

Low-Voltage, Single-Supply 232-Standard Interface Solutions

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ABSTRACT

Features and key specifications for Texas Instruments (TI[™]) new low-voltage, single-supply, 232-standard-compliant interface devices are presented, as well as test data comparing TI versus competitor same-function devices. Evolution of the 232 standard is discussed relative to TI devices that have significantly greater capability than the 232 standard requires.

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Introduction

This application report provides information about the new low-voltage, single-supply interface devices produced by TI. These devices include the popular three-driver/five-receiver data terminal equipment (DTE) device named MAX3243, and specifically are designed for the personal computer (PC) serial port I/O, which usually is used with the 9-pin DB9S connector. Another new device is the five-driver/three-receiver data-circuit-terminating equipment (DCE) device named MAX3238, which is the complement to the DTE device, and is commonly used on the peripheral end of the connecting cable. Also, the MAX3221, which is the smallest possible 232-standard solution, with one driver and one receiver per package, is another member of this family of devices.

Until now, TI has not offered pure CMOS 232-interface devices that operate at supply voltages from 3 V to 5.5 V. These three new TI integrated circuits (ICs) incorporate an on-board charge pump that, with four small external capacitors, produces the necessary driver output voltage swings to comply with the 232 standard. Low-power, high human-body model (HBM) electrostatic discharge (ESD) serial-port protection, high data rates, and small packages are some of the benefits featured on these devices.

Brief Overview of the 232 Standard

Officially, the 232 standard's title is *Interface Between Data Terminal Equipment and Data Circuit-Terminating Equipment Employing Serial Binary Data Interchange*. The current revision is TIA/EIA-232-F (previously referred to as RS-232). In the past, the term RS-232 has been used interchangeably with its official name; however, the RS prefix no longer is used and this document follows the new terminology. The standard discusses electrical specifications, i.e., the physical layer, protocol requirements, and specifies a 25-pin D-type connector that attaches to the cable. Discussion of mechanical characteristics is beyond the scope of this application report. Protocol is discussed briefly only for signals used with the 9-pin connector.

The 232 standard was introduced in 1962 as an effort to standardize the interface for serial communication between the DTE and DCE. Early adopters emphasized the interface between the PC (DTE) and the modem (DCE). Wide acceptance of the PC quickly ensured that the 232 standard would become the industry standard it is today – a low-cost, serial interface. The mouse, plotter, printer, scanner, digitizer, external modems, and test equipment are examples of peripherals that connect to the 232-compliant serial port.

Throughout the years, the 232 standard has been revised a number of times. Figure 1 shows several transition-time vs data-rate changes made to the standard. The latest revision makes 232 specifications agree with international standards ITU-T V.24, V.28, and ISO/IEC 2110.



Figure 1. Changes in 232-Standard Transition Times vs Data Rates

TIA/EIA-232-F places four constraints that limit the unit interval, which is the inverse of data rate, and transition time, defined by the slew rate. Referring to Figure 1, the first constraint is set by the top line. The standard specifies that for any data rate less than or equal to 40 bps, the transition time cannot exceed 1 ms. The second constraint is illustrated as the 4%-limit line drawn diagonally from 40 bps to 200 kbps. This specifically states that for any unit interval less than or equal to 25 ms (equivalent to \geq 40 bps data rate) but greater than or equal to 50 µs (equivalent to \leq 20 kbps data rate), the transition time must not exceed 4% of the unit interval. The third constraint shown pertains to the bottom line, which places a maximum limit of 30 V/µs on the driver output slew rate. This maximum slew rate translates to a minimum transition time of 0.2 µs because 232 voltage swings should never fall below 6 V (0.2 µs = 6 V/30 V/µs). Finally, the fourth limit is data rate, which is widely known in the industry to be only 20 kbps. Hence, the dashed line in Figure 1.

ANSI TIA/EIA-232-F Electrical Specification

It is evident from the specifications in Table 1 that the 232 standard was created when power considerations were not scrutinized closely and data signaling rates, although adequate at that time, were extremely slow when compared to today's standards. To be compatible with the original standard, receiver inputs and driver outputs must tolerate ± 25 -V voltage swings. Continually increasing demand for speed has forced discrete-IC manufacturers to violate the 232-standard maximum 20-kbps specification at the expense of ignoring the 4% limit shown in Figure 1. Today, 250-kbps data-signaling rates are common. Table 1 presents key 232-compliant specifications and features. Figure 2 graphically shows driver and receiver compatibility specifications.

PARAMETER		SPECIFICATION/ FEATURE	COMMENT
Mode of operation		Single-ended	One driver and one receiver per line
Maximum cable chara	acteristic	2500 pF	Most cables typically are 15 m to 20 m long
Maximum data signal	ing rate	20 kbps	Most discrete-IC vendors violate this specification.
Maximum common-m	ode voltage	±25 V	
Driver output levels	Unloaded	≤±25 V	
	Loaded	± 5 V to ± 15 V	
Driver load		3 k Ω to 7 k Ω	
Driver slew rate		≤30 V/μs	
Driver output short-circuit current		≤100 mA	Driver output to any conductor
Receiver input resistance		3 k Ω to 7 k Ω	
Receiver sensitivity		±3 V	Minimum required input voltage
Noise margin		±2 V	Difference between minimum driver output voltage level and minimum receiver input voltage

Table 1. Key 232-Standard Electrical Specifications





Figure 2. Driver and Receiver Voltage Levels

232-Standard Handshaking Signals for DB9S Connections

As mentioned previously, the 25-pin connector is recommended in the 232 standard. However, in today's applications, the 9-pin D-type connector, commonly known as the DB9S connector, is chosen by most users because not all handshaking signals are necessary in most applications. Typically, the handshaking signals used with the DB9S connector are sufficient to support error-free data transmission. A description of the protocol signals and their corresponding DB9S connector pins are presented in Table 2.



Table 2. 232 Signal Descriptions

TERMINAL		DESCRIPTION				
NAME	NO.	DESCRIPTION				
DCD	1	Data carrier detect. The on condition of this signal line, as sent by the DCE, informs the DTE that it is receiving a carrier signal from the remote DCE that meets its criteria.				
DSR	6	Data set ready. A signal turned on by the DCE to indicate to the DTE that it is connected to the line.				
RD	2	Receive data line. Serial data from DCE to DTE. When DCD is in the off condition, the RD signal must be in the MARK state.				
CTS	8	Clear to send. Turned on by the DCE to inform the DTE that it is ready to receive data.				
RTS	7	Request to send. Indicates the DTE is ready to transmit data, and then the DCE must prepare to receive data. After some delays, the DCE turns on the CTS line to inform the DTE it is ready to receive data. Once communication is complete, the DTE turns off the RTS signal. After a brief delay to ensure that all transmitted data has been received, the DCE turns off CTS.				
TD	3	Transmit data line. Serial data from DTE to DCE.				
DTR	4	Data terminal ready. This signal, in conjunction with DSR, indicates equipment readiness.				
RI	9	Ring indicator. Turned on by the DCE while ringing is being received.				
GND	5	Ground				

Technology Evolution

The minimum driver output voltage level specified for loaded conditions is ± 5 V (see Table 1). For 232 drivers that transmit data through a cable, traditionally, this voltage level has been accomplished using multiple supplies to the discrete IC. Because ± 12 V has been available readily on most systems, this voltage was applied to the 232 device to produce the necessary bus-voltage swings. The trend in recent years is to reduce power dissipation, especially in portable applications. Therefore, board designers have been reducing power-supply voltages, or eliminating them altogether, wherever possible. The lowering of supply-voltage levels is a trait common to the semiconductor industry. Consequently, the burden of generating the necessary 232-compliant voltage swings is on the 232 driver IC. This can be accomplished with an internal or external charge pump or dc-to-dc converter.

TI devices incorporate the 232 driver and charge pump in a single solution that makes it easy to design a circuit board using only a single supply. The new MAX3243, MAX3238, and MAX3221 devices operate from 3 V to 5.5 V and require only four small external capacitors (0.1 μ F) for the charge-pumping action. These devices are built in the state-of-the-art LinBiCMOS process, making them the only 232-compliant devices in the broad TI product portfolio with a wide low-voltage supply range.

Key Features and Related Benefits

Table 3 presents important features and their respective benefits per device. TI plans to offer equivalent devices that operate at 1 Mbps for each function, and have the exact pinouts as their parent devices. These 1-Mbps devices will be SN75C3243, SN75C3238, and SN75C3221 (see Table 5). Their part numbers imply that their functionality is identical to their respective parents.

To transmit at 1 Mbps, the driver output slew-rate maximum specification of 30 V/ μ s is violated. Thus, basically, the only two electrical differences between the MAXxxx devices and their respective 1-Mbps versions is slew rate and data rate. The faster SN75C devices are specified for 24 V/ μ s to 150 V/ μ s driver output slew rate.

FEATURE	MAX3243	MAX3238	MAX3221	BENEFIT
Single-chip solution for 232 DTE serial port	r			Ease of design and board-space savings
Single-chip solution for one driver and one receiver 232 serial port			r	Ease of design
Single-chip solution for 232 DCE serial port		~		Ease of design and board-space savings
Low standby current (10 µA max)	~	~	~	Reduced power consumption and increased battery life for battery-powered systems
Operates from a single 3-V to 5.5-V power supply	~	~	~	Ease of design
Always-active receiver	~	 ✓ 		Listens to bus while in standby
Mouse driveability	~			Ease of design
Accepts 5-V logic levels with 3.3-V power supply	~	~	~	Allows mixed-voltage systems
Drop-in replacement for Maxim equivalent	v	V	~	Second-source supplier
232 bus-pin ESD protection exceeds ±15 kV HBM	~	V	~	Best-in-the-industry serial-port HBM ESD protection
Auto-powerdown	V	V		Device automatically powers down with serial-port inactivity
Auto-powerdown plus			~	Device automatically powers down with serial-port and logic input inactivity
1-Mbps data-rate equivalent	SN75C3243	SN75C3238	SN75C3221	Increased data throughput

Table 3. Device Features and Benefits



Typical Design Applications

Because the 232 standard has existed since the 1960s, transmitting data via the serial bus is not new. However, offering the end-equipment manufacturer flexibility to automatically power down host and peripheral interface devices to conserve power is new to TI 232 devices and is especially useful in battery-powered environments. In addition, as the 232 devices are powered down, one receiver always is kept active to signal the supervisory circuitry of incoming data.

Most integrated circuits are designed with an external pin that can be biased appropriately to remove power from the device, thereby forcing a high-impedance state at all outputs. The MAX3xxx devices (and planned 1-Mbps siblings) also have this functionality, but go one step further by allowing the 232 device to monitor serial-port activity and power down after 30 μ s of inactivity. This feature is called auto-powerdown. The MAX3238 and SN75C3238 also monitor the digital logic input side (driver inputs) and, if auto-powerdown criteria are met, the device powers down after 30 s of inactivity. This important feature is called auto-powerdown plus.

For the host to communicate with the peripheral in a two-way serial-data transmission application, conversion from TTL/CMOS to 232 levels must be accomplished. This is the sole purpose for the existence of any 232-compliant device. The onboard charge pump of these devices is capable of producing greater than ± 5 V from a single 3-V power supply, thus complying with the 232 standard.

To take advantage of these powerful functions, 232 applications can be designed so that the serial port draws current from the power supply only during transmission of data. Figure 3 shows an application where the MAX3243 is used on the host (DTE) and the MAX3238 is used on the peripheral (DCE). The host can be a PC, for example, while the peripheral can be an external 56-kbps modem. Four small 0.1- μ F external capacitors (C1–C4) are required to produce the 232-compliant driver output voltage levels. C_{BYPASS}, also 0.1 μ F, is recommended to decouple noise on the supply, and should be placed as close as possible to the V_{CC} terminal.

Auto-powerdown can be deactivated by enabling the MAX3243 on the host using the RI signal from the peripheral, if it is connected to RIN2. The always-active receiver output, ROUT2B, which is electrically connected internally to RIN2, notifies the microprocessor, via the microprocessor supervisory circuitry, of impending data. Toggling RIN2 with the RI signal automatically powers up the MAX3243.



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A 232 connection that requires only one driver and one receiver can be made with only three wires; one to transmit data, one to receive, and ground. No handshaking is used in this type of application, so coordination between the two units is critical. Typically, this type of application is used to allow for a large margin of error due to the lack of handshaking, and today is used widely in handheld, portable applications. Therefore, the receiver on either side must accept data as fast as the driver can transmit. In this application, the unit accepting data cannot notify the transmitter that it is going too fast.

Figure 4 shows a null-modem connection for a three-wire transmission scheme, for which the MAX3221 is ideal. A null modem can connect two DTEs to make them believe handshaking is occurring.



Figure 4. Null-Modem Wiring Diagram

The theory of operation is that the DTE thinks it is talking to a DCE rather than another DTE. TD must be connected to RD when transmitting from one DTE to the other. Signal ground also must be connected so both remote grounds are common to each DTE. The DTR signal is looped back to DSR and DCD on both sides to make the transmitting unit think that the receiving unit is ready to accept data. When DTR is asserted active, DSR and DCD immediately become active. This causes the transmitting DTE to think the virtual modem to which it is connected is ready and waiting.

Because both DTEs communicate at the same speed, handshaking is not used; thus, RTS and CTS are connected together on each side. When transmission is requested by a DTE, RTS is asserted high and gets immediate feedback that it may send data. RI is not connected on either side because this function is not used.

Analysis of TI and Competitor Equivalent Devices

This section compares performance of the TI MAX3243 and that of a competitor's '3243. Specifically, driver output voltage, supply current, and slew rate are examined closely to provide a good understanding of the charge-pump and output-driver performance, and an estimate of the expected supply current. All measurements are taken at room temperature with a 5-k Ω load resistor at the driver output. The analysis was performed only on the '3243 device, but similar results can be expected from the MAX3238 and MAX3221.

Driver Output Voltage

Figures 5 and 6 indicate the performance of the onboard charge pump by displaying the competitor's '3243 driver output voltage at different data rates and different numbers of switching outputs. Figures 7 and 8 show the driver output voltage of the TI MAX3243. This data was obtained while switching one output at 250 kbps, switching two outputs at 120 kbps, and switching three outputs at 50 kbps. Both devices meet the minimum 232 driver output voltage requirement ($V_{OH} = +5 V$ and $V_{OL} = -5 V$).



Figure 5. Competitor's Device Output Voltage vs C_L (V_{CC} = 3.3 V)



Figure 6. Competitor's Device Output Voltage vs C_L (V_{CC} = 5 V)



Figure 7. TI Device Output Voltage vs C_L (V_{CC} = 3.3 V)



Figure 8. TI Device Output Voltage vs C_L ($V_{CC} = 5 V$)

Supply Current

Figures 9 and 10 show the supply current of the competitor's '3243 device while varying data rates and number of switching outputs. Figures 11 and 12 show TI MAX3243 device supply current. The TI MAX3243 slightly outperforms the competing device at the highest current, which is at 250 kbps, 2500-pF load, and one switching-output driver.



Figure 9. Competitor's Device Supply Current vs C_L (V_{CC} = 3.3 V)



Figure 10. Competitor's Device Supply Current vs C_L (V_{CC} = 5 V)



Figure 11. TI Device Supply Current vs C_L (V_{CC} = 3.3 V)



Figure 12. TI Device Supply Current vs C_L ($V_{CC} = 5 V$)



Slew Rate

The 232 standard specifies the slew rate of the output driver to be no greater than 30 V/ μ s. With other specifications, the maximum limit on slew rate ensures that any 232-compliant connection behaves as a lumped load rather than as a transmission line. Therefore, it is critical that an output driver not exceed this slew-rate specification limit for true 232 compliance.

Figures 13 and 14 show the slew-rate performance of the competitor's '3243 at different combinations of data rates and numbers of switching outputs. Figures 15 and 16 show the slew rate of the TI MAX3243. From these graphs, two important observations can be made. First, the TI MAX3243 has a tighter slew-rate variation. Second, and possibly more important, various combinations of data rate versus switching outputs do not seem to affect the TI MAX3243 as much as it does the competitor's equivalent.















Figure 16. TI Device Slew Rate vs C_L (V_{CC} = 5 V)

Figures 17–20 further strengthen the observation regarding the TI device's better slew-rate performance compared to the competitor's equivalent. These data points were generated while switching one output at 250 kbps, switching two outputs at 120 kbps, and switching three outputs at 50 kbps. From top to bottom, curves in Figures 17–20 represent capacitive loading of 0 pF, 500 pF, 1000 pF, 2000 pF, and 2500 pF, respectively. The nearly flat slew-rate response of the TI MAX3243 is superior to the competitor's device.



Figure 17. Competitor's Device Slew Rate vs Data Rate (V_{CC} = 3.3 V)



Figure 18. Competitor's Device Slew Rate vs Data Rate (V_{CC} = 5 V)



Figure 19. TI Device Slew Rate vs Data Rate (V_{CC} = 3.3 V)



Figure 20. TI Device Slew Rate vs Data Rate (V_{CC} = 5 V)

Conclusion

A summary of the laboratory experiments is shown in Table 4. TI MAX3243, MAX3238, and MAX3221 integrated 232-compliant solutions offer superior performance for single, low-voltage applications that operate from 3 V to 5.5 V. Comparison of the TI MAX3243 with the leading 232-compliant competitor's equivalent shows the same, or better, performance from the charge pump that provides the necessary voltages for the output driver.

TEST TEST CONDITIONS		TI MAX3243	COMPETITOR'S '3243	COMMENTS	
Output voltage (V _O)	V_{CC} = 5 V, C _L = 2500 pF, One output switching at 250 kbps	V _{OH} = 5.2 V V _{OL} = -5.1 V	V _{OH} = 6.4 V V _{OL} = -5.5 V	Both meet the minimum	
vs Load capacitance (C _L)	V_{CC} = 3.3 V, C _L = 2500 pF, One output switching at 250 kbps	V _{OH} = 5.5 V V _{OL} = -5.0 V	V _{OH} = 5.6 V V _{OL} = -5.5 V	232 requirement for V _O (V _{OH} > 5 V and V _{OL} < –5 V).	
Supply current (I _{CC}) vs Load capacitance (CL)	V _{CC} = 5 V, C _{load} = 2500 pF, One output switching at 250 kbps I _{CC} = 35.75 mA I _{CC} = 38.25 mA Red		Reduced power consumption		
	V_{CC} = 3.3 V, C_{load} = 2500 pF, One output switching at 250 kbps	I _{CC} = 36.25 mA	I _{CC} = 38.25 mA	on TI MAX3243	
Slew rate	V _{CC} = 5 V	Slew-rate variation is 8.9 V/µs.	Slew-rate variation is 13.5 V/μs.	TI MAX3243 has better	
Siew Idle	V _{CC} = 3.3 V	Slew-rate variation is 7.6 V/µs.	Slew-rate variation is 11.4 V/μs.	slew-rate performance.	

Table 4. Summary Comparison of TI MAX3243 and Competitor's '3243
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Orderable part numbers of devices soon to be available are listed in Table 5.

PART NUMBER	PACKAGES AVAILABLE	TEMPERATURE RANGE	MAXIMUM DATA RATE	PIN COUNT
MAX3243C	Wide-SOIC, SSOP, and TSSOP	0°C to 70°C	250 kbps	28
MAX3243I	Wide-SOIC, SSOP, and TSSOP	–40°C to 85°C	250 kbps	28
SN75C3243	Wide-SOIC, SSOP, and TSSOP	0°C to 70°C	1 Mbps	28
SN65C3243	Wide-SOIC, SSOP, and TSSOP	–40°C to 85°C	1 Mbps	28
MAX3238C	Wide-SOIC, SSOP, and TSSOP	0°C to 70°C	250 kbps	28
MAX3238I	Wide-SOIC, SSOP, and TSSOP	–40°C to 85°C	250 kbps	28
SN75C3238	Wide-SOIC, SSOP, and TSSOP	0°C to 70°C	1 Mbps	28
SN65C3238	Wide-SOIC, SSOP, and TSSOP	–40°C to 85°C	1 Mbps	28
MAX3221C	SSOP and TSSOP	0°C to 70°C	250 kbps	16
MAX32211	SSOP and TSSOP	–40°C to 85°C	250 kbps	16
SN75C3221	SSOP and TSSOP	0°C to 70°C	1 Mbps	16
SN65C3221	SSOP and TSSOP	–40°C to 85°C	1 Mbps	16

Reference

Information about 232-compliant connections can be accessed at www.beyondlogic.org.

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