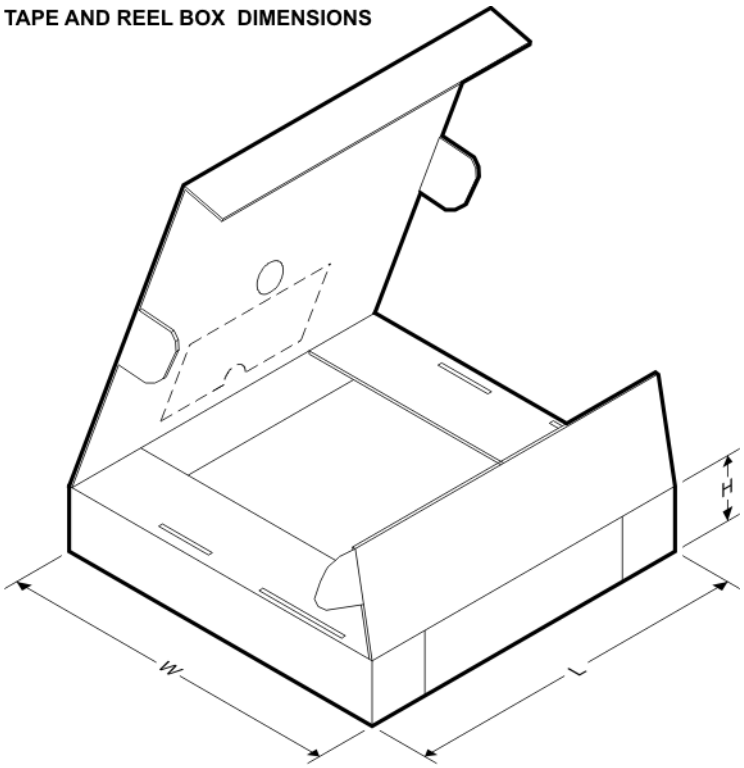
QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LP8758A1B0YFFR	DSBGA	YFF	35	3000	180.0	8.4	2.28	3.03	0.74	4.0	8.0	Q1

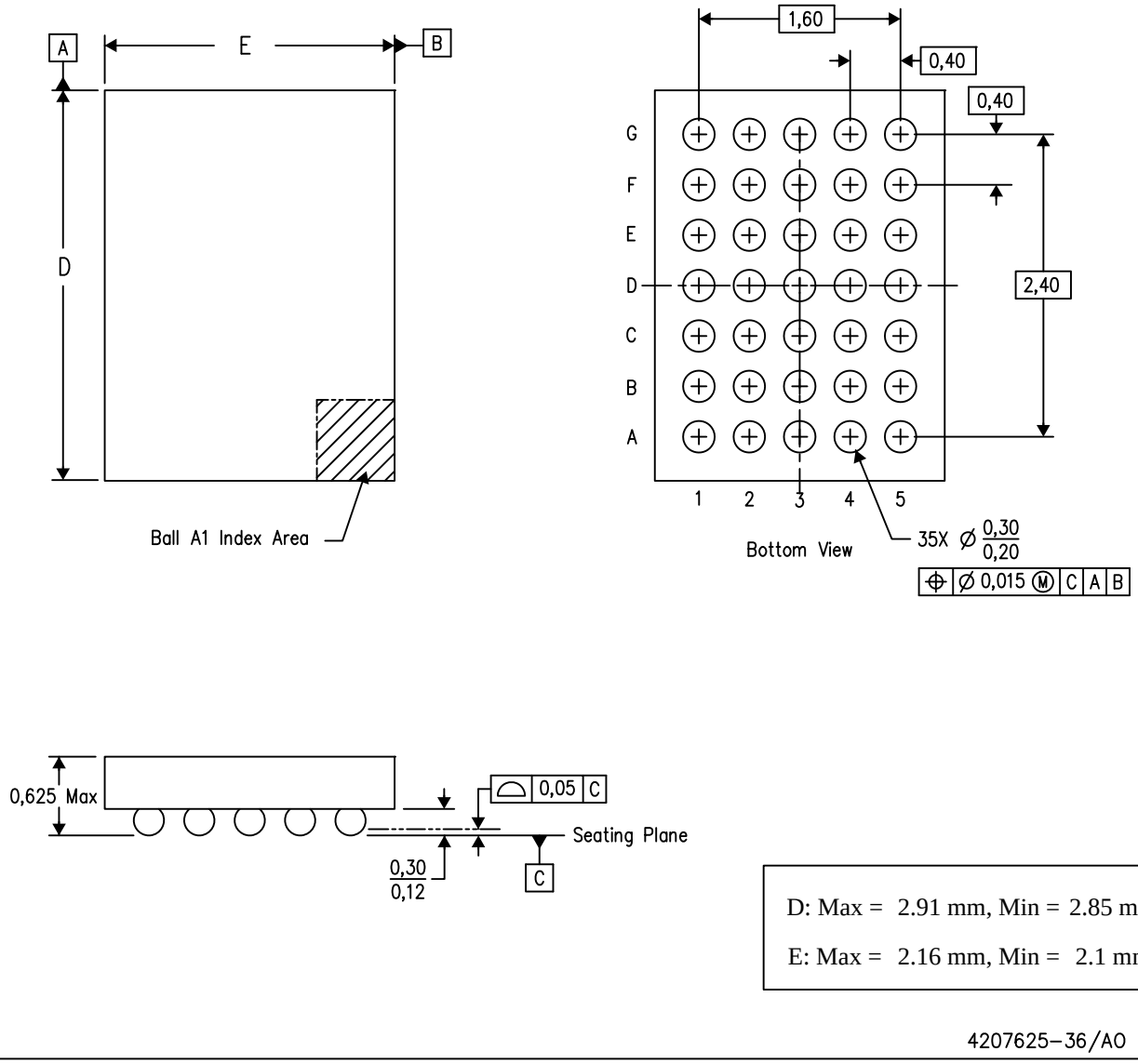
TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LP8758A1B0YFFR	DSBGA	YFF	35	3000	182.0	182.0	20.0

YFF (R–XBGA–N35) DIE–SIZE BALL GRID ARRAY



D: Max = 2.91 mm, Min = 2.85 mm

E: Max = 2.16 mm, Min = 2.1 mm

4207625–36/AO 12/13

NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M–1994.
B. This drawing is subject to change without notice.
C. NanoFree™ package configuration.

NanoFree is a trademark of Texas Instruments.

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